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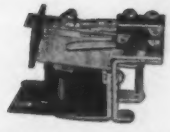

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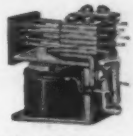

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The Solar System
electronics edition • September 1947

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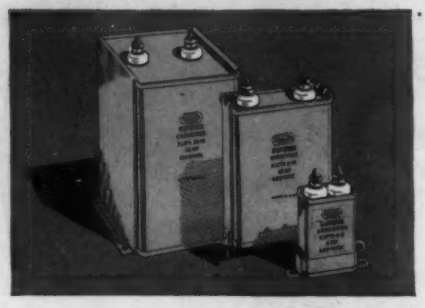

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Capacitors are furnished in gray-lacquered hermetically-sealed fabricated terneplate containers with high-voltage porcelain terminal bushings.

Type XLXPL units are supplied with hook-on flat-flange mounting brackets; Type XLXPJ units are supplied with hook-on spade-bolt brackets; and Type XLXPU capacitors have wrap-around universal mounting brackets.

Maximum capacitances for standard voltage ratings are as follows:

WVDC	Max. Mf.
12,500	0.5
10,000	1.0
7,500	2.0
6,000	2.0
5,000	4.0
4,000	6.0
3,000	8.0
2,500	12.0
2,000	15.0
1,500	15.0
1,000	15.0
600	25.0

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3328

SOLAR CAPACITORS
"Quality Above All"

BUSINESS BRIEFS

By W. W. MacDONALD

Electronic Devices are being used in only five per cent of the applications where they could be useful, according to Edward U. Condon, Director of the National Bureau of Standards.

Several Letters ask if privately-owned shore-to-ship stations are being experimentally licensed for operation in the 152-162 mc maritime mobile band. As near as we can find out the situation appears to be as follows:

Nine frequencies have been allocated for experimental use in that range but no construction permits have yet been granted for private service. A hearing is to be held by the FCC at some future date on applications filed by Foss Launch and Tug of Seattle, Meseck Towing Lines and Moran Towing and Transportation of New York. Decision on this case will probably determine future policy, so the thing for other interested companies to do is to file experimental license applications in Washington on the odd chance that all such applications on hand when the hearing is held may be considered at one time.

Meanwhile, common-carrier experimental maritime mobile station construction permits have been assigned to AT&T in New York, Western Union in the same city, and to Wayne Miller in Mobile, Alabama.

Geological Radio Licenses total 500, as against 131 in 1935. Practically all of them involve a plurality of mobile and portable units, and are held by oil companies or their subsidiaries.

Noise is a major problem in Canada because of the distances radio signals must travel to reach remote areas. The subject came in for a good deal of attention at a recent technical meeting north of our border, with manufacturers of electrical appliances and power companies as well as receiver makers exhibiting great interest.

It is our guess that suppression of noise will soon be a major problem in the United States, what with television and many new mobile services rapidly gaining headway in the field. Cooperative effort will be needed. Count us in.

Packaged Wiring is still very much in the news. We hear that a manufacturer of industrial controls already preforming busbar and dropping it into channels moulded in chassis may soon go a step farther and use sprayed wiring. Experiments conducted in this firm's development lab indicate that it may not be necessary to use a screen process. Sprayed metal sticks to roughened portions of plastic chassis, and is readily wiped off smooth areas.

RCA Earnings in the last 10 years are as follows:

YEAR	GROSS INCOME	PROFIT AFTER TAXES
1937*.....	$112,639,498	8.0%
1938*.....	99,968,110	7.4
1939*.....	110,494,398	7.3
1940*.....	128,491,611	7.1
1941.....	158,695,722	6.4
1942.....	197,024,056	4.6
1943.....	294,535,362	3.5
1944.....	326,421,913	3.1
1945.....	279,503,615	4.0
1946.....	236,980,770	4.6
AVERAGE.	$194,475,505	4.9

* Including foreign subsidiaries. Other years include domestic subsidiaries only.

Factory Tele Service and installation stations may be necessary in order to get the new video business rolling fast and still make equipment stick but many people, including service-wise John Rider, believe that much of this work will ultimately have to be handled by independents.

Remembering our own early radio servicing experiences dating back to 1924, and following conversations with many friends still in the game, we are inclined to agree. And offer television receiver manufacturers this advice for what it is worth: Start out the way you think best. But encourage

FOR { ACCURACY AT HIGH SPEEDS DEPENDABILITY—DURABILITY

an Impulse Register that has gained WORLD-WIDE* RECOGNITION

Cyclotron Specialties Impulse Register was originally designed to meet the exacting requirements of radioactivity research workers. Through outstanding performance, this counter has become a standby in many fields of scientific work *throughout the entire world.*

The Cyclotron Specialties Register is unique in its ability to operate at exceptionally high speeds with complete accuracy and without adjustment or maintenance. It is unexcelled for high speed impulse recording and mechanical operations requiring counting in precise quantities.

The operation of the Cyclotron Specialties Impulse Register is entirely automatic. Originally designed for our Geiger-Muller Counter Sets, it has found numerous other applications . . . both industrial and research . . . including :

PRECISE CONTROL OF QUANTITIES (Packaging, Sorting, etc.)
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Inquiries for special types and modifications including higher speeds, electrical reset, flush panel mounting and higher totalization will now receive prompt attention.

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These are only a few of the many users of Cyclotron Specialties Impulse Registers :

GENERAL ELECTRIC COMPANY
FORD MOTOR COMPANY
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SOCONY VACUUM OIL
WESTINGHOUSE ELECTRIC CORPORATION
BELL TELEPHONE LABORATORIES
U. S. ATOMIC ENERGY COMMISSION
GENERAL FOODS CORPORATION
NORTH AMERICAN PHILIPS CO.
RADIO CORP. OF AMERICA
EASTMAN KODAK COMPANY
DOW CHEMICAL COMPANY
U. S. RUBBER COMPANY
MONSANTO CHEMICAL COMPANY
AMERICAN CYNAMIDE & CHEMICAL CORPORATION
Over 600 American and Foreign Universities

★ SPECIFICATIONS of Cyclotron Specialties Impulse Register No. 401-A

ACCURATELY REGISTERS UP TO 60 IMPULSES PER SECOND

Main, easily-read sweep dial reads 0 to 100 directly. Sweep dial plus auxiliaries read 0 to 9,999 impulses without extra equipment.
4000 ohm D. C. resistance.
Operates on as low as 100 milliwatt.
Small, compact, light weight. Attractive plastic case with non-removable-top binding posts.
Durable, rugged construction to withstand unavoidable accidents.
Dimensions : 3" x 4" Weight : 2 lbs.

Immediate Delivery in Reasonable Quantities

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precision resistors
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AKRA-OHM
WIRE WOUND
Designed to meet
Joint Army-Navy
Specification R-93
PRECISION SWITCHES
350 designs for instruments and high-grade electrical apparatus
HIGH-VOLTAGE TEST EQUIPMENT
1.5 to 200 KV
BRIDGES
Wheatstone, Kelvin-Wheatstone, Percent Limit and Fault Location Types
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VOLTAGE DIVIDERS (DECADE POTENTIOMETERS)
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From 200 micro-ohms to 100 ohms
ATTENUATORS & CONTROLS FOR COMMUNICATION EQUIPMENT
14 standard Shallcross Akra-Ohm Precision Resistors and 50 available adaptations combining special terminals, mountings and power ratings match almost any specification.
Normal accuracy is ±1%. Tolerances as close as 0.05% where required. Temperature coefficient to meet any requirement. All types can be wound with either manganin, copper-nickel, or nickel-chrome or other wire alloys as desired. All designs are highly moisture repellent and unaffected by humidity. Normal resistance ranges from fraction of an ohm to 20 megohms.
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independents to acquire the necessary knowledge in every possible way. Above all, don't cut them off.

Microwave Relay Equipment for television studio-to-station and/or network operation interests at least 25 per cent of the people now on the air, holders of construction permits, applicants, and consulting engineers active in the field. That's the return Philco's Les Woods received when he sent out a letter inviting such people to write in for technical data.

Home Receivers have long been the dog and other communications apparatus the tail in the electronics field. Now it looks like the tail might wag the dog in our time, what with buyer's strikes in the household-goods market as contrasted with continuing healthy demand for transmitters and receivers among old and many new commercial users.

Voltage Regulators are attaining considerable popularity among manufacturers of all kinds of electrical as well as electronic equipment manufacturers for the performance of production tests. So far, we have heard of their use in test setups for vacuum-cleaner motors, switches, circuit-breakers and clocks. Details in our feature pages soon.

It Has Often Been Said that the electronics field sells a lot of things to itself. Nowhere is this fact more noticeable than among manufacturers of transmitters and r-f heating apparatus who happen also to make tubes. Frequently they will be found using tubes made by somebody else. And the answer is that so many types are required that no one manufacturer would find it economically sound to make them all.

Speaking Of Heating, industrial plants in which the use of electronic methods represents a potential saving of, say, 10 percent over older methods should be willing to plough back 2 percent for the services of a maintenance man who really understands the gear. Other-

wise, the saving permitted by the use of electronics may never be achieved.

When plants using steam engines as a source of power added electricity they employed an electrical as well as a steam engineer. They should be equally aware of the fact that a maintenance man who can handle 60-cycle stuff in his sleep may not necessarily feel at home with tubes.

Radio Stations in the United States total 125,000. This figure includes 1,200 broadcast stations (not counting the 1,000 or more under construction) and 70,000 amateur stations.

There are 325,000 licensed commercial radio operators, and 75,000 licensed amateur operators.

FCC Says the nation's seven major networks and 924 a-m broadcast stations had 34,831 employees in February 1947, as compared with the 29,405 employed by nine networks and 876 stations in October 1945.

Employees other than executive, supervisory and non-staff were receiving an average of $61 weekly, for about 39 hours.

Deliveries of domestic broadcast transmitters and studio equipment by RMA member-companies totaled $3,257,394 in the first quarter of 1947, as against $2,863,440 in the last quarter of 1946. (Television equipment sales were lumped into the 1946 figure, are not included in the 1947 report.) A-M equipment sales jumped from $788,567 to $1,135,868, while f-m equipment sales went from $389,844 to $943,812.

Radios In Use in homes, estimated by the FCC's economics division: A-M 66,000,000. F-M 500,000. Television 7,000.

Around This Office nearly everything is translated into electronic terms. The writer, for instance, uses a Norma automatic pencil that has leads of four different colors. The one you want slides into writing position when the proper button is pressed. It is known locally as *the pencil with the three shortwave bands.*

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Beaver gears are used in well-known makes of modern products such as automobiles, radios, automatic tuning devices, phonographs, clocks, timing mechanisms, motion picture projectors, chronometers, gauges, optical equipment, instruments, machine tools, and many others.

Beaver engineers will design or redesign gears to meet performance and cost requirements. Send us your blueprints for estimate.

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Describes Beaver Gear's facilities and methods; also includes table of gear tooth parts.

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The Truth About Electrical Contacts

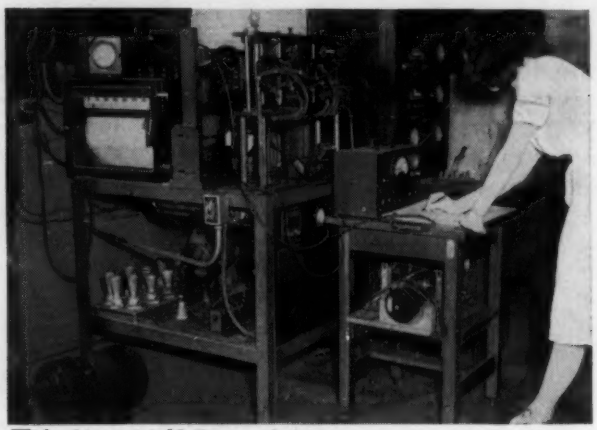
IN theory, the perfect contact material should combine non-sticking properties, low contact resistance, high thermal and electrical conductivity and resistance to electrical erosion with high strength and hardness. In actual practice (because operating conditions differ widely) only two or more of these properties usually predominate.

How does Mallory vary the properties so the *right* ones obtain? You can be sure it isn't done by guesswork. Although Mallory has designed more than 5,000 different contacts—has had 20 years of experience in metallurgy generally—it believes in rigid control. This control is accomplished by a series of spectrographic, chemical, electrical and microscopic tests such as you see at the right. They reveal the truth about chemical composition, grain structure, physical properties.

Yes, you're *sure* when you order Mallory contacts. Furthermore, you get the benefit of Mallory design experience, its manufacturing facilities for producing *every* kind of contact including contact assemblies. What's more, Mallory has developed eight basic contact types that meet thousands of typical applications—save time and money involved in designing "specials." Send for the Mallory Contact Catalog.

This Mallory spectrographic device tests the purity of metals and alloys.

Here Mallory learns what happens in the electrical circuit when contacts are opened and closed.



With this unit, Mallory technicians measure contact resistance and temperature rise.

P. R. MALLORY & CO. Inc.

MALLORY ELECTRICAL CONTACTS & CONTACT ASSEMBLIES

P. R. MALLORY & CO., Inc., INDIANAPOLIS 6, INDIANA

CROSS TALK

▶**JUKE . . .** Last month we rejoiced in Dr. Harry Olson's experiments (p 80, August) that proved that audiences like to hear popular music as played, without acoustic modification of the tonal range. But that was at Princeton. Now comes word from Minnesota that chills us to the bone. It seems that a bass viol player in one of the leading local dance orchestras approached the chief engineer of the Minnesota Electronics Corporation, asking for an amplifier system to be used with the orchestra. The specifications were usual in all but one respect. Said the viol player, "Build the amplifier so my bass viol will sound as loud as those you hear in the juke boxes. People are so used to the oomph, oomph of the boxes that they now squawk for it in the band!"

This is, to be sure, an isolated instance. But we hope we hear of no more like it. As an arbiter of musical taste we think an RC time constant is a very poor substitute for a bandleader's judgment. We rally to the Olson standard, while there is yet time.

▶**TALK . . .** The late Professor Kenrick made quite a stir many years ago when he devised a gadget for attachment to a radio set which would silence the spoken word. It worked on any break in the modulation, such as occurs between words. Last month a Reuters dispatch from Johannesburg revealed that a South African gadgeteer had invented a similar device, but with a fiendish twist. His gadget automatically tunes in another station when the spoken word interrupts the program for more than a minute. Coming at a time when competition has reared its head in the broadcasting business as never before, this incident should remind us all that there's very little wrong with the broadcasting game that less talk (and better talk) wouldn't cure.

▶**TELEBANK . . .** Firm in our resolve to keep our readers informed of new and interesting applications of television in the industrial and commercial fields, we have to report the Franklin Square (Long Island) Bank. A television system between the teller's window and the bookkeeper department provides quick identification of check signatures. This saves the depositors' time, probably doesn't cost too much, and may stay in use for some time. But we'd like the story much better if it weren't for the fact that the same bank has a children's department with a low counter, and offers a lollipop to each child making a deposit. We understand they use a *miniature* image orthicon at *that* window.

▶**HIGHBROW . . .** One of our staff, in recommending against publication of a paper dealing with operational calculus (on the score that few readers could follow it), delivered himself of an intra-office opinion which we think should arouse some argument. Our man claimed that only one percent of the members of our profession understand operational calculus, important tool that it is, well enough to use it in their work. Moreover he says, "Those in the saddle can expect to be displaced by young upstarts that come out of college knowing operational techniques and capable of using them. Moreover, the run-of-mine engineers in small companies cannot expect to compete, or enable their companies to compete, with the big operators who can afford a few of the highbrow boys to do this sort of stuff. What industrial and economic trends may be caused by this condition is anybody's guess." If this thesis is valid, and we think it is, it's food for plenty of thought. The number of practicing engineers who can use advanced calculus and vector analysis in the design of circuits and antennas is small, far too small in relation to the need for such ability. Will the older, less knowing group endanger their jobs or their companies' competitive position by speaking in cycles and decibels, rather than in Fourier transforms and dels? Maybe so. In any event, the future is for the bright young lads. The new crop just graduated, we hear, is a very superior grade of hombre, older than usual, more serious and better trained.

A MEMORY TUBE

Pattern is produced, stored, and scanned on dielectric screen by electron beams. Special cathode-ray tube can be used to store for protracted periods television pictures, radar indicator patterns, oscilloscope traces, or other information

By ANDREW V. HAEFF

Naval Research Laboratory
Washington, D. C.

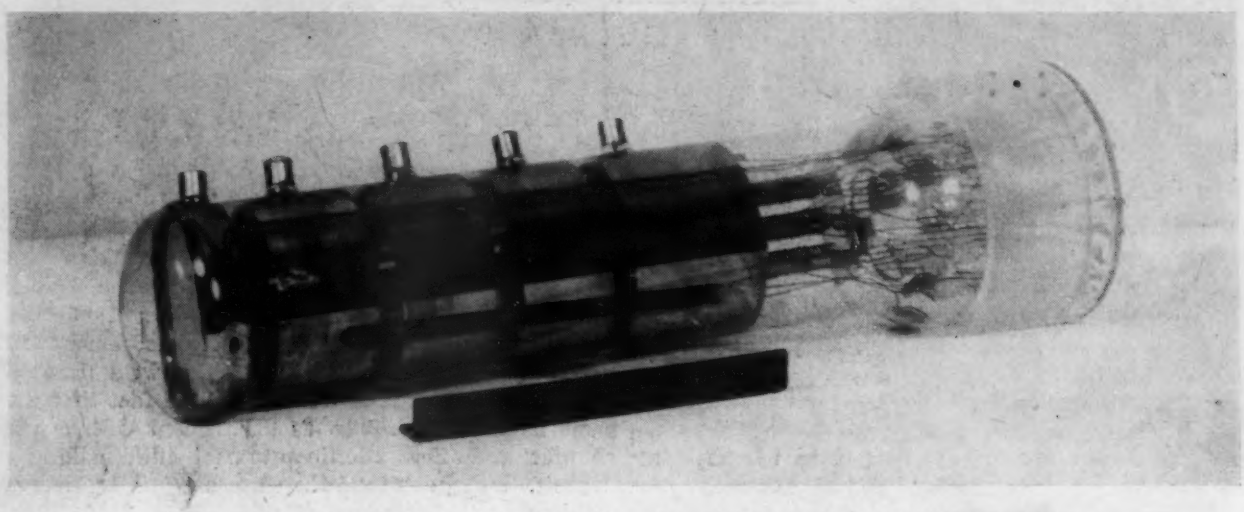


FIG. 1—By constant electronic bombardment, image is indefinitely retained on dielectric screen

AN ELECTRONIC TUBE having a controllable duration of image storage has been experimentally developed. The tube, shown in Fig. 1, can be used for rapidly recording electrical signals, storing them for any desired period, and simultaneous multiple reproduction of the recorded signals. Specific applications of this storage tube include use in simultaneous multicolor and three-dimensional presentation of radar or sonar data, daylight viewing, automatic recording and reproducing electrical transients, and storing and reading binary numbers of electronic digital computers.

Storage of Charges

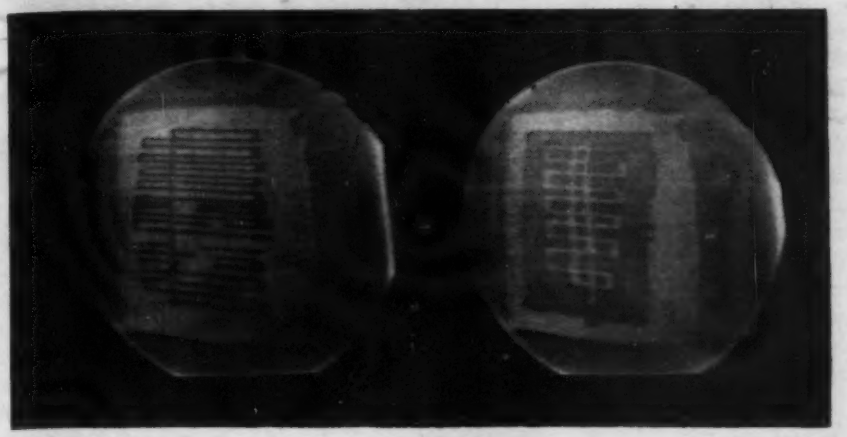
Operation of the tube is based on secondary emission properties of a dielectric target bombarded by electrons. Three electron beams of different energies scan the target. One beam writes or paints the stored image or picture, another holds the recorded information, and the third reads or electrically reproduces the recorded information.

A pattern of charges deposited on the surface of a good insulator can be retained for a long period. Deterioration of the pattern or electrical picture finally occurs from leakage of the charges over the surface and through the body of the insulator, or due to irradiation of the surface, either by a corpuscular beam (with associated secondary emission effects) or by electromagnetic waves (photoemission and photoconductivity). Leakage effects can be minimized by using a dielectric of extremely high resistance.

If it is desired to generate electrical signals which correspond to the stored-charge pattern, it is necessary to explore the surface of the dielectric point by point. This is most conveniently done by scanning the surface with a focused electron beam. This process, however, usually results in a complete erasure of the stored picture after a few scans, as is commonly observed in television pickup tubes.

In the novel storage tube here described, the effect of erasure is completely neutralized by constant bombardment of the surface by a stream of electrons of suitable energy, thereby continuously regenerating the picture. Thus the picture can be retained for any desired period of time, even though it is being continuously scanned in order to obtain electrical output.

A schematic representation of the basic memory tube is shown in Fig. 2. The cathode-ray type tube must have at least two independent sources of electrons and a suitable



Depending on relative tube voltages, writing can be either black on white or white on black

Using potentiometers controlled by position of writing stylus to produce vertical and horizontal deflection signals, pattern can be traced stroke by stroke and all lines forming the writing or picture are retained by screen

dielectric target on which the charge pattern can be stored. In front of the target is a metal screen or collector. The velocity of the electrons while passing the screen and approaching the target is determined by the potential difference between the screen and the particular cathode from which the electrons originate The electron beams will be referred to as the writing, holding and reading beams.

Functions of Beams

Writing is done with a focused beam of relatively high current density. This beam writes the desired electrical picture by being directed through a deflection system to different parts of the target and by being modulated either in intensity or velocity. The elements of the target explored by the writing beam either lose or accumulate electrons and thus assume a potential different from that of the unexplored part of the target surface.

The function of the holding beam is to retain the painted picture for a desired period of time by spraying the entire area of the target with electrons of suitable velocity and in sufficient numbers. This holding can be accomplished either by a diffused beam bombarding the whole surface simultaneously or by a focused beam which scans the surface in a selected time sequence.

The reading beam is a focused beam of constant intensity that scans the dielectric surface. In bombarding the target surface, the reading beam generates secondary emission current, the instantaneous magnitude of which depends upon the potential of the particular element of the target being scanned at that moment. This secondary current is partially intercepted by the metal screen in front of the dielectric target. Thus the current flowing to this metal screen constitutes the electrical output of the device. The output current is fed through a suitable amplifier to the grid of a conventional cathode-ray indicator tube, the electron beam of which is made to scan in the same manner as the reading beam of the memory tube. The intensity modulation of the scanning beam of the indicator tube is converted into light-intensity variation over the surface of the cathode-ray screen, and thus the electrical picture painted on the surface of the storage tube target becomes visible on the face of the indicator tube.

Target Characteristics

The mechanisms of writing, holding, and reading are based on the

secondary-emission properties of the target. Figure 3A illustrates a typical characteristic of a dielectric surface under bombardment by electrons. At low primary energies (when the incident velocity is less than that corresponding to the critical voltage) the net current to the target is negative, so that the target element bombarded by low velocity electrons tends to accumulate electrons and thus to assume a more negative potential. At the critical voltage, corresponding to unity secondary emission ratio, the net current to the target is zero because as many electrons leave as arrive. At higher energies the net current to the target is positive.

If the potential of the collector is below the potential of the target element, then the secondary electrons, under the influence of the retarding field of the collector, will be returned to the target surface. On the other hand, the secondary electrons flow to the collector where its potential is higher than the target potential. The net current to the target element as a function of its potential can be represented as is shown in Fig. 3B.

When a stream of electrons is directed towards the target surface, any element of the target which has a potential below critical value tends to accumulate a negative charge and will finally reach the cathode potential. Any element charged to a potential above the critical value will tend to approach the potential of the collector. Thus the cathode and the collector potential are two stable points at which the target elements can maintain themselves under continuous bombardment by a stream of electrons. Any tendency to lose the charge, due either to leakage of the dielectric or to bombardment from other sources, will be opposed by bombardment from the first source. This characteristic is made use of in the memory tube for maintaining the painted picture for any desired period of time.

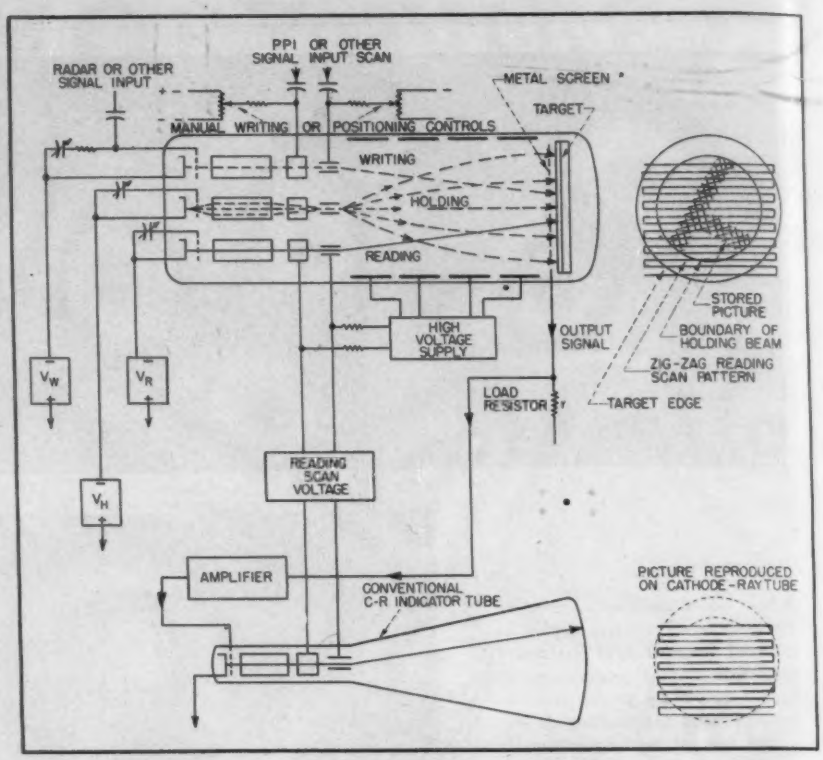


FIG. 2—Memory tube stores picture which can later be reproduced on other cathode-ray tubes

Producing the Image

To paint the electrical picture, the target is scanned with a focused electron beam from a separate cathode and at such velocity and current density that it can override the tendency of the holding beam to maintain the potential distribution over the target surface. As the writing beam is moved over the surface, it leaves a trace which assumes collector potential. When the target surface has luminescent material deposited on it, the positively-charged trace of the writing beam becomes continuously visible under bombardment by the holding beam. The remaining portions of the target remain dark because only the positive areas of the target are bombarded by electrons of sufficient energy to excite the phosphor. This process of painting the electrical picture is referred to as analogous to writing with white chalk on a blackboard.

If the cathode producing the writing beam is maintained at a potential below the collector potential by an amount less than the critical energy, the target elements under its bombardment tend to change their potential in the negative direction. Thus, if the surface of the target was originally charged to the collector potential, the writing will appear in the form of dark lines. This painting is analogous to writing with a black pencil on white paper. Thus by the use of a suitable phosphor on the target surface, it is possible to observe the electrically painted picture directly on the target of the storage tube. In spite of the relatively low electron velocities thus far employed in tests, the picture can be made sufficiently bright with moderate currents because the bombardment and luminescence take place all of the time.

Electrical Output

For many applications it is convenient to obtain electrical signals from the memory tube that correspond to the stored picture. Then the picture can be reproduced in television fashion on the face of any number of conventional indicator tubes.

To obtain electrical output from the memory tube, a third electron beam is used. It scans the surface of the target in television fashion or in the zig-zag pattern shown in Fig. 2. Velocity of the electrons of the reading beam is relatively high, thus their bombardment of the surface produces appreciable secondary emission current from the target. The reading beam is adjusted to give maximum signal output without disturbing the potential

distribution on the target surface, which has some areas at the potential of the collector and some at the potential of the cathode of the holding beam.

This process of reading is analogous to scanning the target of a monoscope tube used in television. The net secondary current flowing from positive areas is different from the current flowing from negative areas so that scanning of the areas by the reading beam produces variations in current to the collector electrode. These current variations are converted into voltage variations in the load resistor and, after suitable amplification, are used to modulate the beam of a conventional indicator tube, which reproduces the picture in television fashion.

If the velocity of the reading beam is adjusted to a value required for the holding beam, then the continuously scanning reading beam can perform the holding and the reading functions simultaneously. For this method of operation only two guns are needed, one for writing and one for holding and reading.

Image Resolution and Stability

High stability of the image, and high signal output were obtained by operating the tube in either the two-beam or the three-beam mode. With writing beam currents of the order of microamperes, and the beam diameter of the order of one millimeter, the writing speed is of the order of several kilometers per second. An increase in current results in a corresponding increase in writing speed. Thus it is possible to record very fast transients even with moderate beam currents.

The number of discrete elementary areas which can be resolved on the storage tube target is determined primarily by the spot diameter of the writing beam. It is expected that with well focused beams about 10,000 elements can be resolved.

Because differences of potential between positive and negative elements of the target are of the order of a hundred volts, corresponding differences in secondary emission currents are quite high. Thus, scanning the target by the reading beam produces current variations of the order of the reading beam current itself. With a load resistance of the order of 1,000 ohms, a signal output of the order of several millivolts is obtained with the reading-beam current of several microamperes, so that a simple amplifier of rather low gain can be used satisfactorily.

High resistivity of the target material contributes to a very high stability of the electrical image. It is possible to observe the image, which is continuously scanned by the reading beam, for hours without noticeable deterioration of detail. This stable condition is realized over a considerable range of current densities and electron velocities of the holding beam. If the potential of the holding beam cathode exceeds the value corresponding to the condition of stable operation, excessive potential gradients along the dielectric surface, in the regions of transition from positive to negative areas, result in spreading of the positive areas over the rest of the target surface. On the other hand, when the holding beam velocity approaches the critical value, the positive areas tend to shrink and the picture is gradually washed away. Lowering the holding beam velocity below the critical value makes it possible to erase the picture rapidly. A suitable adjustment of the holding beam velocity thus makes it possible to obtain the desired persistence of the image; in other words to control the effective memory of the device. Several other methods of control of memory, such as pulsing of the holding beam in a specified manner, also appear promising. The adjustable persistence rather than infinite persistence is extremely useful for such applications as radar target presentation where it is desirable to plot automatically the track of the moving target for a specified time so that the length of the track will be proportional to the velocity of the target. This development thus presents a promising solution of the problem of controllable persistence which has not been solved by development of special phosphors.

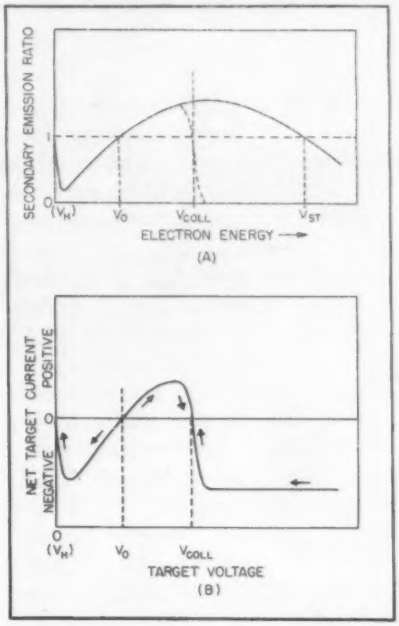


FIG. 3—Secondary emission phenomena give the dielectric target its required characteristics

In studying the operation of the memory tube under different conditions it was observed that, in addition to the normal mode of operation which is explained on the basis of the secondary emission characteristics, it is possible to operate the tube under several different conditions. This observation indicates the existence of additional stable points and a rather complex nature of the secondary emission characteristic.

Other interesting effects, such as generation of charge waves propagating over the surface of the target at rather low speeds, appearance of scintillations or brightly lighted spots apparently associated with ion bombardment, and other effects, were observed in connection with this study. Their description lies beyond the scope of this paper but their mention is intended to stimulate interest in the use of the memory-tube technique for general studies of the secondary emission and other properties of dielectrics.

The author takes this opportunity to express his appreciation of the many contributions of his colleagues and assistants in this project, Mr. F. H. Harris and Dr. S. T. Smith, who by their enthusiasm, resourcefulness and industry helped to achieve the successful solution of many difficult problems which were encountered in the course of this study, and to thank the other members of the Laboratory who gave their whole-hearted cooperation in this project.

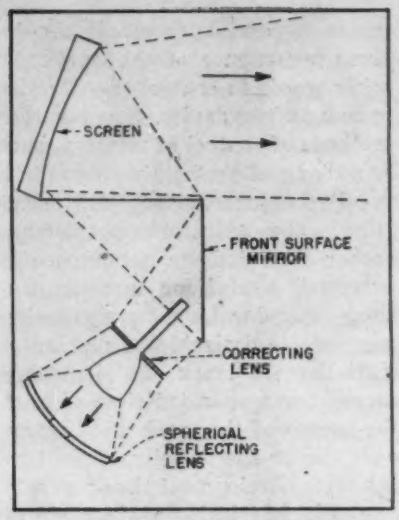


FIG. 1—Elements of the projection system, viewed from the side of the cabinet

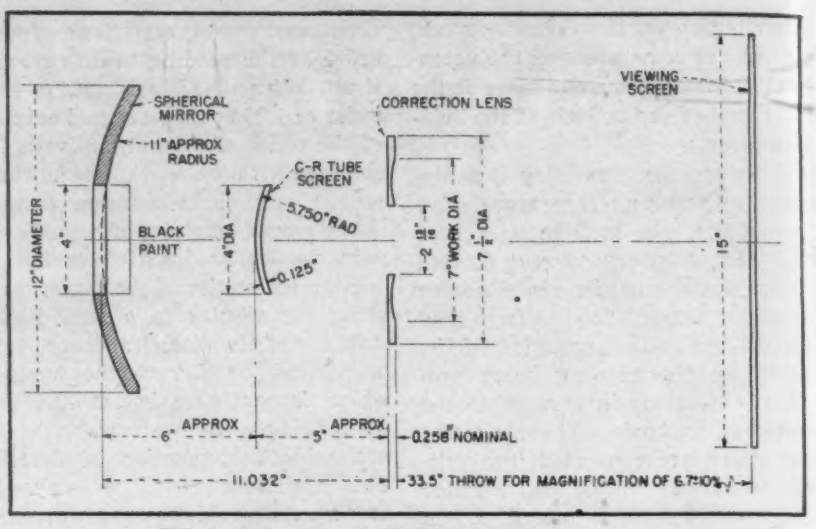


FIG. 2—Dimensions of the Schmidt projection system, arranged for rectilinear throw

A New Television PROJECTION SYSTEM

Combination of a Schmidt optical system, a new phosphor, directional viewing screen, keystone projection, and ingenious cabinet arrangement produces a 15 by 20-inch picture of exceptional brightness and contrast

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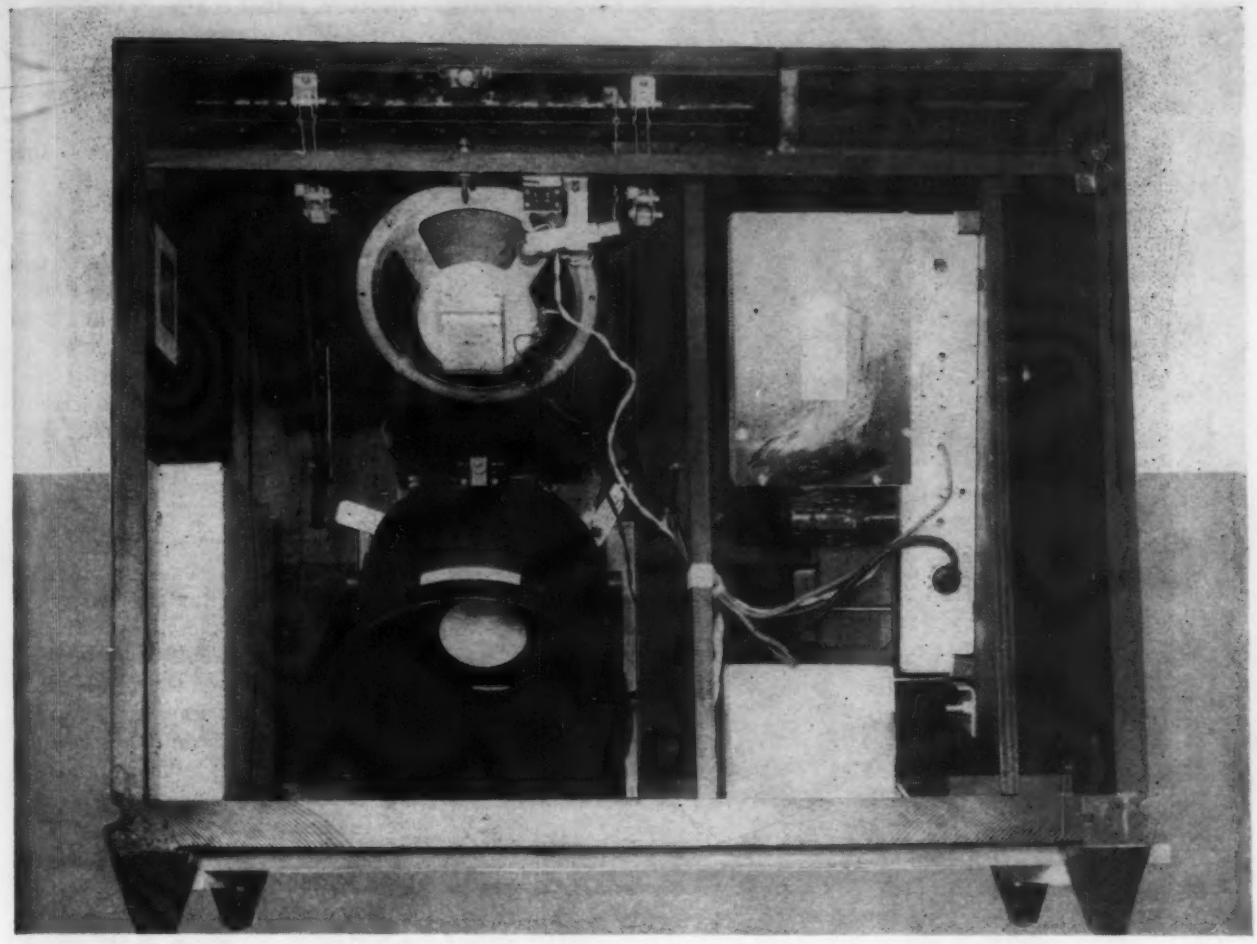
EARLY this year, members of the Federal Communications Commission[1] saw a developmental model of a new television receiver, producing a projected picture of unusual brightness and contrast, and having great resistance to the effects of external light. This receiver, developed by Philco Corporation engineers, has now reached the production stage and was announced to the trade in June as model 2500.

The receiver produces a picture 15 by 20 inches in size, of excellent brightness and contrast. The area contrast of the picture is approximately 70 to one, and the picture may be viewed without loss of contrast in an ambient light level of well over 100 footcandles (such as is produced in a living room by direct sunlight through several unshuttered windows). This exceptional performance is made possible by a new projection technique described in this paper.

The elements of the projection system are shown in Fig. 1, which represents a side view of the cabinet. The 4-inch projection tube produces on its screen an image, about 2¼ by 3 inches in size, with a highlight brilliance of about 2,000 footlamberts (aluminum-backed tube). The light is collected by the front-surface spherical mirror which reflects it through the correction lens. This lens removes the spherical aberration introduced by the mirror and passes the corrected image to the front-surface mirror on the inner surface of the cabinet. This in turn reflects the image to the under side of the cabinet lid, which is covered with a directional viewing screen of unusual construction.

The screen confines the major portion of the light, reflected forward to the audience, to a solid angle 60 degrees wide by 20 degrees high. This beam-projection system introduces an optical gain of some 10 times relative to a perfectly diffusing screen, with corresponding increase in brightness. Great care has been taken in the cooperative design of the optical system, including the screen, to eliminate the formation of a dark central zone in the viewing area, from which the picture cannot be viewed properly.



Rear view of Philco model 2500 projection receiver, with spherical mirror removed from housing (lower left). The chassis at right tunes eight stations, contains 29 tubes

This zone has been one of the principal objections to systems employing a directional screen and a centrally vignetted light source. The screen is also designed to reflect a negligible amount of light which may fall on it from the surroundings, so the contrast of the image is preserved even in a brightly-lighted room.

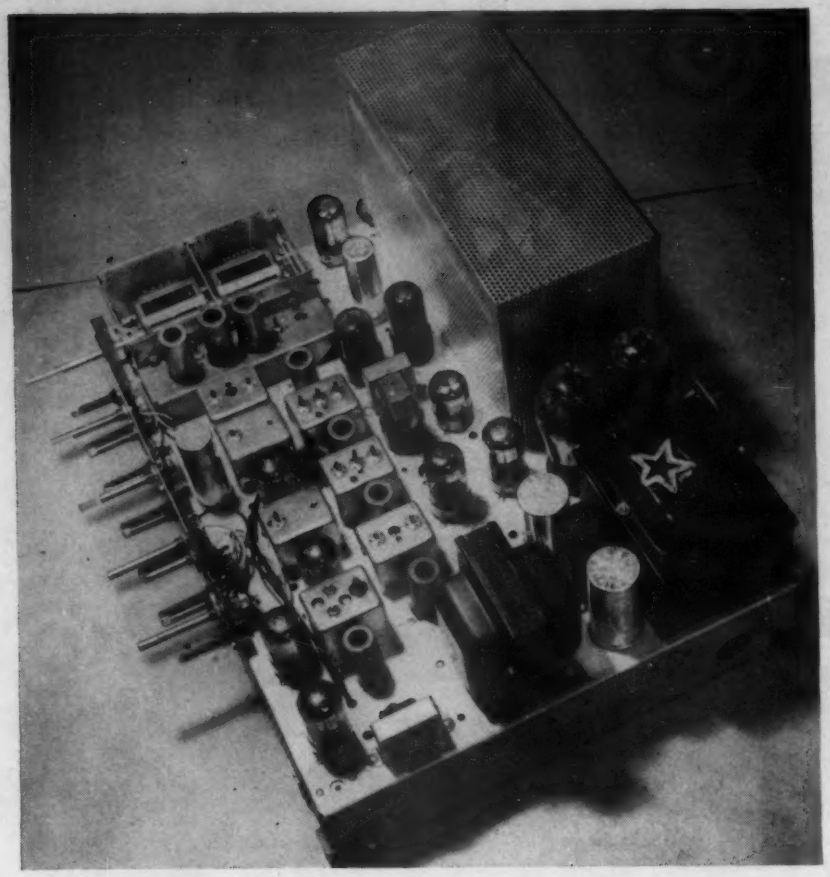
Inspection of Fig. 1 shows that the optical distance from the screen of the c-r tube to the top of the viewing screen is somewhat greater than the corresponding distance from the tube to the bottom of the screen. Thus, if the image on the face of the c-r tube were rectangular, the image viewed on the screen would be wider at the top than at the bottom. To correct this geometric distortion, a compensating geometric distortion of the image is produced on the c-r tube. This system, as described later, employs fixed magnetic deflection elements near the c-r tube, and does not involve any modification of the deflection circuits, which are those that would produce a conventional rectangular pattern. In addition, compensating corrections of the optical system need to be incorporated.

Details of System

Details of the elements of the projection system (projection tube, Schmidt system, viewing screen and keystone projection system) are described below.

The projection tube screen has a four-inch diameter face, and a radius of curvature of 5.75 inches, with a thickness of 0.125 inch. The tube operates at 20 kilovolts. It employs a new phosphor which produces a truly black and white picture when operated at the correct anode voltage. Moreover, the color of the fluorescence is independent of beam current, so the picture does not display the characteristic yellow tinge in the highlights, typical of earlier projection tubes. Screens both with and without aluminum backing have been produced. The aluminum-backed tube gives greater brightness and contrast at the 20-kilovolt level, and is now being used in production.

The electron gun of the tube operates with an average beam cur-



Chassis of the projection receiver, showing the rotary mechanism of the station selector (top, left). The receiver contains 29 tubes

rent of 80 microamperes, with 600 microamperes peak in the highlights. It requires a swing of approximately 80 volts. The beam cuts off at −100 volts ±30 volts.

Wide Aperture Optics

The dimensions of the modified Schmidt projection system are shown in Fig. 2, arranged in the conventional rectilinear form. The action of the correction lens can be seen from Fig. 3 and 4. The point *O*, on the surface of the c-r tube screen emits light along the typical rays 1, 2 and 3. After reflection from the spherical mirror these rays follow the paths 1′, 2′ and 3′. If these rays continued without correction they would impinge on the optic axis at widely differing points *A*, *B*, and *C*. Consequently at the image plane *I* the point object *O* would be reproduced as a circle of light. This defocusing, an effect known as spherical aberration, can be removed by the correcting lens.

The correcting lens is so formed and positioned that a reference ray, say number 2, passes through it undeflected, reaching the optic axis at the image plane *I*. A ray closer to the axis, say number 1, passes through a portion of the correcting lens having positive curvature and so is bent toward the axis along the path 1″, falling on the axis at *I*. Another ray, say number 3, further from the axis, passes through a portion of the correcting lens having negative curvature and so is bent outward along ray 3″, reaching the axis also at *I*. Hence all rays starting at the point *O* come to focus at the same point *I*. If other points of the c-r tube screen are arranged along a curved surface, as in Fig. 1, the resulting image lies in a flat surface. Theoretically the surface of the c-r tube screen should be an ellipsoid, but in practice a spherical surface suffices.

The correcting lens may introduce other aberrations of its own, but by using a lens of low power, the aberrations may be kept negligibly small. It must be remembered that the 525-line television image is, in itself, rather coarse-grained compared to the images usually met in projection systems, so that the allowable defocusing and other aberrations in the projection system are greater than they would be if the image had finer detail.

Calculations show that the radius of curvature of the c-r tube screen should generally be 0.53 times the radius of the mirror. In the Philco system, these radii are 11 and 5.75 inches respectively. The throw of the system (distance from the correcting lens to the viewing screen) is 33.5 inches. The Schmidt system is one which must be designed for a particular value of magnification.

The efficiency of the optical system is here defined as the fraction of the total light flux emitted by an element of the c-r screen at the center, which the optical system accepts and focuses on the corresponding point of the image. The efficiency is equal to $\sin^2 u$, where u is the semi-apex angle shown in Fig. 3. The corresponding f/number of the lens is $1/(2 \sin u)$. The f/number of the system must be used with caution, however, in comparing the system with a lens, since a portion of the center of the spherical mirror is masked off (actually painted black to reduce halation) because light reflected from this region is blocked by the c-r tube.

In this particular system the efficiency is about 30 percent, taking into account the central masking. The corresponding f/number is f/0.8 and the linear magnification 6.7 times. The actual dimensions of the c-r tube image are 2-1/16

Projection mechanism

inches high by 2-9/16 inches wide at the top and 2-31/32 inches wide at the bottom. When magnified by a mean value of 6.7, these dimensions become 15 by 20 inches, as viewed by the audience on the screen.

Keystone Correction

The relative positions of the various parts of the system are shown on Fig. 4. To produce proper focus over the area of the image, the tube axis is placed at an angle β to the optic axis, such that $\tan \beta = m \tan \alpha$ where α is the angle shown in Fig. 4, and m is the reciprocal of the magnification. In the Philco system, α is 24°30′, β is 3°54′ and m is 1/6.7.

As previously mentioned, when the projection system is placed in the cabinet as in Fig. 1 the unequal distances between top and bottom of the viewing screen and the c-r tube produce geometrical distortion. To correct the distortion, the scanning pattern on the screen of the c-r tube must have a trapezoidal shape, as shown in Fig. 5. This shape might be introduced by modulating the amplitude of the horizonal scanning wave (this is the technique used in conventional iconoscope cameras), but a simpler system was devised using magnetic pole pieces at the edges of the c-r tube screen. The arrangement is shown in Fig. 6. Two bar magnets are fixed to opposite sides of the tube. Two pole pieces *A* produce lines of flux across the tube, just inside the screen. These lines are essentially straight and parallel. They cause the scanning beam to be deflected sharply upward just before it hits the phosphor. Consequently the scanning pattern is laid down on the phosphor at an angle and its shape, when projected on the tube face, is trapezoidal.

The sharp deflection imposed on the scanning pattern causes the pattern to fall on the screen in an off-center position. This is corrected by a supplementary magnetic field, produced by the fringing field between the "point" poles *B* at the opposite ends of each magnet. The flux lines of this field are curved, and they are opposite in direction to, as well as weaker than the lines produced by the pole pieces *A*. Each field produces curvature and displacement in addition to the keystoning. By adjusting the angle of each bar magnet, the curvature and displacement are cancelled, while the necessary amount of the keystone correction remains.

Independent adjustments of the position of the cathode-ray tube are provided axially and about two axes perpendicular to the optic axis. These adjustments permit sharp focus to be obtained within the range of magnifications for which the optical system is designed.

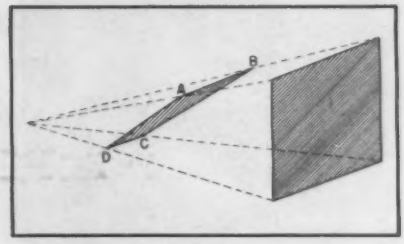


FIG. 5—Projection of a trapezoid at an angle produces a rectangular image, correcting keystone distortion introduced by folded geometry of projection system

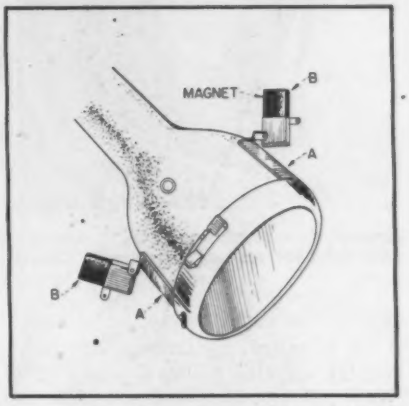


FIG. 6—Bar magnets and pole pieces for producing trapezoidal pattern on c-r tube screen

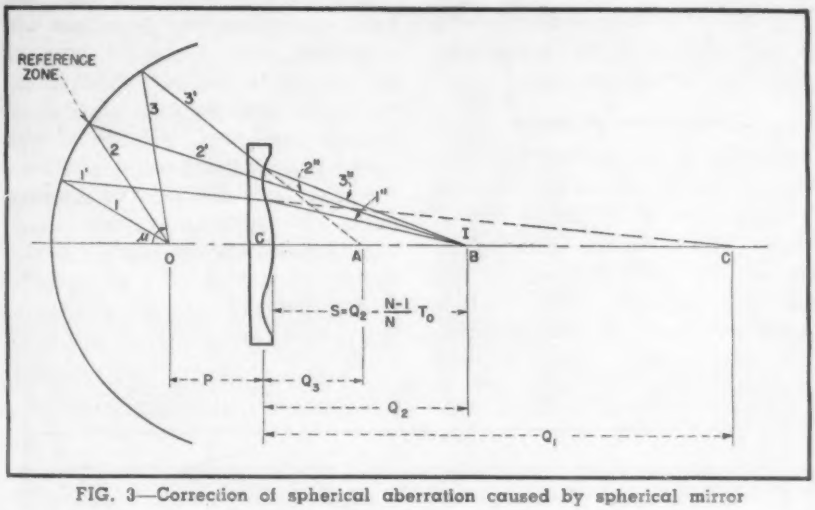


FIG. 3—Correction of spherical aberration caused by spherical mirror

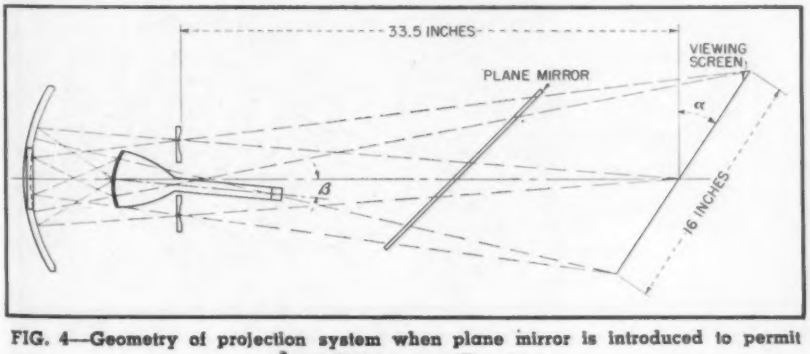


FIG. 4—Geometry of projection system when plane mirror is introduced to permit installation as in Fig. 1

Viewing Screen

The high brightness and contrast of the image are accounted for by the viewing screen on which the image is projected, and its inclination. Conventional viewing screens, such as are employed in motion picture houses, are non-directional, that is they reflect an equal amount of light flux in all directions. Such screens are suitable if the incident light is very bright, and if it is

Phileo system is not equal in both coordinate directions, the dark spot is not circular. It is, rather, elongated horizontally throughout the width of the viewing area, and it possesses appreciable illumination, although distinctly less than the top and bottom areas.

This rather serious defect of the system has been overcome by covering the viewing screen with many millions of lens-like hemispherical droplets, each of which redistributes the light falling on it throughout a narrow predetermined angle. The angle, by suitable choice of the size and refractive index of the droplets, causes each elemental reflection to spread out so that it reaches the central region of the screen, which would otherwise remain under-illuminated. The droplets thus diffuse the light, but by an amount small compared with that produced by the vertical grooves. Hence the tendency to increase the width of the beam in the horizontal direction is negligible.

The droplets are formed by applying a lacquer, of proper index of refraction, in a fine mist-like spray. To keep each droplet distinct and separate from its neighbors, several incomplete sprayings are applied, with time between each to permit drying. Hence one set of droplets does not coalesce to those previously applied.

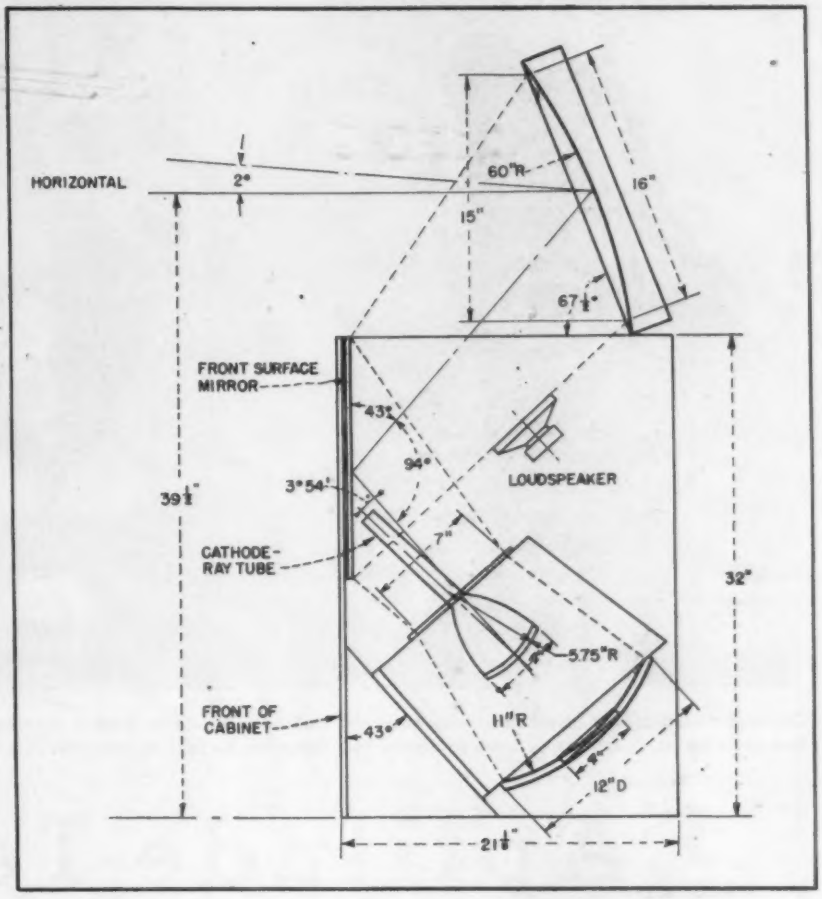


FIG. 9—Dimensions of the projection system, as viewed from side of cabinet. Loudspeaker is located close to center of image, adding to realism

Cabinet and Chassis

A side view of the chassis with principal dimensions is shown in Fig. 9. One important item in the arrangement is the loudspeaker, which is so positioned with respect to the viewing screen that sound reflected from it appears to emerge directly from the center of the screen. This assists materially in creating the illusion of reality. To maintain the elements of the optical system in proper relation, it is essential that the lid of the cabinet be raised to the precise angle of proper inclination and remain in that position rigidly. This is accomplished by a rugged spring-loaded elevating mechanism with a positive stop.

The electrical system of the receiver includes 29 tubes, five of which are rectifiers. Five manual controls are provided, for station selection, contrast, brightness, volume, and tone. Provision is made for eight r-f channels, installed at the factory to meet the particular situation in each major market. The i-f bandpass is approximately 4.0 mc, and the sensitivity is approximately 100 microvolts for a solidly synchronized picture. Automatic gain control is provided in the picture circuit. One unusual electrical feature is a stabilized synchronization circuit which, in addition to maintaining sync in high noise level, eliminates nonlinear scanning and defocusing of the beam.

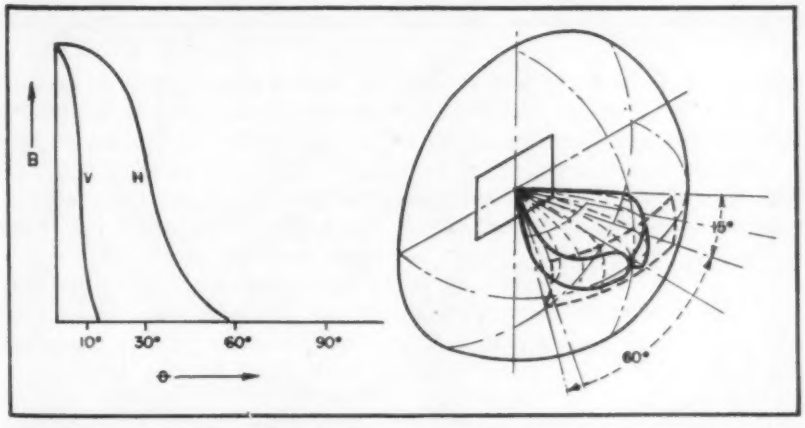


FIG. 8—Rousseau and angular diagrams of the directional pattern of Philco viewing screen

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(1) FCC Views Television Advances, ELECTRONICS, p 80, March 1947.
(2) Maloff and Epstein, Projection Television, *Jour. S.M.P.E.*, June 1945.
(3) Maloff, and Epstein, Reflective Optics in Projection Television, ELECTRONICS, p 98, Dec. 1944.

Closeup view of turret switching arrangement used for station selection in the receiver

required to supply equal brightness to all viewers regardless of their position relative to the screen. But since television receivers must often be operated in brightly-lighted rooms, it is important to conserve light. This can be done by imparting a directional characteristic to the viewing screen.

Light Distribution

A typical Rousseau diagram of a directional screen is shown in Fig. 7, together with the light distribution as a function of the angle. This ideal screen radiates equal flux within a 20-degree vertical angle and a 60-degree horizontal angle. The reflected picture is invisible outside these limits, but is equally bright wherever viewed within them.

The choice of the angular dimensions of this ideal screen is based on statistical studies of Professor Hooton of Harvard University carried out for the Army Air Forces. In this study the anatomical statistics of male and female pilots were studied to determine their seated heights. This information, combined with tests to determine the optimum height of a television viewing screen above the floor, leads to the conclusions that the viewing space can be compressed to a 20 by 60-degree solid angle. If all the light impinging on a screen is reflected in such a pattern, the image will be 17.2 times as bright as if reflected from a perfectly diffusing, nondirectional surface.

Since the desired beam is wider than it is high, it is not possible to use symmetrical optical elements, such as small glass beads, to produce the beam. The asymmetrical nature of the pattern also makes it possible to introduce another important property, namely to reduce the susceptibility to the effects of external light. Light entering the system from sources outside the reflected beam is not reflected back to the audience, and the picture thus retains its contrast even when viewed in a high ambient illumination.

The actual directional characteristic of the Philco screen is shown in Fig. 8. The major portion of the illumination, as shown in the Rousseau diagram at the left, is contained within plus or minus 10 degrees vertically and 30 degrees horizontally, producing an overall projected beam 20 degrees high by 60 degrees wide. The illumination drops substantially to zero at plus or minus 15 degrees vertically and plus or minus 60 degrees horizontally. Tests with this distribution have shown it to be thoroughly adequate for living-room use.

Construction of Screen

The screen consists of a curved sheet of metal, formed as a section of a cylinder of 60 inches radius. This is fitted under the lid of the cabinet with the axis of the cylinder horizontal. On the surface of the screen are many thousands of vertical grooves. These grooves have a random distribution of sizes and cross-sections, but the net result is as if they had cross-sections in the shape of minor arcs of circles. The curvature of the surface tends to confine the reflected light into a narrow beam, measured vertically, while the grooves tend to spread the light over a horizontal sector measuring some 60 degrees. The radius of curvature of the surface is twice the product of the projection throw and the viewing distance, divided by the sum of these distances.

One objection to the use of such a reflecting surface, when an ordinary lens system is used, is that the beam is confined excessively in height. When a Schmidt optical system is used, the light spreads through a wider vertical angle, but another difficulty arises. This is the formation of a dark spot at the center of the screen, due to the fact that the effective light source is a correction lens of annular shape (no light comes from the center of the source). Although each point of the correction lens illuminates the whole viewing screen, the relative directions and angles of the rays from correction lens to screen are such that each elemental area of the picture is lacking in rays directed toward the central zone of the viewing area. Hence if one views the screen from a point on the optic axis the picture appears dark, with somewhat greater illumination toward the edges. If the picture is viewed at a point sufficiently removed from the optic axis, it appears uniformly illuminated. Since the optical power of the

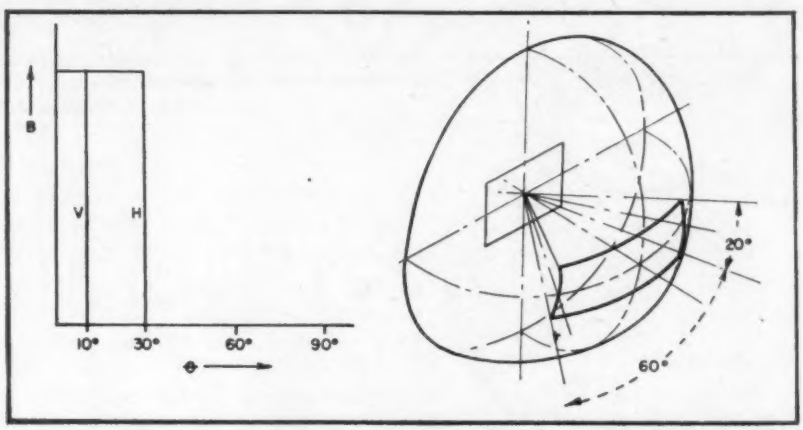


FIG. 7—Rousseau diagram (left) and angular distribution of ideal beam-projection directional viewing system, having a gain of 17.2 ovef a perfect diffusing surface

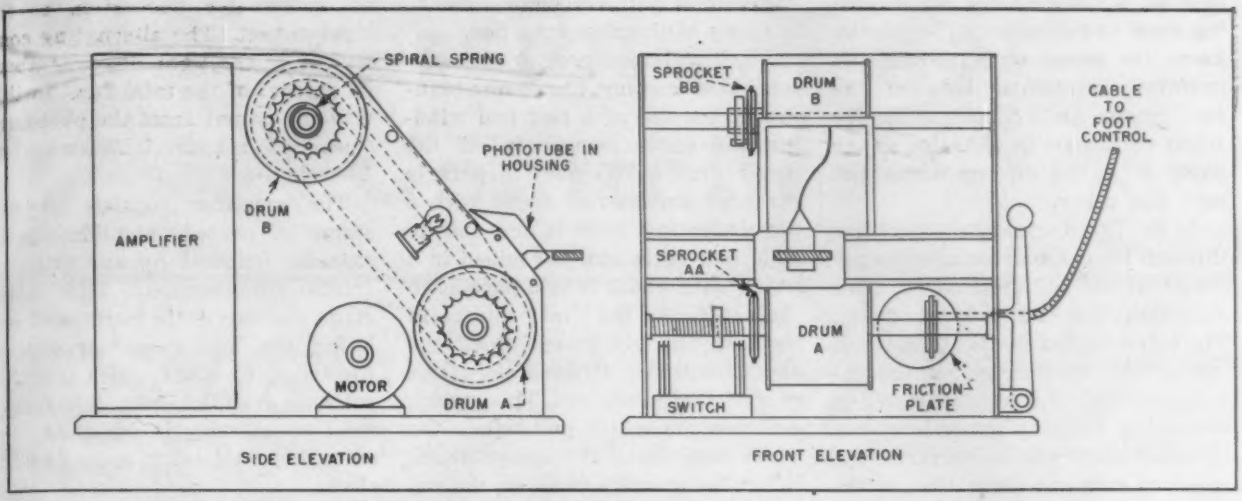


FIG. 1—Mechanical arrangement of photoelectric aptitude tester for sewing machine operators

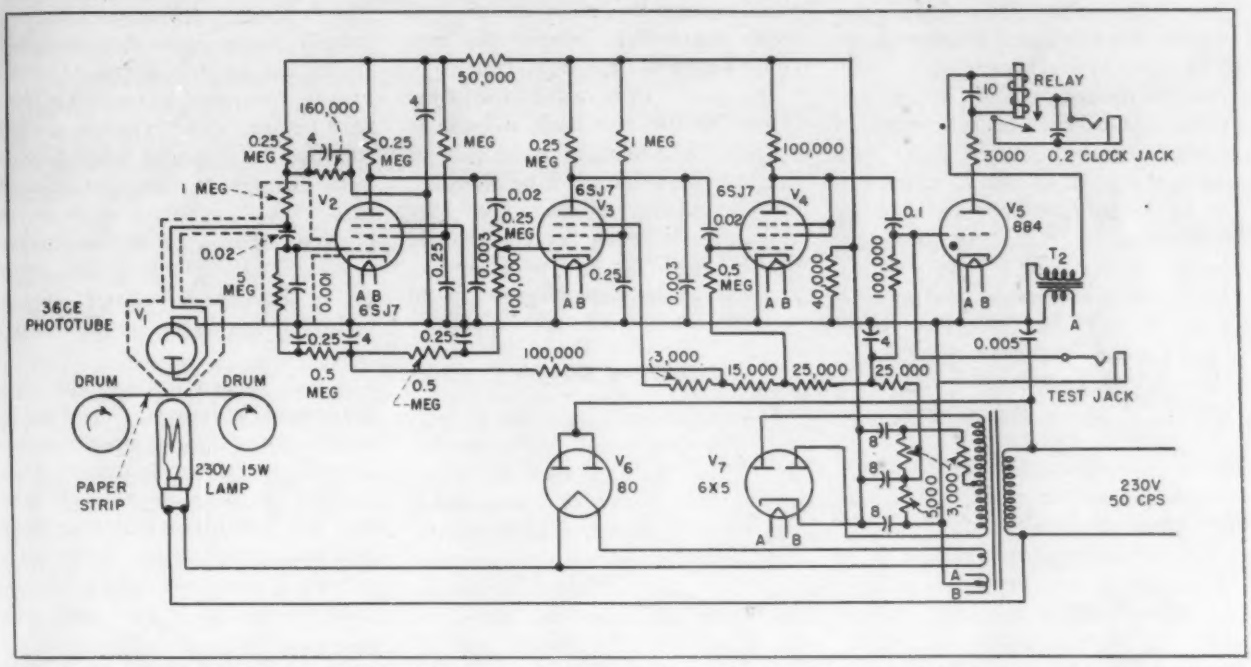


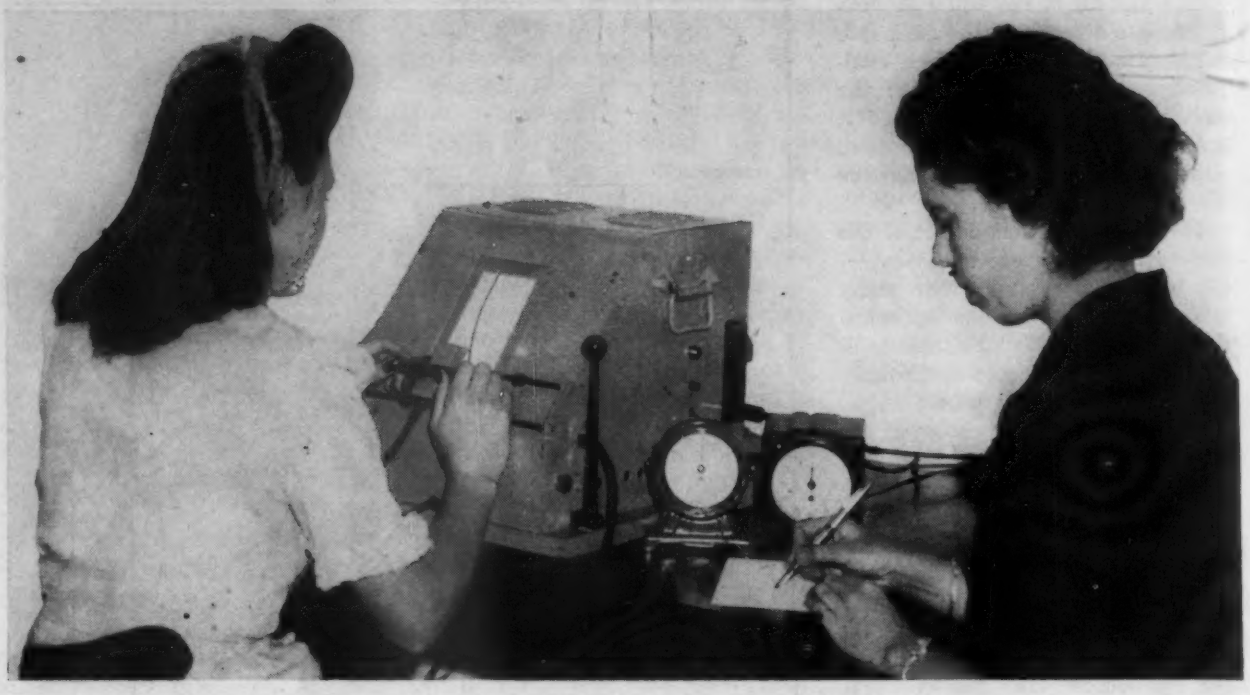
FIG. 2—Circuit of photoelectric amplifier used to start electric stop clock when phototube goes off moving line, and stop clock when phototube is on the line

The wavy line, approximately 15 feet long, is drawn in India ink on a strip of paper which can be rolled from one roller to another. The pointer is in the form of a hollow cone with a hole at its apex and encloses a phototube. A lamp placed behind the paper strip shines through this hole except when the hole is just over the India ink line. The response of the phototube to the light operates a relay through an amplifying system, switching on the electric stop clock to record the amount of time the pointer was off the line.

Mechanical Details

Details of the mounting and drive for the two paper drums are shown in Fig. 1. The main frame consists of two end plates with a flat top and a 60-degree front panel all mounted on a wood baseboard. Between the end plates two half-inch steel shafts are mounted, each carrying a brass drum $3\frac{1}{2}$ inches in diameter and $3\frac{1}{2}$ inches long with $\frac{1}{4}$-inch flanges. Drum A is keyed to its shaft, which is driven by a motor, while drum B rotates freely on the upper shaft.

Each shaft carries a sprocket on the left-hand side of the drum, the sprockets being joined by a chain drive. Whereas sprocket AA is firmly attached to the left-hand end of drum A, sprocket BB is freely mounted on a bushing projecting from drum B. A spiral spring is anchored to an extension of the bushing beyond the sprocket, and after taking several turns round the



Operator at left moves phototube along horizontal rail with hands to keep it over moving line, while controlling motor speed with foot pedal so as to run paper from one drum onto the other as fast as possible. Total time taken for a test is clocked by observer

Aptitude Tester for Garment Workers

Operator attempts to keep phototube aimed at wavy black line while controlling speed of paper strip with foot pedal. Total time that phototube is off the line is correlated with time taken for test to find aptitude for operating sewing machine

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National Physical Laboratory, Council for Scientific and Industrial Research, Union of South Africa

IN the science of aptitude testing, many devices have been evolved to measure ability of a subject to perform one or more operations involving a straight eye and steady hand. Every new research project involves the planning of one or more of these devices and adds to the stock of current psychological apparatus. The South African National Bureau for Personnel Research and the South African National Physical Laboratory recently co-operated in planning a set of aptitude tests, for one of which it was necessary to test the ability of a garment worker in handling a sewing machine. It was decided to use electronic methods, and the machine which was evolved is described.

General Description

In this hand-foot co-ordination machine, a wavy line is moved towards the subject at a speed which can be controlled by the subject with a foot pedal. The subject endeavors to keep a pointer in contact with this line, and a clock automatically records all those intervals of time when the pointer fails to make contact with the line. When the line is comparatively straight the subject can speed up its movement but when there are sharp changes of direction the speed needs to be reduced. The time of the whole experiment, taken in conjunction with the total time during which the pointer was off the line, indicates the aptitude of the subject for machine sewing.

bushing is fastened to a pin projecting from the sprocket. This spring keeps the paper strip tightly and uniformly stretched between the two drums and compensates for slight variations in diameter as the paper is wound off one drum and onto the other.

As the two drums rotate together through the chain drive, more paper comes off the full upper drum than goes onto the empty lower drum. The spiral spring is able to unwind, thus taking up the slack in the paper and keeping it stretched. When more than half the paper has been transferred to the lower drum the reverse happens and the spiral spring is wound up again.

The motor shaft is at right angles to the driven shaft and carries a large smooth-faced friction plate. The driven shaft carries a small rubber-rimmed wheel which can slide on the shaft on a keyway. A spring forces the friction plate against the small wheel, which can be moved up and down its shaft by a Bowden cable leading to a foot pedal on the floor. Thus continuously variable speeds are possible from zero, when the small wheel contacts the center of the friction plate, to a maximum when it is pulled out to the edge of the plate. The motor is a synchronous type so as to give accurately reproducible speeds for given positions of the foot pedal.

The phototube is mounted in a small metal box provided with a conical extension in front, which projects down until it almost touches the paper, and has a fine hole bored in it through which the light from the lamp behind the paper can reach the tube. The box slides on a pair of horizontal guide rods, and is provided with a suitable handle so that the subject under test may follow the movements of the ink line from side to side.

The output of the phototube passes to an amplifier located at the back of the apparatus. An electric stop clock cable is arranged to plug into this amplifier with a telephone jack. The sloping front panel has a window 5 inches long and 3 inches wide through which the paper can be seen. Behind the paper at this point is a metal gate with a well-fitted glass window to keep the paper in a uniform plane close to the nose of the phototube box.

Limit switches prevent the subject from keeping the drums turning at the end of a test and winding the paper completely off the upper drum. The lowest shaft is threaded and carries a nut with a pin projecting from it and able to slide in a horizontal slit milled in a brass plate which is held parallel to the shaft. As the shaft rotates in the nut, the pin moves along the shaft, eventually striking one of the pair of limit switches. This switch reverses the motor and winds the paper back onto the upper drum. When the starting point on the paper is reached the nut and pin strike the second limit switch, which opens the main circuit and stops the motor, leaving the machine ready for the next test.

In order to shorten the time taken for the run back, a lever is provided at the right hand side of the machine. This can be operated by the tester, pulling the gear change cable much farther than is possible with the foot pedal and so increasing the paper speed by 50 percent.

Electrical Circuits

It was decided that a line $\frac{1}{32}$ inch to $\frac{1}{8}$ inch wide would be representative of the markings used on tailoring material in practice, and that a movement of about $\frac{1}{32}$ inch off the line should be sufficient to set the stop clock going. Therefore, a hole not larger than $\frac{1}{16}$ inch diameter was the largest that could be used to admit light to the phototube. This called for several stages of amplification before the signal obtained from the phototube could actuate a relay. Furthermore, as a d-c amplifier is necessarily an elaborate piece of equipment, the output from the phototube should preferably be a-c which can be amplified by a conventional amplifier.

For illuminating the paper strip an ordinary 230-volt, 15-watt lamp was placed at the back of the paper and supplied from the 50-cycle mains via a half-wave rectifier. In this way the lamp received current on alternate half cycles only and the cooling time constant of the filament was sufficiently short to allow an appreciable fluctuation of the light output. The alternating component of the light flux was about 50 percent of the total flux. In this way the output from the phototube contained an appreciable amount of 50-cycle a-c.

The amplifier consists of two stages of pentode amplification in cascade, followed by one stage of triode amplification. This latter stage also serves the purpose of isolating the high-gain stages from the following stage which is a thyratron (gas-filled relay) having a mechanical relay in its anode circuit, this final relay operating the clock.

Resistance-capacitance coupling is used throughout the amplifier. The frequency response is made to have a rising characteristic at 50 cycles and to cut off the unwanted higher frequencies resulting from tube noises. The thyratron plate is supplied with a-c in such a phase with respect to the signal on its grid that it fires for part of the positive cycle. Thus the current flowing in the plate circuit is pulsating d-c, which is used after filtering to actuate the mechanical relay.

A gain control is incorporated in the second pentode stage so that the sensitivity of the apparatus can be controlled to a certain extent. Fixed bias is used on all the tubes to forestall any difficulties which arise occasionally from phase shifts when cathode biasing is used. For the same reason the bypass and blocking capacitors are large enough to cause negligible phase shift at 50 cycles.

For ease of servicing, type 6SJ7 tubes were used for all three amplifier stages. The somewhat unconventional way in which the a-c voltage for the thyratron plate is obtained was necessitated by the fact that no power transformer could be obtained having a second insulated high-voltage winding.

Validation tests on this machine are at present being made by the Bureau for Personnel Research. Apart from small alterations and adjustments it is proving satisfactory in operation, and a first estimate of its ability to indicate aptitude for certain types of garment work is promising.

Reversible-Motor Controller

Circuit sets limits of rotation of a continuously reversing motor for such applications as repeatedly tuning through a radio band or controlling reciprocators in factories

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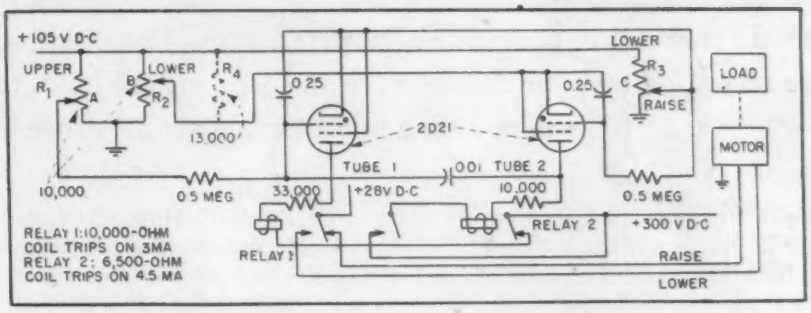
UPPER AND LOWER LIMITS of a sector-scanning radio receiver are set by the circuit shown in the accompanying drawing. It can also be used with any device that oscillates between occasionally changed limits. Devices to which it might be adapted are radio transmitters, radar equipment, automatic factory techniques using milling machines, lathes, or variable-depth boring tools; in short, any device that must move repeatedly between two points within its complete range.

Circuit Operation

The circuit is relatively simple. Either end of the automatic scan can be set without regard to the other by means of linear potentiometers R_1 and R_2 at a remote point. Potentiometer R_3 is geared to the turning mechanism of the remotely controlled load so that its full resistance is traversed with the complete range.

Relay 1 controls direction of travel of the motor. The load will be raised when Relay 1 is de-energized, lowered when it is energized. With the potentiometers in the positions shown, Relay 1 will be initially de-energized and the motor will raise the load until the voltage at C of R_3 very nearly equals the voltage at A of R_1. Relay 1 will then be actuated by the firing of Tube 1 and the motor will reverse, lowering the load. Tube 1 will remain conducting until the voltage at C of R_3 nearly equals the voltage at B of R_2. At this condition Tube 2 will fire and Relay 2 will be actuated momentarily, thus de-energizing Relay 1. This action will deionize Tube 1 and reset the circuit to the original condition.

Upper and lower limits must not be crossed. Some mechanical means should be provided to prevent this possibility. On the other hand, if by some means, the load is moved past the lower limit, the load will be raised into the desired sector when the circuit is energized. If the load is moved past the upper limit, it will be lowered into the sector when the circuit takes control.



Circuit automatically reverses motor when load reaches preset limit

Design Considerations

Potentiometers R_1, R_2 and R_3 should preferably be supplied from the same d-c voltage, but, if necessary, separate regulated supplies having the same output voltage can be used, thus saving one control wire.

Potentiometers R_1 and R_2 can be arranged to indicate sector limits on the same linear scale, but some error in the lower limit indication will result from a small current passing through R_3 when Tube 1 and Relay 1 are energized. This error can be minimized by (A) a low resistance potentiometer at R_3, (B) a sensitive coil for Relay 1, (C) a higher d-c voltage on all potentiometers, or (D) a compensating resistance (R_4). If separately calibrated indicator scales are used for R_1 and R_2, this error is eliminated.

Values shown on the diagram were chosen for remotely tuning a radio receiver. Both upper and lower limit potentiometers indicate on the same linear scale and are arranged mechanically so that electrical crossover is impossible. Using inexpensive three-watt, wire-wound potentiometers and R_4, maximum error in indication of sector limits is about 3 percent of the total tuning range. The circuit is stable down to a minimum sector of about 0.25 percent at a sweep period of 30 seconds over the entire range. Tripping is consistent within about $\pm$ 0.05 percent for a given set of tubes.

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and
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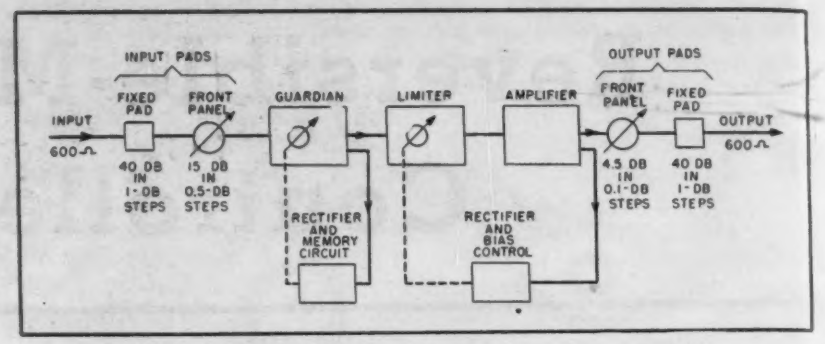


FIG. 1—Functions of the various stages in the amplifier

Automatic Gain Control and Limiting Amplifier

An agc amplifier regulates the audio signal applied to a peak limiter so that the output is held constant to provide a high percentage of modulation in a transmitter or high level for recording or public address. A memory circuit holds the gain constant for a predetermined time to preserve dynamic range of program

The automatic gain control and limiting amplifier (Progar) to be described was designed to bridge the gap between the physical limitations of broadcasting, recording, and public address equipment and the ability of operating personnel to maintain an efficient program level within these limits.

In broadcasting, dynamic limits are sometimes exceeded because of an operator's desire to maintain a high percentage of modulation at all times. Early attempts to alleviate this situation made use of peak limiting or compressor amplifiers at the transmitter. Unfortunately, this led to relaxing of critical monitoring. Operators relied upon the peak limiter to maintain a high average percentage of modulation, disregarding the fact that this statement holds true only when the proper signal level is fed into the peak limiter.

Guardian Amplifier

The Langevin 119-A unit was designed to be inserted between studio audio facilities and the transmitter. (Several have also been built by Royal S. Howard of the Associated Broadcasting Company in San Francisco.) Figure 1 shows the sections into which it is divided, consisting essentially of an automatic gain control or guardian amplifier that provides a regulated audio flow and an improved type of peak limiter amplifier.

The input signal is first fed into the guardian amplifier and from there branches into two circuits, part going to the peak limiter and part going into the rectifying and memory circuit. The output of the rectifying and memory circuit controls the gain of the guardian amplifier. The output of the limiter also branches into two circuits, part going to the output load and part to the rectifier and bias control circuit which provides fast gain reduction action of the peak limiter amplifier.

Figure 2 shows the combined action of the guardian and limiter circuit. The lower graph represents the guardian action. The vertical axis represents the peak value of the input signal, and the horizontal axis represents the peak value of the signal at the output of the guardian circuit, which is also the input to the peak limiter. The vertical axis of the upper graph represents the peak value of the output signal.

Gain Control Action

The input controls are adjusted so that the average peak input signal is −31 db. The average peak into the limiter occurs at point *A*, which is at the knee of the curve, and the average peak output is +22 db, which is made equivalent to 80 percent modulation. As the body of the program material falls below the average peak, the full dynamic range of the material is reproduced due to the linear action of both amplifiers. Any transient peaks above the average peaks are compressed in the limiter section.

If the average peak value of the input should increase and remain there for any reason, as from −31 db to −21 db, the input to the limiter will increase to *B* and the average peak output will be +23 db, or 90 percent modulation. Under this condition, part of the body of the

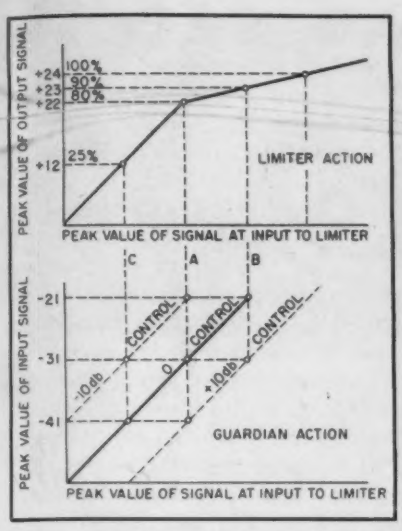


FIG. 2—Circuit action of the combined amplifiers at various levels

Authors Guenther and Jurek check a specification sheet for the Progar amplifier

program below the average peaks will be compressed in the limiter. The gain of the guardian amplifier is then reduced to the −10 db control line so that the increased input produces an output from the guardian back at *A*, and the output from the limiter is again back at +22. The time for the guardian section to decrease its gain is much slower than the limiter attack time but is still too fast to be detected by the ear.

If, while the limiter is in this condition, the operator changes the input level back to normal (to −31 db), the input to the limiter will drop to *C* and the output signal to +12 db. The guardian section does not act immediately, but the memory circuit holds the gain constant for a predetermined time and then allows the gain to increase gradually to normal as represented by the zero control line. This wait time is relatively long and in present models is adjustable from 2 to 8 seconds. This long delay is provided to allow for station breaks and fade-outs as used in dramatic shows. If the action were extremely fast, the amplifier would try to maintain constant output continually and destroy the dynamic range of the program.

Should the average peak input signal level drop below its normal value, as from −31 to −41 db, with the corresponding output of +12 db, the guardian amplifier gain will again be held constant by the memory circuit for the predetermined time and will then slowly increase to the +10 db control line, bringing the output of the guardian amplifier back to normal. Upon restoring the input signal to normal again, the input to the limiter will be at *B*, and the limiter will instantly reduce this increased input signal to a safe output value until the guardian gain changes again.

The combination of the two amplifiers maintains a steady audio flow into the transmitter at all times. As is seen from the lower graph in Fig. 2, the gain-changing action of the guardian amplifier is not strictly of the compressor type, due to the action of the memory circuit. Therefore, even though the average dynamic range is changed, as it will be in any gain-changing amplifier, the dynamic expression is not affected.

Circuit

Figure 3 shows a schematic of the complete unit. The audio section of the amplifier consists of four push-pull cascade stages with transformer input and output. The first two tubes are the guardian control section; the second stage is the limiter amplifier control section; the third stage is the linear amplifier section; the fourth stage is the power output section.

The control circuit for the guardian section consists of a two-stage resistance-coupled pushpull linear amplifier operating into a full-wave rectifier. The output of the rectifier feeds an RC circuit which builds up the d-c potential in proportion to the incoming program level. A portion of this potential is fed through the conducting diode, V_{12B}, into capacitor C_{20}. The voltage across C_{20} is then fed into the guardian amplifier control tubes. With the correct input signal applied to the tubes (−31 db), the bias of the tubes is established by the control circuit, mentioned above, to set the gain of the tubes in the center of the operating range. The purpose of this condition is to allow the tubes to increase or decrease amplification by approximately 15 db, thereby effecting the agc action.

If the incoming signal should be

increased, additional potential is developed in the first RC circuit of the control section. Portions of this voltage are almost instantly transferred through conducting diode V_{12B} into C_{22}, increasing the control voltage on the tubes and thereby reducing the gain of the first stage and reducing the signal to the proper level going into the second stage. This circuit may be referred to as a reverse-acting control circuit whereby the operation of the circuit depends upon permissible error, which in this case is 1-db change in the output for a 10-db change in the input. Therefore, the action taken by the control circuit must be such that it produces enough control signal to reduce the incoming signal 9 db with only a 1-db increase in the output.

If the average peak of the input signal should drop from its normal value, the charging voltage will be reduced across C_{20}. If the signal is slightly reduced, no effective change will be produced on C_{22} because this capacitor may only be charged by V_{12B}, or discharged by V_{12A} and then only if the voltage of C_{22} equals the voltage of C_{20}.

This circuit is the heart of the guardian circuit, and is referred to as the memory circuit. This explains the reason why during a symphony concert the pianissimo passages are not affected, as some portions of it keep the first RC (C_{20}) charged sufficiently so as to prevent C_{22} from being discharged, thereby keeping the gain constant in the guardian circuit. Should the program drop completely, capacitor C_{20} would start discharging because no signal is being fed into the control circuit. As soon as the voltage across C_{20} equals the voltage across C_{22}, the conducting diode (V_{12A}) would start discharging C_{22} at the same rate along with C_{20}. This would reduce the control voltage, allowing the gain to increase in the guardian stage. This gain will increase to the limit determined by the setting of the guardian increase limit control. The purpose of the guardian increase limit control is to limit the maximum gain of the control amplifier when no signal is present. In average studio conditions, a 5-db limit is recommended. Ten and 15-db limits are provided for unattended service if the noise level of the preceding equipment will allow this additional gain.

A Guardian Action Disabling switch is provided for testing this circuit for normal gain conditions and for tube balance checks. This switch substitutes a fixed bias in place of the control circuit bias.

Guardian action is observed by measuring the combined plate currents of the guardian control tubes. This measurement is only relative as it measures the plate current of the tubes and is calibrated in terms of audio signal. A calibration control is provided to set the meter at zero position when the guardian action switch is in the disabled position.

The peak limiter section is composed of the last three pushpull stages, with a biased diode rectifier connected across the output. This rectifier rectifies a portion of the output voltage whenever that voltage exceeds the predetermined

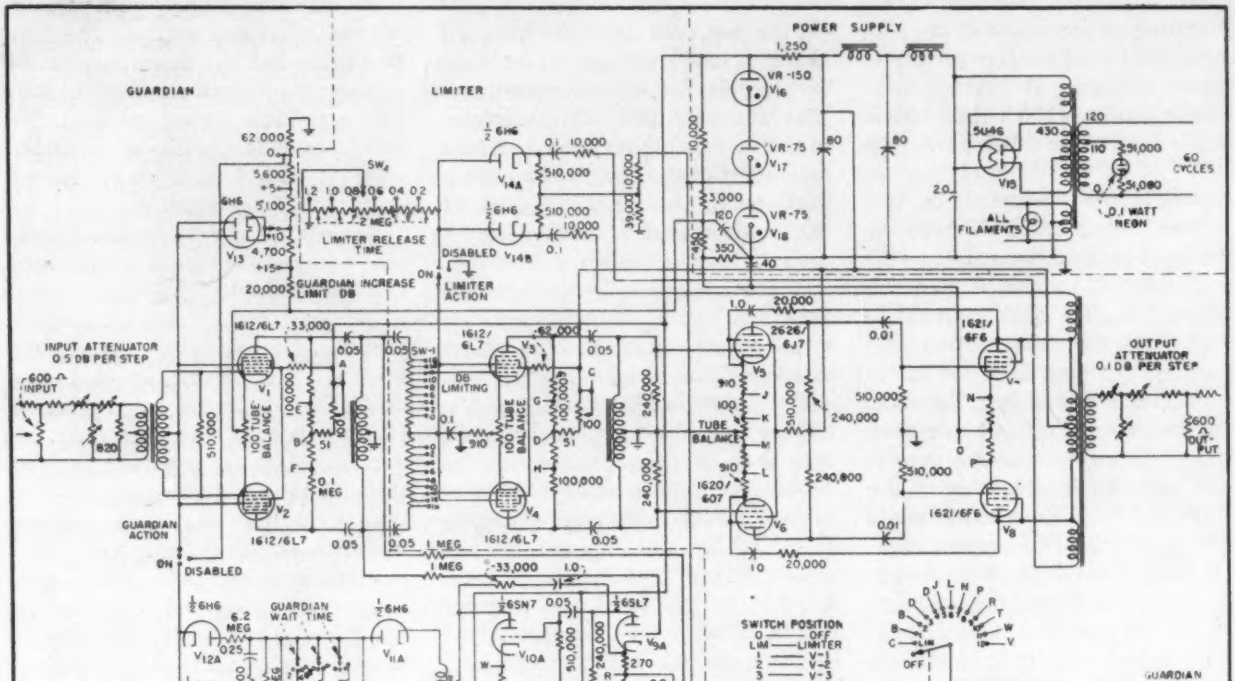
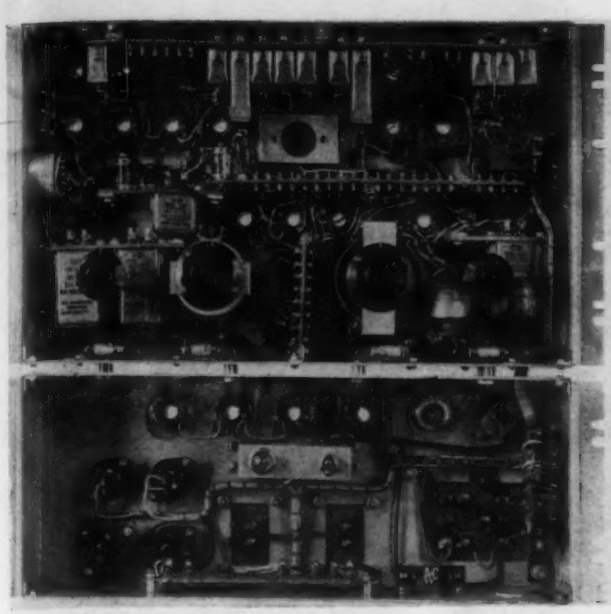
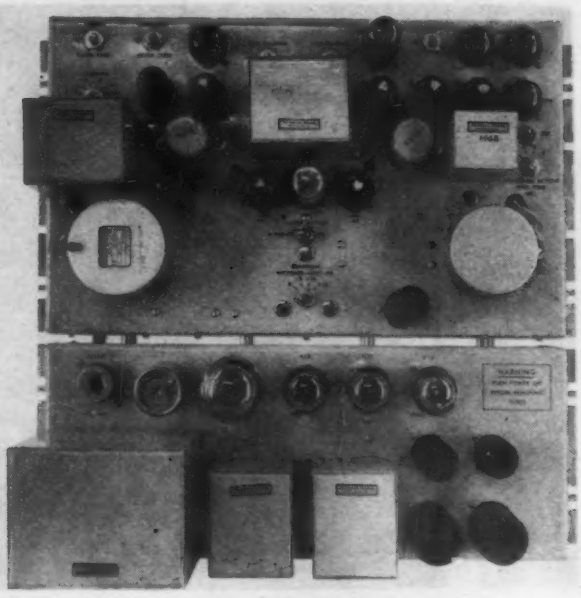


FIG. 3—The complete circuit is basically a four-stage audio amplifier, with added control circuits for the special functions



Front view of amplifier chassis with meter panel removed



Screwdriver-adjusted controls are accessible from the rear

amount and applies that rectified voltage as a bias to the limiter input stage.

The limiting action is also a reverse-acting control circuit, and is designed to operate at a 10:1 control ratio, whereas a 10-db increase in input signal produces only a 1-db increase in output signal above the predetermined peak value.

The output of the guardian section is so adjusted that when the limiting control is in the zero position no limiting action takes place on normal program material. Due to the slow attack time of the guardian circuit normal transients appearing in the signal have no effect on the guardian control circuit and are passed on to the limiting section where they are reduced by the control action of the limiter. This also explains why steep wave front tones such as certain piano tones and cymbal crashes have little or no effect on the guardian action, and if dangerous to the transmitter are reduced to a safe value by the limiting action.

Circuit Tracking

Reverse-action control circuits were selected for both of the control circuits in place of forward-acting control circuits because of the extended range of control without cutting off the amplifier. Forward-acting circuits give difficulty in maintaining the control action over a range greater than 13 db. After that point, they have a tendency to cut off the controlled amplifier completely, thereby producing noticeable holes in the program material if the peaks should exceed the normal operating range encountered in practice.

The forward-acting control circuits receive their control voltage from the input of the amplifier and must develop the control voltage that will track along with the control action of the variable-gain tubes. The control circuit rectifiers produce a voltage that is linear to the input voltage, while the controlled tubes require a logarithmic voltage. Therefore, the tracking of a forward-acting circuit is difficult to control over an extended range.

In the reverse-acting control circuit, tracking is no problem as the circuit is selfbalancing, working on permissible error, and control action is automatically extended beyond the required 30 db.

Semifixed and variable attenuators are provided at the input and output to adjust the input signal and provide the unit with the proper input signal of −31 dbm (db referred to a zero level of 1 milliwatt), and to adjust the output of + 22 dbm to correspond to 80 percent modulation at the transmitter. These input and output levels must be established before putting the unit into operation.

Semifixed pads with a 40-db range are provided at the input and output to allow the unit to be substituted in place of the conventional line or program amplifier or inserted as a zero-gain device at the transmitter audio input.

Common Power Source

The power supply is conventional with VR regulator tubes stabilizing the critical voltages such as the guardian circuit and the limiter diode bias. It is recommended that the power supply be operated from the same a-c power source as the modulating stage of the transmitter. This is to allow for the variations in the line voltage that will change the modulation capabilities of the transmitter and will also change the output capabilities of the Progar, and will therefore maintain a uniform maximum operating point of both units simultaneously. If a well-regulated power supply is used, a constant fixed output will be maintained, and as the line variations affect the modulation capabilities of the transmitter the full benefits of the control action could not be utilized.

For network operation, it is advisable to use one unit at the studio with the limiter action disabled, and a second unit at the transmitter with the guardian action disabled. The reason for this recommendation is that if the two limiters are used in cascade with different release times, a pumping action will be noticeable on some program material.

Mass-Production Speed

Ultraprecision electronic balancer brings gyro rotor up to test speed of 12,000 rpm in a few seconds and indicates amount and location of unbalance directly, permitting skilled operator to balance a rotor in two minutes by alternately measuring and drilling

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forces on the resilient rotor supports, resulting in voltages that identify both magnitude and angular position of required unbalance correction.

The principle by which the balancing instrument could focus attention on one plane of correction without interference from the other may be demonstrated by reference to Fig. 2. Consider that a localized unbalance mass exists in each of the correction planes and that these masses have any angular configuration with each other. Let the unbalance forces due to each of these masses be represented by W and X respectively. Force X will manifest itself at shaft ends A and B as components X_a and X_b respectively, all having the same angular orientation and of magnitudes such that

$$X_a + X_b = X$$
$$aX_a = bX_b \qquad (1)$$

Similar relations hold for components W_a and W_b. These four components of X and W will produce resultants R_a and R_b at ends A and B respectively.

Electrical transducers at ends A and B can produce voltages which are proportional both in magnitude and in phase to the resultant forces R_a and R_b. Let us, for the sake of simplicity, allow the notations R, W, and X to indicate also the electrical voltage vectors corresponding to these forces. Then the voltages generated at A and B will also be represented by the vectors of Fig. 2. To the voltage R_a let us add a part of the voltage of R_b, reduced in magnitude by the factor b/a and reversed in instantaneous polarity as indicated in the lower right in Fig. 2. Resolving these into their original vector components, we see that $-(b/a)X_b$ is exactly equal and opposite to X_a since from Eq. 1

$$X_a = (b/a)\,X_b \qquad (2)$$

Since the rotor is symmetrical with respect to dimensions a and b, $aW_b = bW_a$ and $W_b = (b/a)\,W_a$. Making the vector summation of voltages R_a and $-(b/a)R_b$, the X component disappears and the remainder becomes

$$W_a - \frac{b}{a}W_b = W_a\left(1 - \frac{b^2}{a^2}\right) \qquad (3)$$

which is a voltage proportional both in magnitude and phase to the original force W only. By mixing the generated voltages in a proportion which is a function of the locations of the balance correction surfaces and the transducers alone, it is possible to determine both the location and magnitude of the unbalance on either correction surface.

FIG. 1—Gyro rotor. Air stream impinging on vanes in center drives rotor at 12,000 rpm during balancing tests

Evolution of Final Design

For balancing, the rotor was initially mounted in the double suspension system of Fig. 3, using rubber as the resilient material and operated above resonance with moving-coil pickups as transducers. The pickup units are similar to the Western Electric D-93306 reproducer for vertical transcriptions. Each pickup was coupled directly to the bearing suspension by means of a small piano-wire rod which drove the pickup sinusoidally while permitting rotary motion of the bearing. This rubber suspension method proved inadequate, however, because it permitted motion axially as well as radially, did not easily allow separate control of compliance and damping, and did not maintain the rotational axis location within sufficient accuracy between test and use.

The final suspension design overcame these limitations, while still maintaining the advantages of free plane motion, by employing ball bearings mounted in plates supported on three horizontal round cantilever rods. This permitted free motion in the planes of the bearings, but not axially. Damping was provided separately by enclosing each cantilever rod within a fluid-filled cylinder, the working end of which contained a thin neoprene dam for fluid retaining. Damping was controlled by means of beads mounted on each suspension rod and also by adjusting the properties of the fluid. Rosin dis-

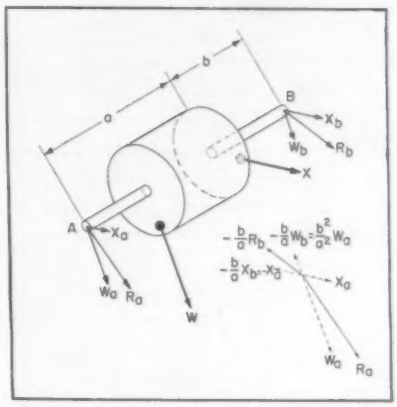
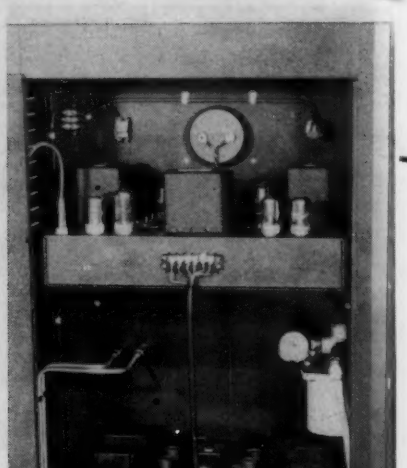


FIG. 2—Force reactions on rotor due to unbalance correction, and vector summation (lower right) for unbalance indication

Precision Balancing at

With this ultraprecision electronic balancing machine, a skilled operator can easily balance 200 gyro rotors a day to an accuracy of a few micro-ounce-inches. A drill press is conveniently located alongside the machine, and the operator alternately measures and drills until the required balance is achieved



Rear view of gyro balancer, showing how amplifiers are arranged on two separate chassis units that can readily be removed for servicing

THE heart of an automatic pilot for aircraft is the gyroscope or gyro rotor, which rotates at 10,000 to 20,000 rpm and must be dynamically balanced to a degree of accuracy hitherto unattainable. Production demands for these rotors in the Jack & Heintz plants rose to an unprecedented total of 35,000 units per month during the war, each balanced to the astonishing accuracy of a few millionths of an ounce-inch. This is the story, told now for the first time, of the ultraprecision electronic balancing instrument which did this work.

The Balancing Problem

The gyro rotor, shown in Fig. 1, is a round rigid body which is to rotate about a rotational axis determined by the positions of the pivots or conical bearing surfaces which form the inner races or cones of the ball bearings. Another axis through the rotor, near the rotational axis but so directed as to pass through the average of the positions of the centers of gravity of successive thin sections taken along the axis, is called the axis of inertia. This is the axis which will produce no pressures on the bearings due to centrifugal forces when rotation takes place about it. Dynamic balancing involves shifting the axis of inertia until it coincides with the rotational axis within the balance accuracy required. (Strictly speaking, the axis of inertia is defined as the z axis about which the summation of the products of inertia is zero.)

The condition of unbalance can be located, measured, and expressed as a weight of w ounces located r inches from the rotational axis. The dimension ounce-inch was found to be inconveniently large for gyro work, hence the micro-ounce-inch (one millionth of an ounce-inch) was used. The required sensitivity for the balancing instrument was between 2 and 5 micro-ounce-inches; actually it was possible in the final design to attain a maximum sensitivity of one micro-ounce-inch.

Principle of Balancing

It can be shown analytically that regardless of the number or distribution of unbalance masses within a rigid body, balance correction can be effected in any two arbitrarily chosen planes at right angles to the rotational axis. Conversely, it can be considered that all the unbalance occurs in the two previously chosen planes of correction.

During rotation, centrifugal action of the unbalance masses at the two correction planes will produce

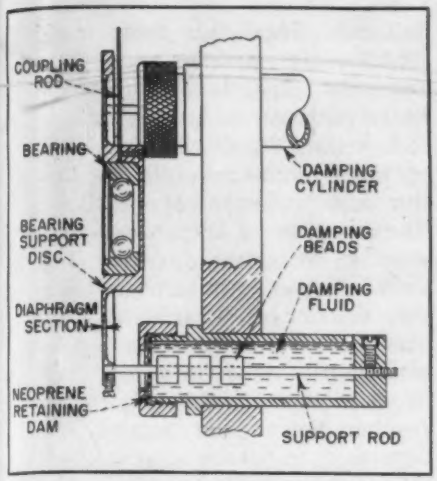


FIG. 5—Design of suspension assembly and damping cylinder

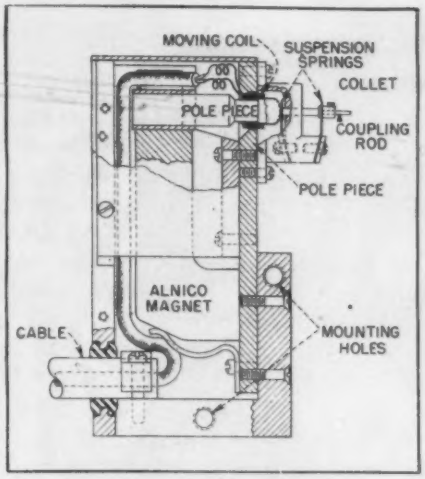


FIG. 6—Construction of pickup unit, showing similarity to p-m dynamic speaker

rms for 1 micro-ounce-inch unbalance.

The foot valve of the pneumatic system is operated only to attain rotor speed quickly, after which opening of the on-off air valve and proper preadjustment of the needle valve maintains rotor speed.

Final Electronic Circuit

In order that the signal mixing be accomplished at reasonable level, the pickup voltages are amplified separately in identical channels. Each channel employs an input transformer, an attenuator, a two-stage electronic amplifier, a special filter, and a gain adjustment, as shown in Fig. 7. The transformers are selected and matched for phase response, and each provides a step-up ratio of 1:150 or a gain of 43.5 db. The attenuators are pushbutton operated and ganged and each provides six steps of 10 db in a total resistance of 250,000 ohms. Each step corresponds to one scale range, the first or most sensitive being 0 to 30 micro-ounce-inches and the last being 10,000 micro-ounce-inches full scale. The attenuator pushbuttons are conveniently located at the front of the instrument table, each inscribed with a letter designating its range.

The amplifiers are identical feedback-stabilized two-stage sections of conventional design. Each provides about 40-db gain with negligible phase shift at 200 cycles. The filters are 100 to 200 cycle band-pass units of special design providing an attenuation of at least 25 db in the stop-band below 60 and above 400 cycles and a transition loss of less than one decibel in the working range of 180 to 220 cycles.

The design item of primary importance for the filter is the rate of change of phase angle with frequency in the working range. This is held within 0.5 degree per cycle and is constant for more than a ±10 percent range of frequency at 200 cycles when used with the circuit terminations shown. The variable resistor in the output termination provides a ±2-db gain adjustment so that both channel signals may be matched overall.

Mixing Circuit

The mixing circuit provides a means of taking the output of one amplifier and adding to it a portion of the output of the other amplifier in reversed polarity as called for by Eq. 3. It also allows the operator to obtain an unbalance indication on either the left or right side of the rotor without interference from the unbalance of the other side. The circuit is switched by two pushbuttons located at the left of the instrument table.

Inverse feedback added to an amplifier produces many advantages but one disadvantage here is that with it the amplifier is no longer a passive network, and a voltage applied at its output end may be amplified to cause disturbing interference. The resistors in each input leg of the mixing circuit are isolating networks to avoid this difficulty. This isolation is effected at the expense of gain, part of which is recovered in the mixing transformer while still allowing a net loss through the mixing circuit of about 28 db.

The focus adjustment potentiometer permits adjustment of the ratio b/a (Eq. 2 and 3) and when calibrated provides a loss of about 15 db for one channel with respect to the other. The phase-adjusting capacitor in the mixing circuit provides for a small phase adjustment of the vector $(b/a)R_s$ (Fig. 2).

Meter Amplifier

The amplifier and filter following the mixing circuit are identical with the input counterparts already described. The amplifier recovers the voltage lost in the mixing circuit. The additional filtering provides appreciable isolation of the fundamental unbalance signal with respect to extraneous bearing vibration components, permitting the operator to obtain an unbalance indication for an unbalance component appreciably weaker than the bearing vibration disturbance. The filter output termination here again provides ±2 db gain adjustment for overall calibration of the signal which at this point has a value of about 1.0 volt rms at full-scale meter deflection.

The meter amplifier provides an additional gain of 17 db to its output, together with signal rectification for the meter movement. One diode of the rectifier provides half-wave rectification for the meter, while the other bucks out the normal zero-voltage diode current.

The meter has a 4,000-ohm 100-microampere d-c movement and is arranged with two scales laid out to be 10 db apart. Above these scales are 12 arbitrary balance numbers laid out 1 db apart for the operator's convenience in remembering or denoting the magnitude of unbalance. These are used with any of the six scale ranges by prefixing the letter designating the scale range. For instance, an unbalance of 200 micro-ounce-inches can be more readily designated by the operator as a reading of C9. These numbers are spaced well within the accuracy needed by the operator in effecting a balance and greatly en-

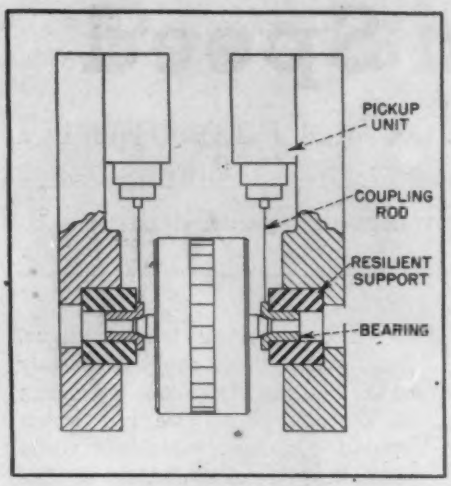


FIG. 3—Double suspension system used initially, having rubber bearing supports

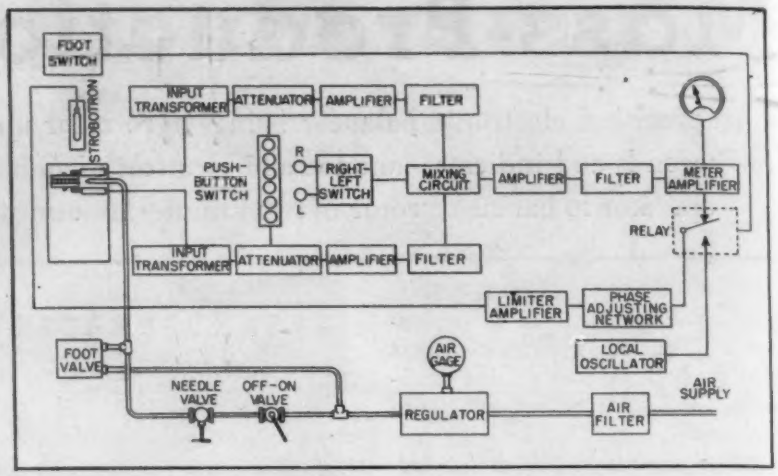


FIG. 4—Block diagram of ultimate balancer design. Air supply system drives rotor at required 12,000-rpm test speed

solved in castor oil worked remarkably well in the early tests.

Phase response for both the mechanical and electrical parts of the instrument proved highly important. Low phase-shift band-pass filters were therefore used in the amplifier channels instead of resonant circuits. Low attenuation was required in the region of 200 cycles, high attenuation in the region of the mechanical resonance of the suspension, and high attenuation for all harmonics of the rotational frequency (400 cycles and above).

Since the phase shift in the filters is only one of several factors adding to error in location, 2 degrees total shift was arbitrarily set as permissible for the filters in the working speed range. Use of two filters in series to obtain the desired attenuation in the stop band cut this allowable phase shift in the working speed range down to 1 degree per filter.

A Strobotac was used initially for speed-measuring purposes, and gave accuracies easily better than 1 percent at 200 cycles, corresponding to a rotor test speed of 12,000 rpm. The use of a stroboscope for speed determination immediately suggested its use also for unbalance location by using the rotor as its own reference. If the corrected unbalance voltage were used to initiate the flash of the stroboscopic lamp, the spot requiring unbalance correction could (with proper phase adjustment) be illuminated. This spot could be easily identified stroboscopically while rotating if the rotor were initially stamped with a series of numbers around its periphery.

The block diagram in Fig. 4 shows the final instrument design, including the mechanical suspensions, the connections for compressed air that supplies motive means for gyros under test, and the electronic circuits.

Final Mechanical System

Details of the final mechanical suspension appear in Fig. 5. The diaphragm-type construction of the bearing support disc is for the purpose of handling secondary axial and angular components of motion of the rotor bearings. These discs are machined of dural (17st) in one piece, with the diaphragm section reduced to a thickness of 0.0050 to 0.0055 inch. The pickup coupling rod is a 0.020-inch diameter wire. The suspension support rods are 0.040-inch diameter steel drill rod, the ends of which are threaded to permit alignment adjustment of the bearing support disc. The damping fluid was changed to one of the Dow Corning Silicones, which proved superior to the rosin-castor oil mixture.

The natural period of the entire suspension system with gyro rotor in place is very nearly 29 cycles, which is about a factor of seven below test speed. The damping produces a mechanical phase angle which changes less than 0.3 degree due to changes of damping factor under normal operation of the instrument and about 0.2 degree due to changes of rotor speed within the normal operating speed range. The total mechanical phase shift error is therefore not over 0.5 degree. The mechanical displacement sensitivity is 0.2 micro-inch peak per micro-ounce-inch of unbalance and is unaffected by the normal speed range or by damping.

One suspension assembly is mounted on a ball-bearing slide to allow about an inch of motion for insertion and removal of test gyros. A calibrated spring acts against this slide to provide end thrust or bearing loading equal to that which obtains in actual assembly, thus reducing errors due to shift of rotational axis. A lock is provided for clamping the slide during test in such a way as not to disturb the accuracy of alignment. The ball-bearing slide reduces variations of end loading that might be caused by friction and at the same time provides for long-time accuracy of adjustment and ease of operation which are so important in production use. A vibration isolation mounting supports the assembly.

A cross-sectional view of one of the pickup units appears in Fig. 6. The impedance is about 6 ohms at 200 cycles, and sensitivity is about 45 db below 1 volt per inch per second velocity. In terms of balancing instrument requirements, each pickup will develop about 1 microvolt rms open circuit for a peak sinusoidal displacement of 0.2 microinch at 200 cycles per second. Overall sensitivity at the 200-cycle operating speed is thus 1 microvolt

lator, permitting use of the strobotron for determining rotor speed.

The 200-cycle oscillator has a more or less conventional two-stage RC feedback circuit except that stability is obtained by means of inverse feedback across the amplifier portion. The stability thus obtained is surprisingly good. The frequency drift, greatest during the first hour of use, is not over 1 percent during this period. Oscillators calibrated an hour after being turned on remain within 2 percent of calibration when similarly checked over a period of months. The variable resistor in the feedback circuit provides a frequency adjustment of about ±7 percent to take care of frequency calibration.

A separate power supply was used for the strobotron to avoid the danger of having firing pulses get back into the amplifiers through the main power supply.

The complete circuit was built up from a number of two-tube subchassis units to facilitate rapid replacement of circuits in the event of trouble in production use and to permit better control of circuit element quality in the manufacture of the instruments. In manufacture these amplifier components were built up complete, then individually checked and adjusted for gain, frequency response, and phase shift. Each subchassis is easily removed for replacement by unsoldering the few leads and removing two screws. Replacement of a pretested amplifier section can thus be made in a few minutes.

After construction of an instrument is completed, some simple means of primary calibration are essential before it can be usefully employed. The four items requiring calibration are: (1) test speed or oscillator frequency; (2) unbalance magnitude or overall gain; (3) unbalance position indication or overall phase; (4) focus or mixing circuit balance.

Calibration Procedure

The primary standard for checking oscillator frequency is the 60-cycle power line frequency. A disc having ten equally spaced holes is mounted on the shaft of a small 3,600-rpm synchronous motor which while running is illuminated with the strobotron excited from the oscillator. A stationary pattern is obtained at 200 cycles by catching each third hole during rotation. The next nearest stationary patterns with this method occur at 150 and 250 cycles; since these are definitely beyond the range of the oscillator, this system provides a positive primary calibration.

To calibrate for unbalance magnitude, a standard rotor was prepared with two small holes drilled on each face, precisely 0.700 inch from the center of rotation. These holes were tapped for a 0-80 screw, and the rotor was then carefully balanced. Next a 0-80 screw was prepared which weighed precisely 0.00143 ounce, and inserted into one of the holes provided for it in the standard rotor. Thus a known unbalance weight was added to the rotor at a known distance from the center, providing a standard unbalance of 0.700 × 0.00143 = 0.001 ounce-inch = 1,000 micro-ounce-inches. Since the 0.700-inch location can be easily determined and maintained to an accuracy of a few tenths of a percent and since the 0.00143-ounce weight can likewise be determined to a high degree of accuracy, this primary standard is dependable to 0.5 percent.

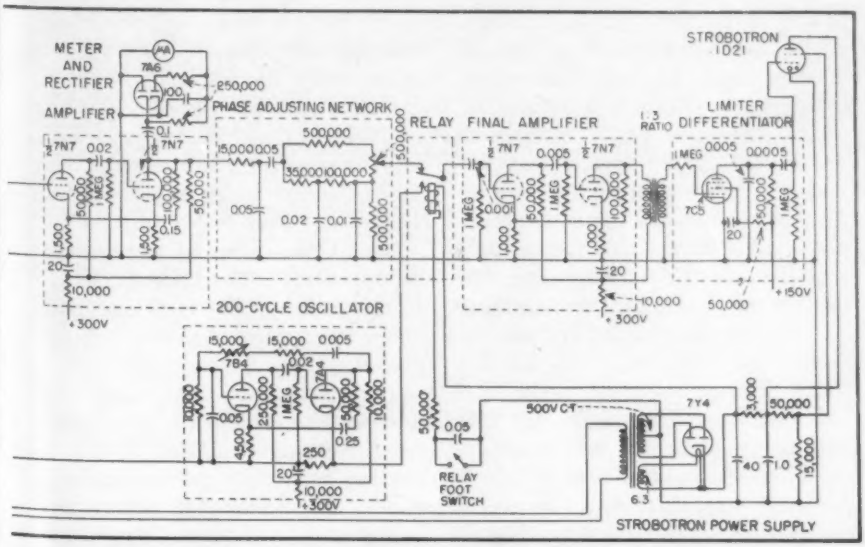
The procedure of calibration thus becomes simple. When the rotor is at test speed with the weight on one side of the rotor, the gain for that channel is adjusted until the meter reads 1,000 micro-ounce-inches or full-scale reading on the D scale. The weight is then transferred to the other side of the rotor and the procedure repeated for the other channel.

With the standard rotor and screw, the calibration of unbalance position indication also becomes simple. With the screw in position and the rotor at exact test speed, the strobotron must illuminate the rotor at the instant the position of the screw is aligned with a reference index line in the operator's field of view. The adjustment in the phase network provides means for a compensating adjustment.

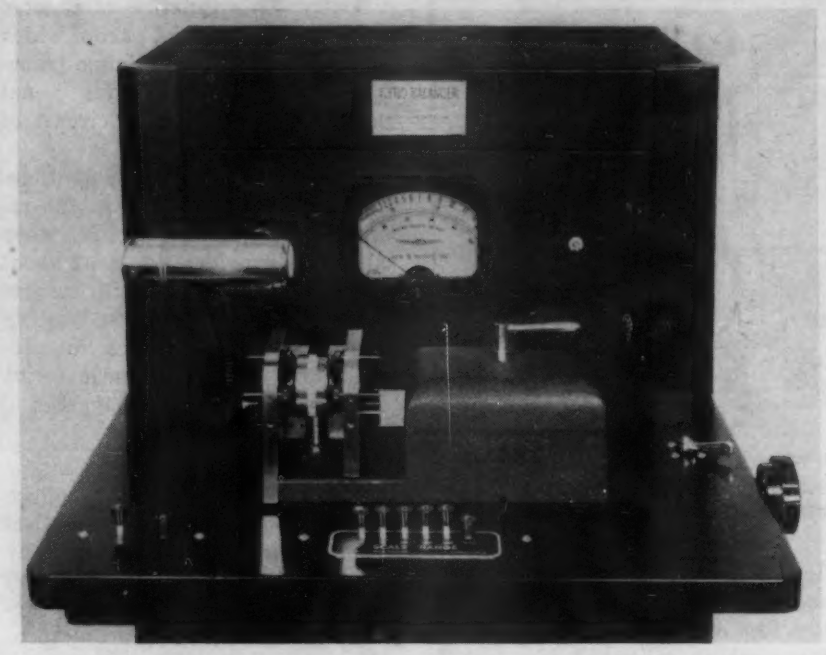
The adjustment for focus is a little more involved and is also performed with the standard rotor and screw as reference. When all the above calibrations have been properly made, and with the standard unbalance on the right side, the standard rotor is operated at test speed. Under these conditions the unbalance reading for the left side should be adjusted for a minimum value. While a null value 20 db below the unbalance reading gave sufficient separation for good production speed, 30-db separation was usually maintained on instruments in production use.

Operational Procedure

With the instruments so designed and constructed and with calibrations properly made, results under actual production-shop operation proved most gratifying. The indi-



was used to offset the severe line voltage fluctuations encountered in industrial plants



Instrument table and front panel of gyro balancer. Speed-checking strobotron is in lamp housing above gyro rotor. Pushbuttons L and R at lower left determine which end of rotor is checked, and buttons at center control range of meter

hance the production use of the instrument.

The phase-adjusting network following the meter circuit makes the necessary calibrating provision for synchronizing the instant of strobotron flash with the location of unbalance on the rotating gyro wheel. This network provides a 50-degree adjustment of electrical phase angle but in so doing adds a transmission loss of 17 db. The loss through the network, however, is constant within ±1 db as the phase angle is changed through the total 50-degree range.

Final Amplifier and Limiter

The signal at this point is again about one volt rms for full-scale meter reading, and is essentially a sinusoidal voltage which may vary by more than a factor of 10 over the useful range of unbalance position indication, which is from about 1/5 of full scale to about a factor of 2 beyond full-scale meter reading. Accordingly, some means must be provided for flashing the strobotron consistently with a high degree of phase-angle accuracy in spite of this large variation of signal amplitude. This function is accomplished by the final amplifier and limiter-differentiator.

The final amplifier and its output transformer serve only to supply a very large signal to the limiter stage; together they produce a voltage gain of about 42 db. Thus for one-volt input, about 125 volts appears at the limiter grid. The limiter is of more or less conventional design and is arranged to utilize only about 1.25 volt of the signal at its grid. The output wave form will, therefore, be a flat-topped wave on which the electrical angular distance represented by the vertical portion of the wave may be considered to be approximately $\sin^{-1}$ 1.25/1.25 or 0.6 degree. For twice full-scale signal this value would be 0.3 degree, and for one-fifth of full-scale signal this value would be 2.9 degrees.

A voltage-regulated power supply to the limiter maintains the limiter output constant in amplitude. Thus we have an output wave constant in amplitude and with less than 3 degrees error in phase angle over a factor of 10 variation of input signal amplitude. If the instrument is calibrated for phase angle at one-third of full-scale meter reading, the angular error in determining unbalance location from this cause is within ±1½ degrees from one-fifth of full scale to twice full-scale indication.

Strobotron and Oscillator

To trigger the strobotron tube, a single sharp pulse is most desirable. This is easily obtained by differentiating the square wave. The 0.0005-μf capacitor and 1-megohm resistor combination comprise the approximate differentiating network. However, to avoid the formation of both sharp positive and sharp negative pulses which may produce double-frequency firing of the strobotron, the 0.005-μf capacitor is used in the plate circuit of the limiter to round off one side of the square wave.

A relay operated by a foot switch may be used to change control of the firing of the strobotron from the unbalance signal to a fixed-frequency signal from a local oscil-

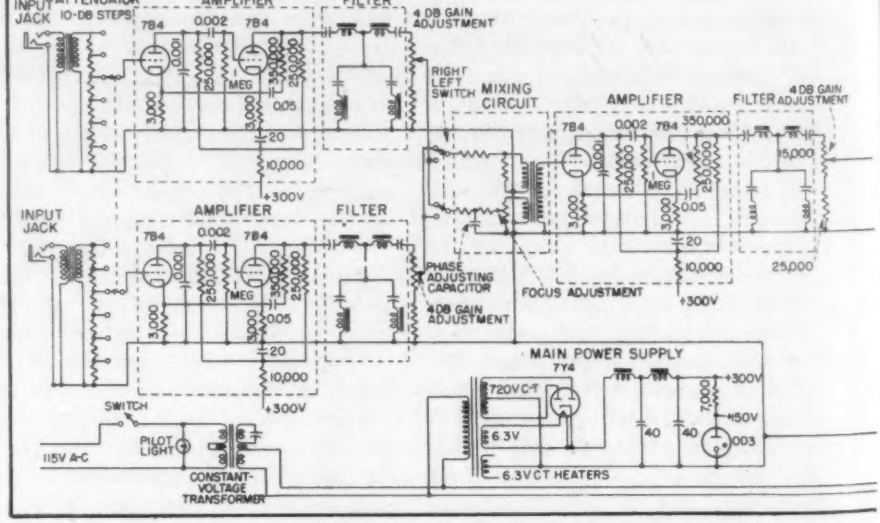


FIG. 7—Complete circuit diagram of gyro balancer. Constant-voltage transformer

cation of unbalance, both as to magnitude and position for both sides of the rotor, can be obtained by an average operator in five to ten seconds, and actually occupies a very minor part of the total balance procedure. A new rotor is inserted in the instrument by the operator, brought up to speed, and checked for unbalance. This data can be recorded on a pad, but is usually remembered by a good operator. The rotor is then removed from the instrument, inserted in a fixture on an adjacent drill press, drilled, cleaned using an air duster, and reinserted in the instrument. This complete process is repeated as many times as necessary (usually three or four) until the desired degree of balance accuracy is attained. This entire procedure is accomplished in two minutes time on the average, including the several drillings and instrument indications.

Initially the drill press was provided with two independent settings, one of which could be set to the scale or range letter and the other to the balance number on the meter dial. This predetermined the depth to which the drill would cut, so that with a standard drill size and point shape, the operator could remove the proper amount of material. It was found, however, that if the operator was allowed to exercise her own skill, she could perform this operation much faster and with no more error. In addition, the psychology of the uncertainty involved greatly aided in increasing her interest in the job and enhanced development of skill in rapidly making a judgment of the amount of material to be removed.

In training new operators, each was provided with a small gage on which were drilled five impressions corresponding to proper amounts of material to be removed for unbalance readings of B7, C7, D7, E7, and F7. By visual reference to this standard, the operator could judge the amount to be drilled out, and in about a week's time would develop skill to the point where she no longer needed the gage.

Accuracy Attained

The speed with which rotor balance can be attained depends primarily on the accuracy of angular location at which correction is made, and this in turn depends on several factors:

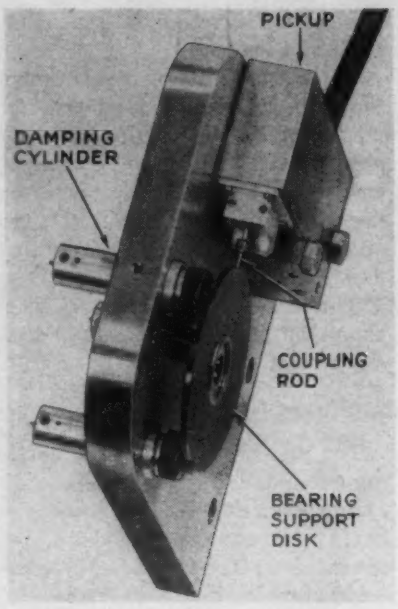
(1) The practical accuracy with which the rotor speed can be matched stroboscopically on production work. This we found to be easily effected within 0.5 percent.

(2) The accuracy with which this speed is maintained while taking an unbalance position reading. This was a function of operator experience and skill; it might have been 3 or 4 percent for beginners, but for experienced operators this was also within 0.5 percent.

(3) The frequency stability of the local oscillator. This remained within 2 percent over long time intervals, and by calibration was maintained within 1 percent.

(4) The overall instrument phase-angle error versus test frequency. This includes the sum effect of mechanical and electrical phase-shift errors and phase-angle stability. In this design the sum total error was maintained within ±2 degrees phase shift per percent frequency deviation.

(5) The phase-angle error versus unbalance magnitude. On this design, the practical maximum error from this source was within 3 degrees phase shift over the useful range when making calibration at full-scale meter reading.



Suspension and pickup assembly used in final design of balancer

The sum effect of items 1, 2 and 3 results in a maximum error from these sources of perhaps 2 percent or 4 cycles. The effect of this on item 4 results in an uncertainty of unbalance position indication of ±4 degrees. Adding to this the error of item 5 gives a maximum uncertainty of 7 degrees, although statistically it would be unusual for all errors to add simultaneously in this fashion. Practically, maximum uncertainty was well within the ±6 degree criterion originally set for the total of all such errors.

The balance accuracy which can be attained on a rotor is not limited by the above errors nor does it depend on the operator's skill. These affect only the speed with which the balancing is performed. The unbalance on any production rotor could be reduced to 10 micro-ounce-inches; with selected rotor pivots and bearings this could be reduced to 3 or 4 micro-ounce-inches. Some rotors, with exceptionally accurate rotor pivots, were actually balanced to 1 or 2 micro-ounce-inches. Tests on such rotors indicated that the balancing instrument could maintain a sensitivity of 1 micro-ounce-inch, better than a rotor's physical stability for this magnitude.

The maximum useful range of the balancer is from 1 to 20,000 micro-ounce-inches. An experienced operator can easily effect a balance on a rotor with an initial 20,000 unbalance down to the production limit of 20 micro-ounce-inches or less in the order of two minutes. The best operators could easily maintain this speed and could average 200 rotors per day per machine while actually operating the machine only 45 minutes each hour to avoid fatigue. The operator holding the record for the greatest number of gyros balanced in one day turned out over 350 rotors.

This amazing speed on so precise an operation is of course partly due to the design of the instrument and the conveniences for the operator included for its use. However, a major portion of the success was due to the excellence of departmental supervision and to the industry, zeal, and enthusiasm with which an unusually outstanding group of girls tackled a tough wartime assignment.

R-F Inductance Meter

Two r-f oscillators, one with unknown L in series and the other with standard C in parallel, are tuned to zero beat. Value of unknown inductance at selected test frequency is then read directly on scale of capacitor with accuracy within 1 percent for 1 μh to 100 mh

By

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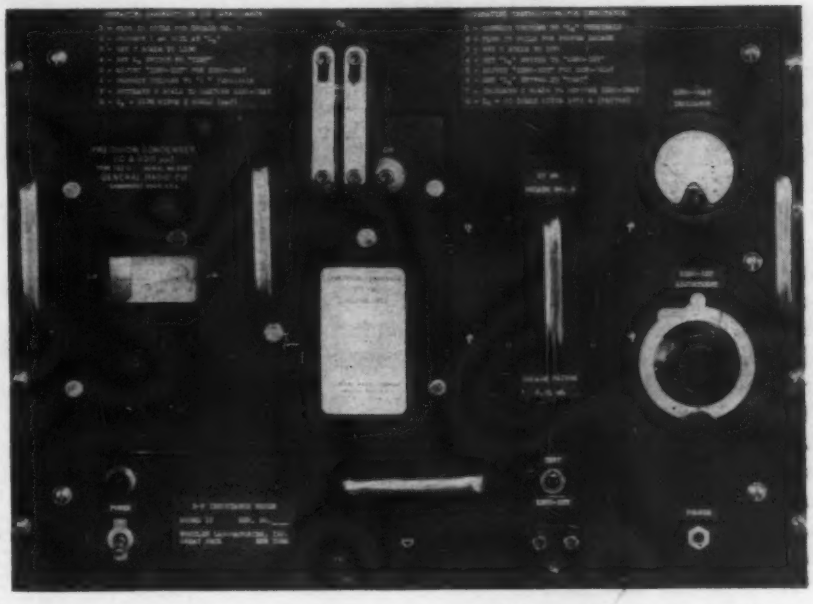


Front panel of r-f inductance meter. Plug-in standard inductances are stored in rack above meter, not shown

THIS direct-reading r-f inductance meter is designed for quickly and accurately measuring inductance of a radio coil. Errors in readings are within about one percent of the observed value for the rated range of inductance, which is unusually close when applied to one-tenth of full scale. This tolerance of error is based on the usual types of radio coils with fairly small inherent capacitance and at least a moderately high ratio of reactance to resistance.

The measurement of small capacitance is a valuable byproduct use of the instrument. It is direct-reading in terms of the difference of two scale readings, and its errors are only those of the precision capacitor included in the inductance meter. This measurement is made at radio frequencies whose values are chosen for convenience and reliability.

Beat-Note Circuit

The block diagram in Fig. 1 shows the arrangement of the inductance meter. Two similar oscillators are tuned to the same frequency by observing a zero-beat indicator. Each oscillator contains a resonant circuit with fixed inductance L_o and fixed capacitance C_o. One oscillator also has a calibrated variable capacitor C connected in parallel. The other has provisions for connecting unknown inductance L in series.

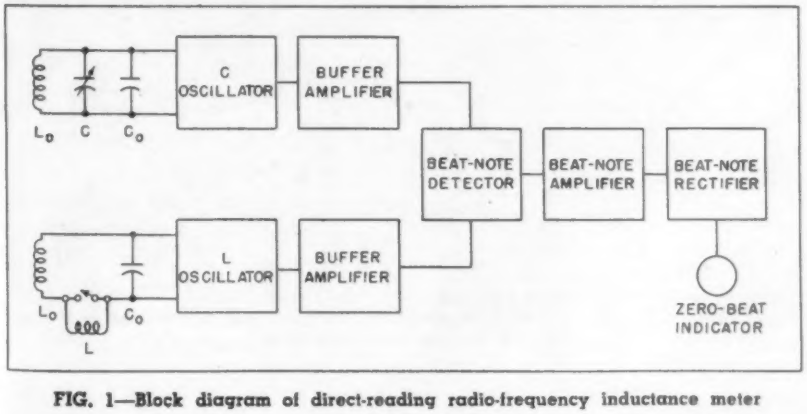


FIG. 1—Block diagram of direct-reading radio-frequency inductance meter

With the standard capacitor C at its reference value (taken as zero) and the unknown inductor L on short-circuit, the two oscillators are tuned to resonance by a fine adjustment. Then the unknown L is inserted and the standard C is increased to restore zero-beat. The latter condition satisfies the equation of resonance of the two oscillators: $(L + L_o)C_o = (C + C_o)L_o$. From this the calibration formula is derived, giving $L = CL_o/C_o$. The factor L_o/C_o is adjusted to be a decimal multiplier, and the scale of C is direct-reading in micromi-

crofarads, so the value of the unknown L is obtained merely by shifting the decimal point.

The circuit of the r-f oscillator and buffer amplifier is given in Fig. 2. The two oscillators and amplifiers are alike except for the differences in the tuned circuit, one including the standard capacitance C and the other the unknown inductance L. The standard inductors L_o are the pair of plug-in coils, one for each oscillator, whose values are equal to the maximum inductance of each range. Each oscillator has a regulator circuit which limits the amplitude of oscillation to a suitable value for the buffer amplifier.

The beat-note circuit is given in Fig. 3. Direct-current coupling of the amplifier to the rectifier-meter combination provides response down to very low frequencies. The full-wave rectifier causes the meter pointer to vibrate at double the beat frequency.

The response curves of the meter are shown in Fig. 4. The curve of Fig. 4A shows the wide peak obtained while tuning one oscillator past the frequency of the other, and the curve of Fig. 4B is an enlargement of the valley near zero beat.

The zero-beat indicator gives essentially a wide peak having a narrow valley in the center. The width of the peak is ±5,000 cycles from the center, so it cannot be passed in quick tuning. The width of the valley is ±100 cycles, the pointer vibrations are within ±10 cycles, and the dead spot in the center is within ±1 cycle. The wide peak is used to find the narrow valley, after which the dead spot gives a reliable indication of zero beat.

This indicator has been found well adapted for use by nontechnical personnel. It cannot be overloaded in normal use. Being visual, it neither causes nor suffers from noise in the laboratory or factory.

Test Frequencies

The choice of test frequencies which are best adapted for the usual radio coils in the respective decades of the inductance range is a compromise which fortunately is not too critical. The final selection is influenced also by the stability of C_o and L_o and by convenience in the use of the decimal values.

The basic value of the inductance of a simple coil in open space is determined by the magnetic energy in the vicinity of the coil but outside the cross-section of the conductor itself. If a shield can is used, this space is confined to the inside of the shield. The inductance would have this basic value if the coil and shield were made of perfect conductors, and its actual value is not much different at the higher radio frequencies.

The resistance of the wire and the shield causes an increase in the actual or apparent inductance of the coil, in proportion to the magnetic energy which is allowed to penetrate inside of either conductor and outside of the shield to the surrounding space. The internal capacitance increases only the apparent inductance, by virtue of the approach to self-resonance.

The magnetic energy in the wire increases the inductance by a constant amount at low frequencies, and by a decreasing amount at higher frequencies. The transition occurs at the frequency where the skin depth is one-fourth the wire diameter if the wires are well separated, and at a slightly lower frequency if they are close together. In the case of fine wire, such as in stranded (Litz) conductors or in multilayer coils, this effect is constant at all frequencies up to the operating frequencies, so it causes no error in inductance measurements at lower frequencies. In a coarse wire, such as the conductors of a transmission line, this effect varies slowly with frequency as governed by the skin effect, so a measurement at a lower frequency is subject to a slight error relative to the operating frequency.

A perfect shield decreases the inductance by a substantial amount, while the imperfection of the shield

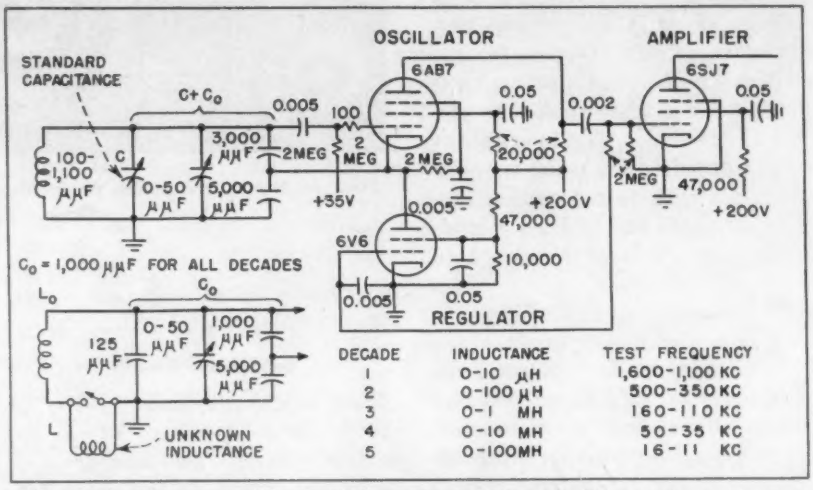


FIG. 2—Circuit for one of the oscillators with its buffer amplifier. Both are identical except for the tank circuits; one has the standard capacitance in parallel, and the other has the unknown inductance in series as shown at left

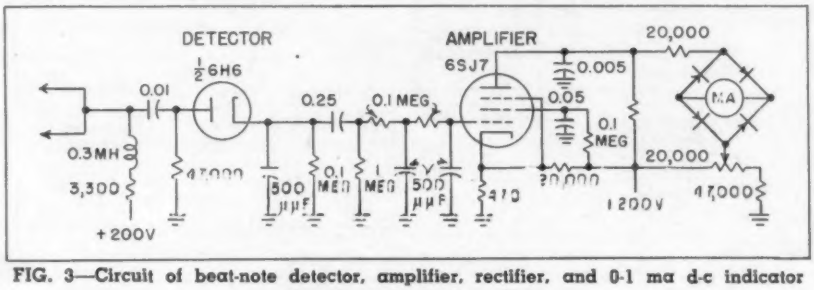


FIG. 3—Circuit of beat-note detector, amplifier, rectifier, and 0-1 ma d-c indicator serving the pair of r-f oscillators

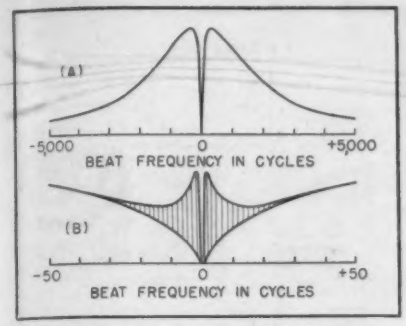


FIG. 4—Response of zero-beat indicator is shown at A, and enlarged view of valley near zero beat appears at B

conductor causes a relative increase. As the frequency is lowered, the inductance increases gradually as governed by the skin effect in the shield, then rapidly as the conductive shielding fails, and finally approaches the value of the inductance without the shield.

The internal capacitance causes the apparent inductance to increase rapidly as the frequency approaches self-resonance in the coil. This is the factor which makes impossible the direct measurement of inductance by a single observation at the operating frequency.

The best frequency for a direct measurement of inductance, with least error from all causes combined, should be much lower than the frequency of self-resonance but much higher than the frequency of shield failure. Fortunately these two frequencies are usually far apart. Three decades is a typical separation in the case of small coils in shield cans, operating around 1 mc.

The internal capacitance is the most typical characteristic for small radio coils, since it is determined mainly by the size, the leads, and the terminals. The typical range of internal capacitance is within 10 $\mu\mu$f, so the inductance meter is standardized for about 5 $\mu\mu$f with a tolerance of ± 5 $\mu\mu$f to maintain the rated limits of error.

Since there are many causes of error and it is desired to hold the total error within one percent, the test frequencies are chosen low enough so that ± 5 $\mu\mu$f causes less than ± 0.25 percent error. In other words, the unknown inductance would resonate at the test frequency with more than 2,000 $\mu\mu$f capacitance, so the test frequency is about 1/20 the frequency of self-resonance, or lower.

Sensitivity of Measurement

The 5,000-division continuous linear scale is available only in the precision capacitor, as distinguished from the inductor and the resistor. Therefore the capacitor is the ideal variable standard in any system of measurement, and especially in a direct-reading system.

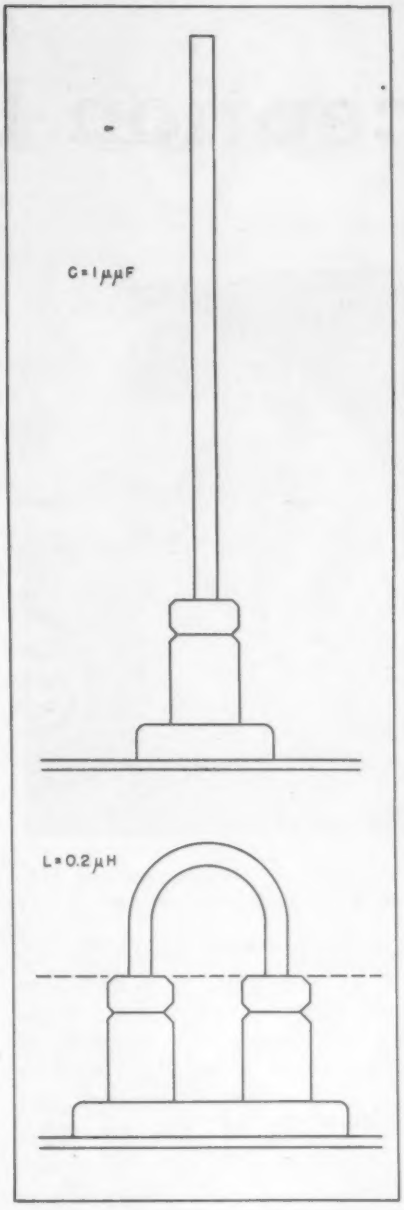


FIG. 5—Examples of small values of capacitance and inductance, shown actual size here, that can be measured with the r-f inductance meter

The inductance meter is a simple way of utilizing the capacitor to read directly in proportion to the unknown inductance.

By making the test at the highest frequency permissible for each decade of the inductance range, it is possible to take full advantage of the long scale of 5,000 divisions for 1,000 $\mu\mu$f, by reading to about $\frac{1}{2}$ division or 0.1 $\mu\mu$f.

In the first three decades (up to 1 mh), the test frequency is above 100 kc so it is possible to take advantage of the low range of the precision capacitor, having 50 divisions per $\mu\mu$f, or the equivalent of 50,000 divisions for full scale. This refinement is available only for a range of 4,000 divisions. Therefore it is useful for absolute measurements only to 8 percent of full scale.

By the addition of fixed capacitors in parallel, this expanded scale becomes available for differential tests of coils nearly alike. By this method, any inductors within the range of 1 μh to 1 mh can be compared to within 1/5,000 of their values, or 0.02 percent. Any larger inductors up to 100 mh can be compared on the standard scale within 1/500 of their values, or 0.2 percent.

Examples of the absolute sensitivity of this instrument, which is attained by the long scale and the beatnote comparison at radio frequencies, are given in Fig. 5. These diagrams are shown actual size.

The capacitance C is a small antenna inserted in the capacitance terminal. Its value is 1 $\mu\mu$f, which is 5 divisions on the regular scale or 50 divisions on the expanded scale.

The inductance L is a small loop of heavy wire connected to the inductance terminals. As compared with a short-circuit (a large flat disc at the dotted line) the inductance is 0.02 μh, corresponding to 10 divisions on the regular scale or 100 divisions on the expanded scale.

By indicating inductance values on the expanded scale of a precision capacitor and using r-f oscillators with a zero-beat indicator, an inductance meter is obtained which has an unusual degree of absolute and differential stability as well as high accuracy.

Built-in and outdoor antennas for f-m reception, front-end and i-f circuit arrangements, and demodulation systems are discussed in this symposium.* Included is a useful method of testing an antenna by comparing the voltage delivered at the receiver to the voltage obtained from a reference antenna

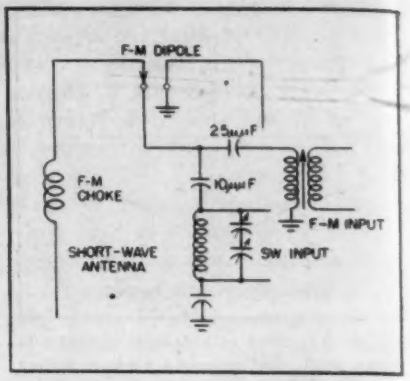
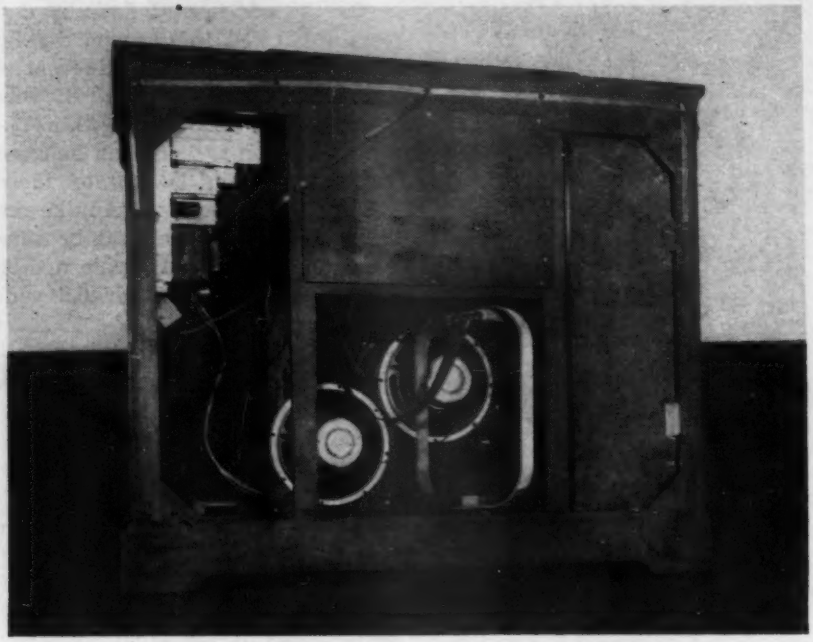


FIG. 1—Simple combination antenna covering one f-m band and short waves

F-M Reception Problems



A simple built-in f-m antenna is a folded dipole made of 300-ohm twin lead and tacked to the back of the cabinet as in this Hallicrafter receiver

THE AVERAGE USER of an f-m broadcast receiver has a right to expect that the installation and operation of his set should be no more complicated than for the standard broadcast set he bought ten years ago.

Since receivers, like population, are concentrated in cities and towns, and f-m transmitters are being located at the centers of these concentrations, f-m coverage is essentially a metropolitan or urban proposition. This means that although some rural areas as well as some outlying towns and villages will be served, the majority of listeners will be relatively close to the transmitters, where high field strengths prevail. For this reason it is feasible to provide built-in antennas in f-m receivers and to expect them to operate satisfactorily in most installations.

Built-in antennas for f-m sets involve some problems that are not found in standard broadcast sets. This is so because the buildings in which the sets are used are large compared with the wavelength. The dimensions of the rooms in which sets are used, and even the dimensions of the radio cabinets, are comparable with the wavelength. Therefore, large standing wave patterns are frequently found to exist, and small differences in the position of the radio set within a room, or even the position of people moving within the room, may make a relatively large difference in the received signal strength.

This means that the performance of built-in antennas is variable, and one type may perform well in a particular location while another type will perform better somewhere else. In most metropolitan locations the signals are strong enough that these problems are completely cared for by the limiter of a sensitive receiver.

Double Antenna

Figure 1 shows one type of built-in antenna. It is a dipole constructed of wire and fitted to the inside of a console. The arms of the dipole are bent or folded so that they will fit into a cabinet of reasonable size. The antenna shown in this figure also includes a length of wire used for shortwave reception. This wire is isolated from the dipole by means of a choke having high impedance to the f-m frequencies and relatively low impedance to the short-wave frequencies. The receiver has separate input circuits

*From papers presented before the Chicago Section and the Indianapolis Section of IRE

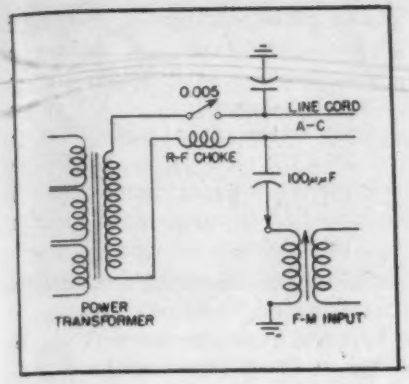


FIG. 2—Circuit for a line-cord antenna suitable for reception of f-m signals

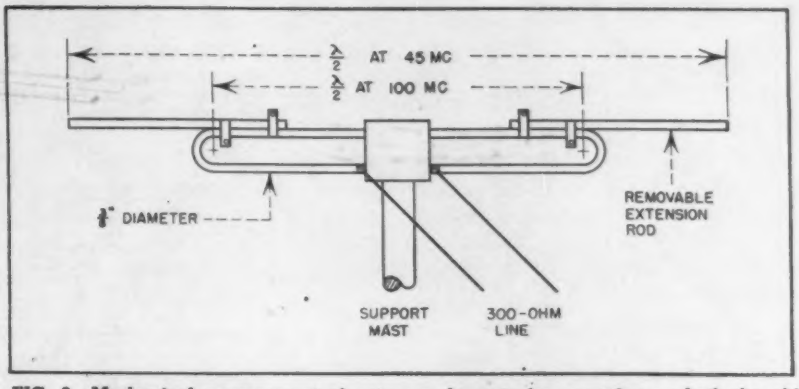


FIG. 3—Mechanical arrangement of antenna elements for reception on both the old and the new bands

and Their Solution

for f-m and shortwave, and these are tied together at the antenna terminals by means of a network which prevents undue loading of one circuit by the other.

Another form of built-in antenna is shown in Fig. 2. This is the familiar line-cord antenna adapted to f-m reception. Since one side of the line cord must be bypassed to reduce line noise in the broadcast section of the receiver, it is essentially the signal voltage between the two conductors of the line cord which is used as input to the f-m circuits.

In some locations the cabinet dipole will work better, and in other locations the line-cord antenna will work better. Since they are inexpensive, both can be provided in a console f-m set and the user may select the one which works better in his location. Table model radios are generally too small to accommodate the cabinet dipole, and for this application the line-cord antenna is preferable. Both of these antennas provide good reception on the low and high f-m bands.

Both F-M Bands

In outlying districts where built-in antennas do not give satisfactory performance, a several-fold improvement may be had by using a simple outdoor antenna with low-loss leadin. Figure 3 illustrates such an antenna which works satisfactorily on both the old and the new f-m bands. This antenna consists of a 100-mc, half-wave folded dipole fitted with extension arms that build up the length to a half wavelength at 45 mc.

The efficiency of the antenna is somewhat better on the low band than on the high band. If low-frequency f-m broadcasting should be discontinued or the band changed, the extension rods can be discarded and the antenna will function with full efficiency on the high band only.

With an outdoor antenna the performance is much less variable than with a built-in antenna. It becomes meaningful to measure the performance and obtain curves showing the effect of frequency over the band in which we are interested. There are several characteristics of antennas that permit critical measurements for the purpose of comparison. For example, it is important in television antennas that the impedance closely match the impedance of the transmission line at all the operating frequencies. This measurement may be expressed by a curve of impedance against frequency, standing-wave ratio of reflection coefficient, or attenuation due to mismatch. Antennas may also be measured in terms of their directional patterns or signal gain.

For f-m service, impedance matching is not a major consideration. It is quite satisfactory to test an antenna in terms of voltage delivered to the receiver. In doing this it is convenient to compare the voltage with that obtained from a standard or reference antenna. The common reference is a resonant half-wave dipole. The setup for making this test may be of interest, and Fig. 4 shows the method used in obtaining a curve for the two-band antenna just shown.

Test Technique

Tests are made using the antenna for receiving. At each frequency an arbitrary field is set up by a third antenna connected to an oscillator some distance away. The reference antenna is then adjusted to resonant length, placed in the test position, and the voltage V_1 delivered to the matched load is noted. The reference antenna has an impedance of approximately 70

Antennas for F-M Receivers

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Engineering Department
Zenith Radio Corporation
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R-F, I-F, and AFC Circuits

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Engineer-in-Charge
Home Receiver Division
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Limiters and Frequency Detectors

By MARVIN HOBBS
Consulting Radio Engineer
Chicago, Illinois

better still, the crystal mixer, are extremely quiet; however, stage gains less than unity are common using these mixers, and a quiet first i-f stage amplifier tube is required if either is used.

The oscillator stability and tuning accuracy requirements for an f-m receiver are considerably greater than for an a-m broadcast receiver. The a-m receiver must be tuned within about ±2 kc in a 1,000-kc band (or the dial must be set to about ±0.2 percent) if noticeable distortion is to be avoided. The f-m receiver must be tuned within about ±20 kc in a 20-mc band, or about ±0.1 percent. The oscillator must be stable to within about two parts per thousand in the standard band, and to about two parts per 10,000 in the f-m band.

AFC Circuit

The nature of the oscillator stability and tuning accuracy problems make the use of automatic frequency control on the f-m band attractive. If the oscillator and mixing functions are performed in separate tubes, afc can be provided by adding a few resistors and capacitors, and by using a dual-triode tube as the oscillator and reactance tube. Such a system is used in several Hallicrafters home receivers.

A circuit of this type is shown in Fig. 6. The plate of the oscillator is capacitance coupled to the grid and plate of the reactance tube. The resultant voltage on the grid of the reactance tube is shifted in phase approximately 45 degrees by resistor R_1, capacitor C_2, and the capacitance from the reactance tube grid to ground. This voltage is then amplified by the reactance tube according to the value of direct voltage on its grid and appears as a current in the oscillator plate feedback winding, shifted in phase from the oscillator plate current by about 135 degrees. The d-c component of the output of the frequency detector can be used as the control voltage applied to the grid of the reactance tube.

The slope of the reactance-tube characteristic can be adjusted so that the afc system takes hold and releases within the frequency limit of the next adjacent local channel (400 kc), and a correction of about five to ten times on the oscillator mistuning can be realized. Under these conditions, the afc operation prevents the receiver from reproducing the two spurious responses, which occur on each side of the desired response points in a receiver without afc.

I-F Amplifier

The i-f amplifier in an f-m receiver is required to produce a major part of the total gain and the selectivity of the receiver. The selectivity is usually made as great as possible without harmful effects upon the signal; that is, when the f-m signal swings back and forth on the curve of the i-f selectivity characteristic, the result is amplitude modulation. The limiter or detector must be able to remove this a-m and restore the signal to its original form to avoid distortion in the audio output. At the same time, a small amount of phase distortion is produced due to the nonlinear phase characteristic of the i-f amplifier when the sidebands of the signal are too far from the center frequency in terms of the bandwidth of the amplifier. As a practical matter, the amplitude variations in a typical i-f amplifier are far more troublesome than the lack of linearity of the phase characteristic.

In the most popular design for i-f amplifiers, the bandwidth of the amplifier is made about equal to the total deviation. This seems to include an adequate allowance for error in tuning, oscillator drift, and other factors in a well designed receiver. Each i-f transformer consists of a pair of tuned circuits coupled to slightly less than critical coupling, so that normal variations in coupling can be allowed in production without some transformers being over-coupled.

It is the usual practice to make the Q of the primary equal to that of the secondary, since the passband is too narrow to allow any appreciable improvement in gain with practically obtainable ratios of Q. Moreover, unequal values of Q for the primary and secondary cause

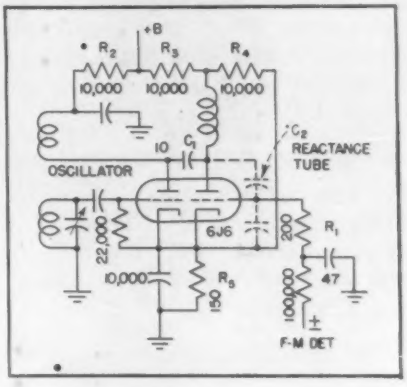


FIG. 6—A tuned-grid oscillator and afc reactance-tube circuit

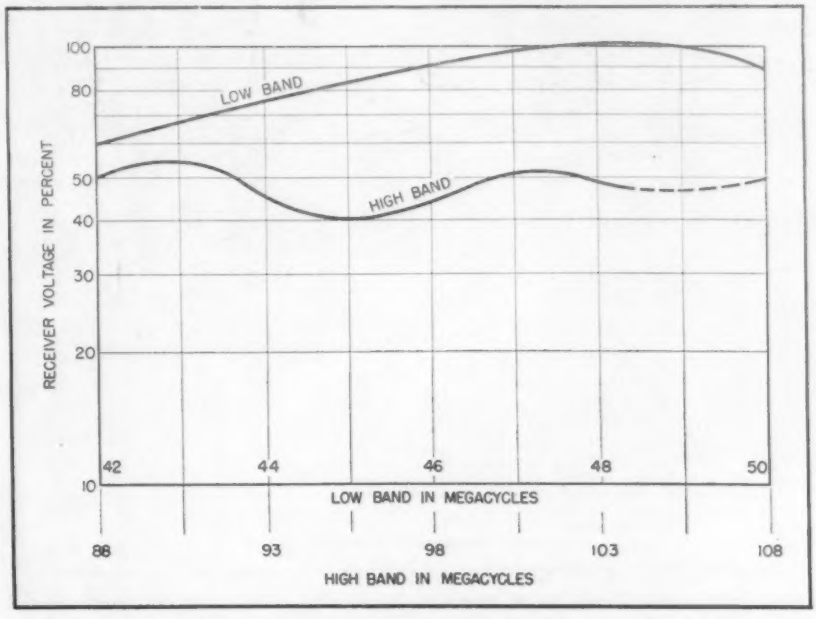


FIG. 5—Receiver voltage for a two-band antenna with 300-ohm line feeding into a 300-ohm input compared with a matched-impedance resonant dipole

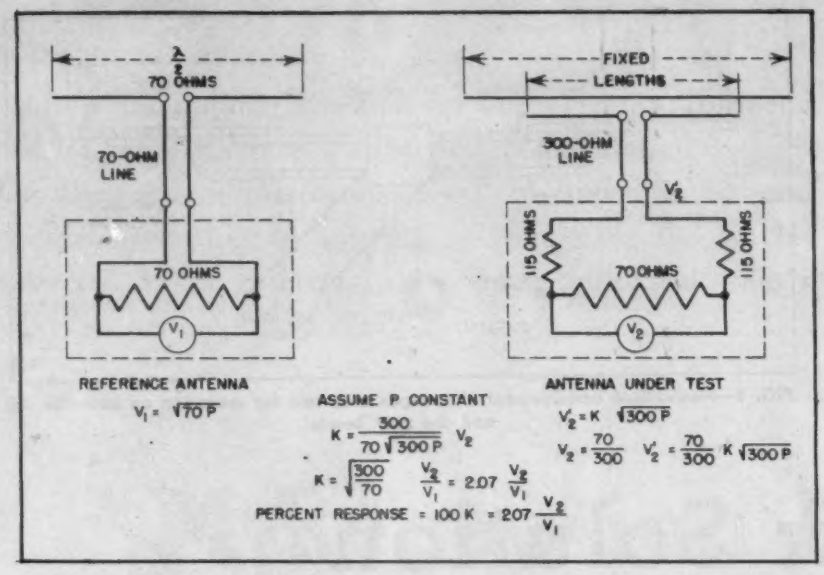


FIG. 4—Method of comparing the two-band antenna against a resonant dipole

ohms, the transmission-line characteristic impedance is 70 ohms, and the load impedance is 70 ohms. The load consists of the input attenuator of a field intensity meter.

The antenna under test is then substituted for the reference antenna at the test position. Since the receiver for which this antenna was designed has an input impedance of 300 ohms, 300-ohm transmission line is used and the field intensity meter input is built up with resistors to match the 300-ohm line. When this is done the field intensity meter input circuit duplicates an ideal receiver input, but the voltage indicated by meter V_2 is equal to only 70/300 of the voltage delivered to the receiver.

By comparing readings V_1 and V_2 of the field intensity meter, the performance of the antenna under test is related to the performance of the standard antenna. Here it is assumed that the power delivered to the receiver by the standard antenna is the maximum power available from the antenna under test, and the fraction of the available power actually obtained is a figure of merit for the antenna. Since receiver measurements are made in terms of voltage rather than power, it is the square root of this figure or the voltage ratio K which is actually used.

After this measurement has been made at a number of frequencies, a performance curve can be plotted. This is shown in Fig. 5. In judging this curve, it should be borne in mind that the reference antenna has been readjusted in length for each frequency. Plotted in this same manner, a simple dipole of fixed length would show considerable selectivity while a simple folded dipole would be relatively broad and comparable to the low-band performance curve shown. If the extension arms of the two-band antenna were removed, the high-band curve would look very similar to the low-band curve as plotted, but the low-band curve would suffer seriously.

Experience indicates that built-in antennas of the line-cord and cabinet dipole types give satisfactory set operation at practically every location in the Chicago metropolitan area on either the low or high bands. At distances of about forty miles, a simple outdoor antenna becomes useful in many installations; and at greater distances, approaching the nominal service area limit, the outdoor antenna is required in all but exceptionally good locations.

R-F and I-F Circuits

In considering the circuit elements between the antenna terminals and the output of the i-f amplifier, it must be noted that the best distribution of gain in an f-m receiver is determined largely by noise considerations. A certain amount of noise is produced by thermal agitation in the antenna circuit. If the maximum signal-to-noise ratio is attained in the antenna circuit and the amount of noise added in later stages is very small compared to that in the antenna circuit, the signal-to-noise ratio of the receiver can be made to approach (as a maximum) the ratio established in the antenna circuit.[1]

The maximum antenna stage gain is $0.5\sqrt{Z_t/Z_1}$ where 0.5 accounts for matching to the source impedance; Z_t is the impedance into which the antenna transformer works, and Z_1 is the source or transmission-line impedance. It is assumed that the impedance of the secondary of the antenna transformer is considerably greater than the input loading of the tube, which is usually the case, in a receiver using an r-f stage.

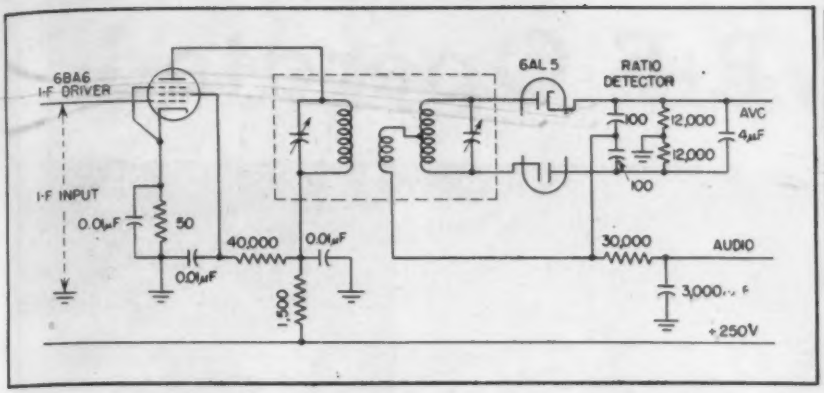
Input Stage

The r-f amplifier tube is chosen primarily on the basis of low input conductance, low noise, fairly high mutual conductance, low grid-to-plate capacitance, and low cost. In an a-m/f-m receiver, signal-handling ability with avc applied may be an important consideration. Since these factors are not combined in any single tube, the final choice represents a compromise.

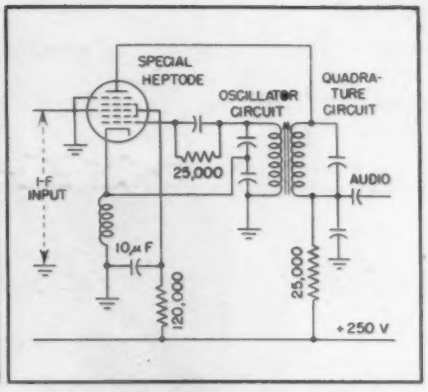
The maximum stable r-f stage gain in a variable-capacitor tuned stage having a solid rotor shaft is about 10 times. If the antenna and r-f circuits are isolated from each other as in permeability-tuned receivers, somewhat greater r-f stage gains can be safely realized.

The amount of r-f stage gain required depends upon the relative amounts of noise at the r-f tube grid and the mixer grid. If the same type of tube is used as r-f amplifier and converter, the mixer tube noise voltage is approximately two times the r-f amplifier tube noise. To avoid any substantial decrease in the signal-to-noise ratio, therefore, a gain greater than five times is required in the r-f amplifier stage.

The noise voltage developed in a mixer tube depends largely upon the type of tube used and its mutual conductance. Of the commonly used types of mixer tubes, the pentagrid converter is the noisiest, the pentode is less noisy (and frequently a good compromise) and the triode is the quietest. The diode, and



Ratio detector system for an i-m receiver employing a transformer having a tertiary winding



Simplified circuit of the lock-in oscillator and quadrature detector combination

tubes to furnish maximum gain. Lower priced receivers often employ a single limiter, which results mainly in a loss of overall sensitivity and a limitation of the range of signal levels over which amplitude modulation is fully suppressed.

The main factors considered in the selection of proper constants for the discriminator include an adjustment of the primary to secondary coupling and circuit Q to provide the optimum degree of linearity throughout the working range of the detector. Circuits employing either variable capacitance or variable inductance tuning have been developed.

The limiter-discriminator combination has the advantages of being unaffected by modulation variations within the operating range of the limiter and of being relatively noncritical in so far as discriminator linearity and balance are concerned. It has the disadvantages of requiring a high gain before the limiter and of a broadening selectivity characteristic with increasing signal level.

Ratio Detector

Due to the inherent ability of the ratio detector to suppress amplitude modulation at comparatively low input levels, it offers a desirable opportunity to those who wish to incorporate reception facilities at a minimum cost. This characteristic is particularly advantageous in those receivers employing common r-f and i-f amplifier tubes for both the a-m and f-m channels and has led to the production of complete a-m/f-m receivers incorporating as few as seven tubes. If the ratio detector is properly designed these receivers can provide acceptable performance. Unfortunately, many of them do not suppress amplitude modulation or provide low distortion levels.

One chief requirement for the proper suppression of amplitude modulation is careful balancing of the two diode circuits. The use of bifilar secondary windings of high Q and a proper choice of circuit constants can produce very beneficial results in this respect. The required linearity for low distortion is not difficult to obtain with a well balanced circuit, but many production sets suffer due to excessive selectivity ahead of the detector. The i-f selectivity often varies due to regeneration or coupling coefficients not controlled in low-cost receivers, resulting in high downward modulation during wide frequency swings and considerable distortion.

These difficulties with the ratio detector can be overcome provided a careful design is made and a high degree of quality control is exercised in production. The chief advantage of the system is its ability to suppress amplitude modulation at lower signal levels than is possible with limiter sets.

The main disadvantages are the problem of handling high percentages of downward modulation, produced by multipath signals and excessive selectivity, and the fact that avc voltages must be applied to several high-frequency stages with the possibility of shifting their tuning and the symmetry of the selectivity characteristic.

These disadvantages can be overcome to a great degree if the audio output requirements for the detector are reduced and provided enough i-f stages can be employed to allow the use of high grid circuit capacitances in the amplifier circuits which are automatically biased.

Locked-in Oscillator

The locked-in oscillator operates in a manner similar to the double limiter arrangement throughout the lock-in range of the oscillator, but it may distort badly when the signal level drops below the lock-in threshold. It has been employed by one manufacturer commercially and in several forms experimentally, such as the use of a submultiple oscillator to convert to narrow band f-m before detection and the use of combined oscillator-detector circuits.

In general the advantages and disadvantages are of the same general nature as those of the limiter system. An important consideration is the necessity for eliminating any extraneous frequency or phase modulation of the locked-in oscillator by filament, screen or plate voltage variations.

The locked-in oscillator has the inherent capability of being less susceptible to interference than any of the three systems described, but practically its entire performance depends on the degree to which the oscillator can stay ideally locked to the desired signal.

References

(1) E. W. Herold, An Analysis of the Signal-to-noise Ratio of Ultra-high-frequency Receivers, *RCA Rev.* 6, Jan. 1942; D. O. North, The Absolute Sensitivity of Radio Receivers, *RCA Rev.* Jan. 1942.

(2) J. R. Nelson, *Proc. IRE*, Feb. 1929.

asymmetry of the amplitude and phase characteristics when either primary or secondary is detuned relative to the other circuit.

The gain of the i-f amplifier is usually made as high as cost and stability considerations will permit. If maximum sensitivity is to be realized, an optimum value seems to be enough gain to assure saturation of the limiter (or a comparable noise value at the ratio detector). This allows satisfactory reception of any signal which is of sufficient intensity to overcome the noise voltage at the limiter or ratio detector. A margin of about 10 db over the i-f gain required to produce this saturation by noise is needed to insure maximum performance under all conditions.

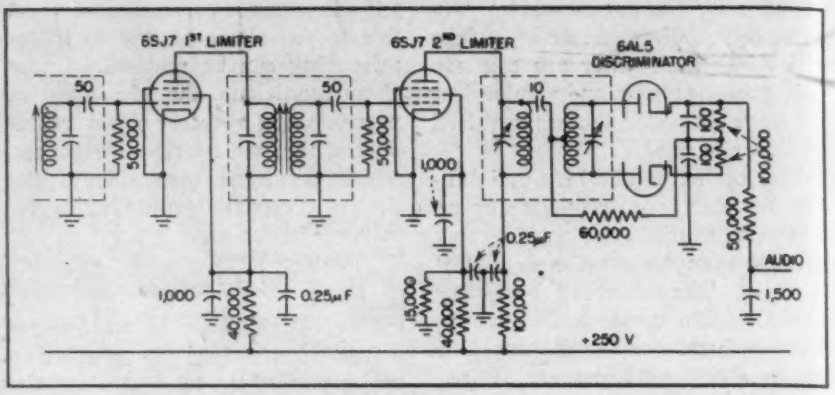
Maximum stable stage gains may be computed by means of a formula given below.[3] It is usually advisable to limit the stage gain to somewhat less than one-half of the calculated value for safety.

$A = \sqrt{\frac{g_m}{\omega C_{pg}}}$ (multiply A by $1/\sqrt{n}$ for n stages) where A = maximum stage gain, g_m is in micromhos, ω is in radians per second, and C_{pg} is the effective plate-to-grid capacitance in farads.

High-Capacitance Grid Circuit

There is one departure from conventional i-f transformer design that has been developed recently. The usual value of tuning capacitance on an i-f transformer primary or secondary is about 50 $\mu\mu$f or less. The input capacitance of the commonly used tubes varies about 2 $\mu\mu$f when the grid voltage is varied from the normal bias value to cut-off. This means that impulse-noise voltages on the last i-f amplifier tube or limiter tube can produce detuning of the i-f transformer secondary, and thereby introduce phase distortion in the signal. The limiter removes most of the amplitude modulation of the carrier due to the noise pulse, but it cannot remove the phase distortion. Remarkable improvements in signal-to-noise ratio have been reported when the secondary tuning capacitances in the last i-f and limiter grid circuits were increased.

The gain of a double-tuned i-f stage is inversely proportional to the square root of the product of the values of the tuning capacitances. This suggests that, since the tube output capacitance is substantially constant, the stage gain obtained from a symmetrical transformer can be duplicated with only the output capacitance and strays tuning the primary, and the secondary capacitance increased by the same ratio by which the primary capacitance was decreased. It is desirable to keep the Q of the primary equal to that of the secondary.



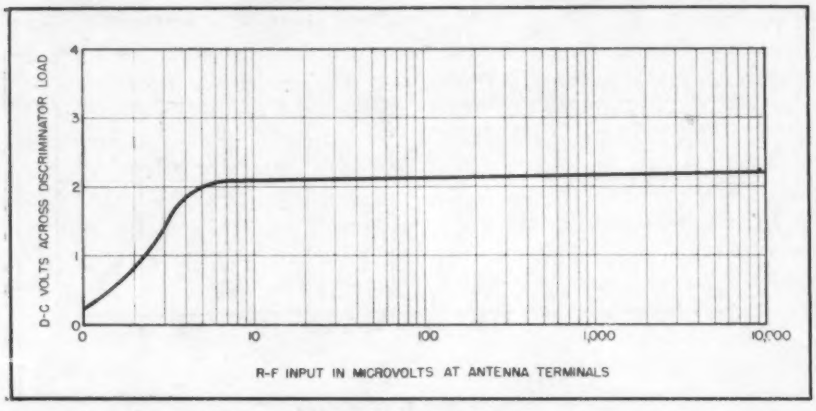
Receiver circuit with a double limiter and a discriminator. Capacitor values in $\mu\mu$f unless otherwise noted

Three types of wide-band f-m detector systems have appeared in production receivers since the war, namely—the limiter-discriminator combination, the ratio detector and the locked-in oscillator. For practical considerations, it seems best to evaluate these systems in terms of the overall performance of the various receivers in which they are employed. However, the detector is often blamed unjustly for the poor performance of receivers in which cost considerations have led to sacrifices throughout the complete circuit. All of the systems mentioned have certain disadvantages as well as their own peculiar advantages. Any of above systems can be made to perform quite satisfactorily.

The Limiter-Discriminator

It has become customary to employ two grid-plate limiters in cascade in the more elaborate receivers using this system. The parameters for each limiter are adjusted so that the maximum range of linearity without appreciable change of output level is realized. Grid circuit constants are chosen to provide a low time constant for optimum limiting of impulse noises.

Since adequate gain must be provided ahead of the limiters to realize full action at low antenna input levels, they are usually preceded by two i-f stages, a mixer, and one r-f stage, preferably with separate



Input-output characteristic of a sensitive limiter-type receiver which shows that practically flat response may be obtained without the use of avc. The long flat portion at high signal levels indicates ability to handle downward modulation. Curve obtained with carrier deviated ±50 kc from center frequency of detector

R-F Operated

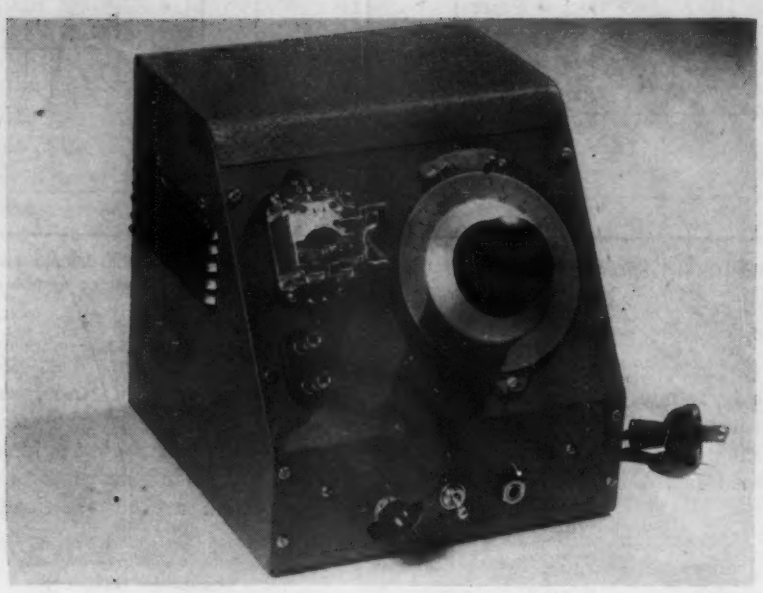


FIG. 1—External appearance of the carrier-operated relay. Shown on panel are the antenna terminals, micro-relay, tuning control, and (at bottom) sensitivity switch, automatic reset switch and jack for headphone monitoring

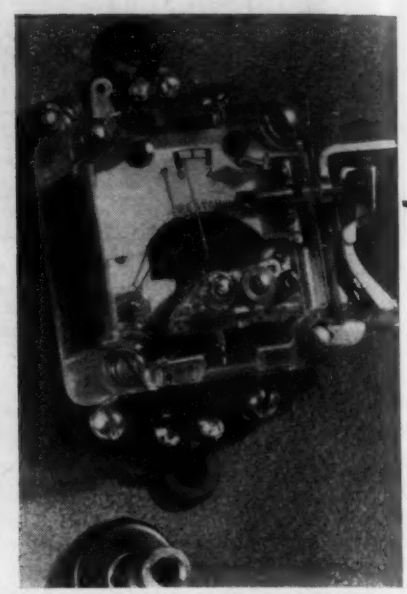


FIG. 2—Close-up view of Weston micro-relay, which closes on 2 microamperes d-c. supplied by tuned circuit and crystal

WHENEVER it is necessary to control a relay at a remote unattended location (a problem which often arises in communications and industrial technology), the supply of standby power assumes importance. If power is available, a relay controlled by a continuously operating radio receiver suffices, but the reliability of such a system is always in some question because of possible failure of power supply or tubes. If a high degree of reliability is required, and if power is at a premium, a wire control circuit is often pressed into service despite its high initial cost.

Several months ago it occurred to the writer that the reliability and economy offered by the wire circuit could be achieved by radio control, provided that all standby power were eliminated in the receiver and sufficient sensitivity achieved to permit operation at reasonably long distances.

The combination of a crystal detector and a sensitive microammeter-type relay suggested itself as a possible solution to the problem. Accordingly available relays of this type were investigated. The most sensitive device available is the meter-type relay which closes a circuit by magnetic attraction between contacts on the meter pointer and the pointer stop. One such meter-relay, the Weston model 813, closes with a current of 2 microamperes through a 1000-ohm coil, which represents the minute power consumption of 0.004 microwatt. It will carry a contact current of 50 milliamperes, which is more than sufficient to actuate a rugged power relay.

Expressed in voltage, the relay will close with 2 millivolts, d-c. Combined with a silicon or germanium crystal rectifier, such a relay will respond to an r-f excitation of from 5 to 15 millivolts, depending on the detection efficiency. Since a 100-watt transmitter will produce a 10-millivolt-per-meter field at about 5 miles, under ground-wave conditions, or at vastly greater distances when skywaves are called into play, it would appear that such sensitivity would serve in many applications.

Among the uses to which such a carrier-operated relay might be put are: a carrier-failure alarm which requires no power while monitoring but which will sound an alarm when the carrier disappears for any period longer than a second or so; a ringing device for radiotelephone circuits to remote locations having limited power resources, such as fire towers; a switch for turning on any remotely located electrical device (transmitter, receiver, photoelectric illumination control, and others) when the periods of operation are so widely spaced that standby power is of economic importance.

Relay Specifications

The device, as finally constructed, responds to antenna current with a sensitivity of approximately 10 millivolts, r-f, and tunes over the broadcast and medium frequency spectrum from 700 to 2,300 kc. This frequency range was selected because of the desire to test on skywave signals, and because many high-powered transmitters occupy this region. Operation at 30 or 40 mc is equally feasible, merely by changing the coils, and is perhaps preferable when the effects of interference are considered. The device consumes no power whatever

Remote Control Relay

Consuming zero standby power, microsensitive relay responds directly to the current in a receiving antenna, closing on 8 to 10 millivolts, r-f. Useful in many applications at unattended locations or as a carrier-failure alarm

By DONALD G. FINK

while in the standby condition. The final control is exercised by a power relay with 10-ampere contacts, which draws four watts of coil current when actuated from a 115-volt line.

The microsensitive relay closes through magnetic attraction, and it remains closed until reset. Since unattended operation is required, an automatic thermal resetting arrangement is provided. When the relay is actuated, it remains closed for from 10 to 60 seconds, and then automatically resets to the standby position. If the control carrier is present continuously, the device continues to recycle until the carrier is removed, thus providing a recurring closure particularly suitable to ringing applications. The device may also be reset manually and the reset mechanism may be disconnected if it is desired that the relay remain closed after the initial actuation.

By adjustment of the tension of the meter-relay coil spring, the microsensitive relay may be operated with contacts normally open or normally closed. In the normally-open position, the relay remains inactive until the control carrier signal appears. In the normally-closed position, with the rectifier polarity reversed, the relay is held open by the presence of the carrier, and closes (actuating the power relay) only when the carrier is interrupted.

Since signals of widely varying strength may be encountered, a sensitivity control must be provided. This is an adjustable shunt across the sensitive relay which reduces its sensitivity by a factor of 10, 100 or 1000 times. A jack is provided so that the station tuned in can be identified by listening with headphones.

In designing the r-f portion of the carrier-operated relay, a choice had to be made between sensitivity and selectivity. It was decided to design for maximum sensitivity, which implies loading the tuned circuit with the low (1,000-ohm) resistance of the meter-relay coil. This heavy loading reduces the overall Q of the two-circuit tuning system to about 7. Consequently the resonance curves (Fig. 5) display a width of about 100 kc at 6 db down at 650 kc, which is not sufficient to separate stations in the broadcast band. This is not a disadvantage if the control signal is substantially stronger than those of adjacent stations, which is ordinarily the case when the device is used as a carrier alarm.

For operation at a fixed frequency a larger number of tuned circuits might also be employed to increase the selectivity, although such a system would be difficult to track over an extended tuning range.

Construction and Circuit

The external appearance of the carrier-operated relay is shown in Fig. 1. The pointer of the meter-relay can be viewed as a tuning indicator. In operation, the sensitivity switch (bottom, left) is set to the minimum position and the tuning dial set to the frequency of the control signal. The sensitivity switch is then rotated clockwise until the meter pointer swings and the magnetic contact points (Fig. 2) close. If the reset switch (bottom, center) is in the upper posi-

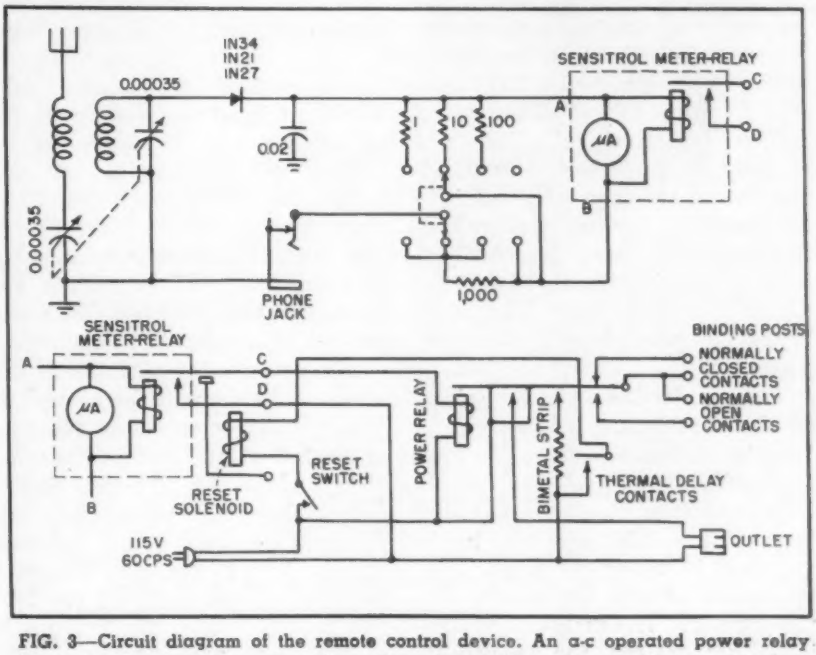


FIG. 3—Circuit diagram of the remote control device. An a-c operated power relay (at bottom), actuated by the micro-relay, closes when the signal is received or removed. A thermal contact resets the relay after 10 to 60 seconds

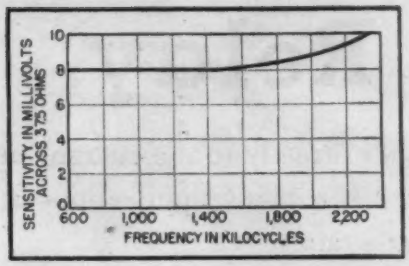


FIG. 4—Sensitivity curve, showing input r-f voltage required to actuate relay at various frequencies from 700 to 2,300 kc. Other frequency ranges can be accommodated with the same sensitivity

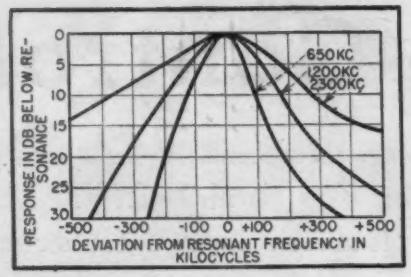


FIG. 5—Selectivity curves at various frequencies. The low selectivity is inherent in the small number of tuned circuits and the heavy loading. Higher selectivity can be obtained at some loss in sensitivity

tion, the relay remains closed indefinitely until the magnetic contacts are separated by manually operating the reset armature to the right of the relay. If the reset switch is in the down position, the thermal reset comes into play and the relay is reset automatically by the solenoid. The relay thereafter remains inactive until the carrier again appears. If the carrier is still present, the relay recloses cyclically until the carrier is removed.

The circuit diagram is shown in Fig. 3. The tuning system consists of two ganged 0.00035 μf capacitors, tuning the antenna circuit in series and the secondary in parallel. This provides a low-impedance input (about 50 ohms). The coils are universal-wound pies, each the secondary of a standard broadcast-band r-f antenna input transformer. The coupling between the two coils is varied until maximum sensitivity is achieved at both ends of the tuning range. The crystal rectifier is not critical. The 1N34 germanium diode serves well, although somewhat greater sensitivity (about 25 percent greater rectified current) may be obtained from selected silicon crystals of the 1N21 or 1N27 type. The 0.2 μf bypass to ground adds about 50 percent to the rectified current.

The sensitivity switch consists of three resistors switched by a two-circuit rotary switch. To keep the loading (and hence the selectivity) constant on all positions of the switch, a 1,000-ohm resistor is switched in series when the low-resistance shunts are across the micro-relay. If constant selectivity is not desired, the sensitivity may be reduced by switching series resistance, which will increase the selectivity as the sensitivity is lowered.

The switch employed should have positive action and should short over from one contact to the next as it is rotated so that the micro-relay is not left unshunted between the low-sensitivity positions. This precaution is necessary when strong r-f signals are encountered, since the 2-microampere coil of the micro-relay may be burned out by currents in excess of one milliampere.

Testing Sensitivity

Testing of the r-f and micro-relay portions of the device is readily carried out with a standard signal generator having a low-impedance output. The measured r-f input required to close the meter relay at various frequencies is shown in Fig. 4. A measurable movement of the meter pointer can be achieved with as little as 2 millivolts r-f input, and the sensitivity of closing can be increased by setting the coil-spring of the meter movement so that the pointer rests, with no current flowing, on the 1-microampere scale division. This adjustment is not recommended, however, since the reliability is somewhat impaired.

The power-relay portions of the device are shown at the bottom of Fig. 3. The micro-relay contacts (repeated at the left in the diagram) close the 115-volt circuit across the power relay coil (Advance type 304B with thermal reset). The power relay contacts thereupon close, applying the 115-volt circuit to an outlet receptacle at the right, and operating independent normally-open and normally-closed contacts connected to binding posts. When the power relay closes, another contact applies the 115-volt circuit to the thermal reset, a resistance coil wound on a bimetal strip. After a period of from 10 to 60 seconds (adjustable by the setting of the contact screw) the bimetal flexes, closing the 115-volt circuit to the micro-relay reset solenoid, which resets the micro-relay. A toggle switch in series with this circuit may be opened to disable the reset circuit.

When used as carrier-failure alarm, the polarity of the crystal rectifier is reversed, and the tension of the micro-relay coil spring is adjusted, by turning the screw on the front face of the relay mount, until the magnet contacts close solidly in the absence of the current. When no carrier is present, the relay then continues to close and reset in the normal fashion. When a carrier of sufficient strength is applied, the reverse current holds the micro-relay contacts open against the tension of the coil spring and the meter remains in the standby position so long as the carrier is present in sufficient strength. Since the time constant of the meter movement, under normal current of 2 to 10 microamperes, is of the order of a second, any interruption of the carrier of this or longer duration actuates the relay, followed by a reset at 10 to 60 seconds.

The sensitivity is remarkable when judged by the usual performance of crystal-detector receivers. The relay will close on signals which are just barely audible in high-sensitivity (Baldwin-type) headphones. It will close solidly on the signal from a 50-kilowatt broadcast station located at 7 miles, when a finger is touched to the antenna binding post. A long aerial (130-foot dipole) provides such a high level of signal that selectivity is the chief problem. The cost of the parts, at current prices, is approximately 50 dollars. Readers are invited to correspond concerning possible uses of the device and variations of the design to meet specific applications.

Adjustable-Bandwidth F-M Discriminator

Use of a cathode follower in place of the pentode amplifier normally used to drive a Foster-Seeley discriminator provides easily adjustable bandwidth control, output-frequency characteristics complementary to the frequency-voltage characteristics of stabilized f-m microwave oscillators, and freedom from critical adjustments

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IN THE COURSE OF research on communication systems, the need arose for a discriminator to be used in frequency-modulation receivers having variable bandwidth. Almost simultaneously the mathematical analysis of a typical discriminator circuit showed a possibility of constructing a discriminator whose bandwidth would be readily adjustable by means of a tap switch, but whose output—frequency characteristics would have a shape independent (in normalized coordinates) of discriminator bandwidth.

This discriminator circuit has proved considerably more tolerant to misadjustment than conventional types, and has four other significant features: (1) the bandwidth is adjustable over more than a four-to-one range by means of a single tap switch; (2) the normalized output voltage-frequency characteristic is independent of bandwidth; (3) the normalized output voltage-frequency characteristic may be made the complement of the frequency-modulating voltage characteristic of a frequency-modulated stabilized microwave oscillator, thereby making possible very good overall system linearity; (4) the circuit is not as critical to slight mistuning as most other discriminators (particularly in mistuning the primary).

The discriminator has three compensating undesirable features: (1) one additional high-g_m tube may be required as compared to other discriminators; (2) sensitivity (output volts per megacycle of frequency deviation) is an inverse-squared function of bandwidth while peak output voltage is an inverse function of bandwidth; (3) larger bandwidths are required for a given amount of distortion than in the case of the Foster-Seeley discriminator.

Electrical Circuit

The circuit of a typical cathode-driven discriminator, given in Fig. 1, operates in the same fashion as the Foster-Seeley[1] discriminator except for one important fact—the stage preceding the rectifier acts as a constant-voltage source (cathode follower) instead of a constant-current source (pentode amplifier). This in general means that the tube driving the discriminator may not

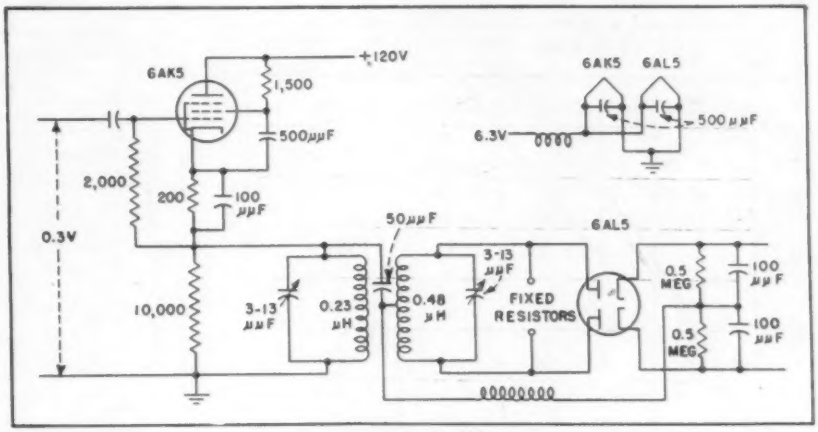

FIG. 1—Cathode-driven discriminator circuit, with values of components for center frequency of 60 mc

give good limiting action, so that the circuit may have to be preceded by a conventional limiter.

The coils forming the transformer in the cathode-driven discriminator may be coupled as tightly as possible, since the output of the discriminator is proportional to the coefficient of coupling between them and the bandwidth is independent of this coefficient. Circuit parameters given in Fig. 1 are for operation at a midfrequency of 60 mc, and all data given were taken about this midfrequency.

Theory of Operation

An analysis of the operation of a cathode-driven discriminator is partially given as an incidental part of a published analysis[2] of the Foster-Seeley discriminator, and the results of this analysis are of interest here. If the detectors are linear, the output voltage as a function of circuit parameters is

$$E_0 = \eta E_1 \left\{ \sqrt{\left[1 + \frac{\alpha Q_2 F}{2(1+Q_2^2F^2)}\right]^2 + \left[\frac{\alpha}{2(1+Q_2^2F^2)}\right]^2} - \sqrt{\left[1 - \frac{\alpha Q_2 F}{2(1+Q_2^2F^2)}\right]^2 + \left[\frac{\alpha}{2(1+Q_2^2F^2)}\right]^2} \right\} \quad (1)$$

where E_0 = d-c output voltage
η = rectification efficiency of rectifiers
E_1 = Peak r-f voltage across transformer primary
$\alpha = Q_2 K \sqrt{L_2/L_1}$
Q_2 = loaded Q of transformer secondary
K = coefficient of coupling
L_2 = inductance of transformer primary
L_1 = inductance of transformer secondary
$F = 2B/f_m$
B = deviation
f_m = midfrequency

Each radical in Eq. 1 represents the output of one of the two half-wave rectifiers. The expression may be simplified to

$$\frac{E_0}{\eta E_1} = \sqrt{1 + \alpha \frac{\frac{\alpha}{4} + Q_2 F}{1 + Q_2^2 F^2}} - \sqrt{1 + \alpha \frac{\frac{\alpha}{4} - Q_2 F}{1 + Q_2^2 F^2}} \quad (2)$$

Now if the detectors are square-law instead of linear, the effect on Eq. 2 is simply to remove each radical, leaving

$$\frac{E_0}{\eta E_1} = \frac{2\alpha Q_2 F}{1 + Q_2^2 F^2} \quad (3)$$

From this expression it is apparent that a plot of $E_0/\eta E_1 \alpha$ as a function of Q_2F or, since $F = 2B/f_m$, as a function of Q_2B, should be independent of Q_2. The effect of a change in Q_2 will be to produce a proportional change in the peak amplitude reached by E_0 since α depends directly on Q_2, and to change $\Delta\nu_p$, the peak-to-peak bandwidth of the discriminator, inversely with Q_2 since the output will be a maximum for a given value of Q_2B independent of Q_2. The curve given by Eq. 3 is identical with the voltage-frequency characteristic of microwave discriminators, and hence complementary to the frequency—voltage characteristic of a frequency-modulated microwave oscillator stabilized by such a discriminator[3]. As can be seen from a comparison of Eq. 2 and 3, the use of linear detectors instead of square-law makes the shape of the output-frequency characteristic somewhat dependent on the value of α, and hence on the Q of the tuned circuit. For small deviations (normalized with respect to peak deviation), or for coefficients of coupling less than unity, the difference between linear and square-law rectifiers is negligible.

Experimental Results

The results obtained from the circuit of Fig. 1 are shown graphically in Fig. 2 and 3. Figure 2 is a normalized plot of the output voltage—frequency characteristic of the cathode-driven discriminator for various loadings, in comparison with two curves taken at different loadings on a Foster-Seeley discriminator. The shape of the characteristic of the Foster-Seeley dis-

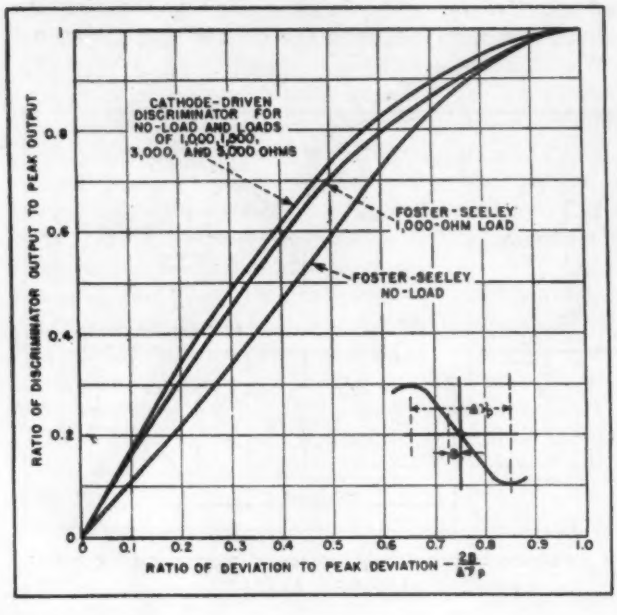

FIG. 2 (left)—Voltage-frequency characteristics of conventional and cathode-driven Foster-Seeley discriminators, showing that the cathode-driven discriminator is independent of loading

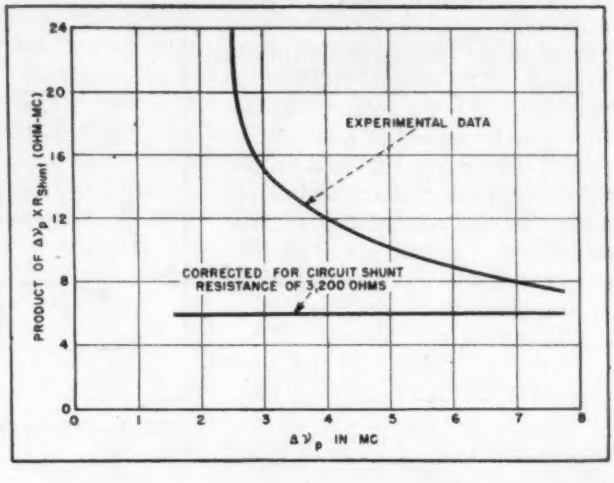

FIG. 3 (below)—Variation of bandwidth-load product with bandwidth for cathode-driven discriminator, showing how circuit losses provide lower limit to bandwidth obtained

criminator is a critical function of loading and coupling, optimum damping being somewhere between the no-load condition (circuit losses correspond to 3,200-ohm shunt resistor) and loaded condition (equivalent shunt resistance of 760 ohms) in the case studied. The normalized curves for the cathode-driven discriminator, however, are independent of loading within experimental error, as shown in Fig. 2.

Figure 3 is a plot of the product of the bandwidth and load resistance as a function of the bandwidth. (Table I is included for supplementary data not included in Fig. 3.) If the circuit Q were infinite, so that all losses were supplied by the loading resistor, this curve should be a straight line parallel to the abscissa since in this case $Q_L = \omega LR$. Therefore $R\Delta\upsilon_p$, where $\Delta\upsilon_p$ is peak-to-peak bandwidth, should be independent of R or $\Delta\upsilon_p$.

In practice, circuit losses provide a lower limit to the obtainable bandwidth. In the case investigated this came at 1.7 mc, equivalent to a shunting resistance of 3,200 ohms. This resistance was assumed to be included in parallel with the various load resistors.

The resistance of the parallel combination was computed for each case and a new curve plotted using the calculated resistance as the load. As is seen in Fig. 3, the resulting curve is within experimental error a straight line parallel to the abscissa. The load resistors used were measured at d-c rather than 60 mc, so that the a-c resistance of the highest-resistance units might be expected to be relatively lower than the a-c resistance of the lower units because of the Boella effect. This might explain the dropping off of the experimental points for small bandwidths.

Adjustment of the cathode-driven discriminator to give symmetrical curves was considerably easier than adjustment of the Foster-Seeley discriminator in all cases. This was especially true with respect to primary tuning in obtaining a symmetrical curve about f_m.

Distortion

Inasmuch as the cathode-driven discriminator is easy to adjust and not critical as to coupling coefficient, one might wish to use it in the detection of frequency-modulated signals from some linearly modulated source such as a broadcast station. In this application one is interested in the nonlinearity of the circuit, which will be somewhat greater than that of the Foster-Seeley discriminator.

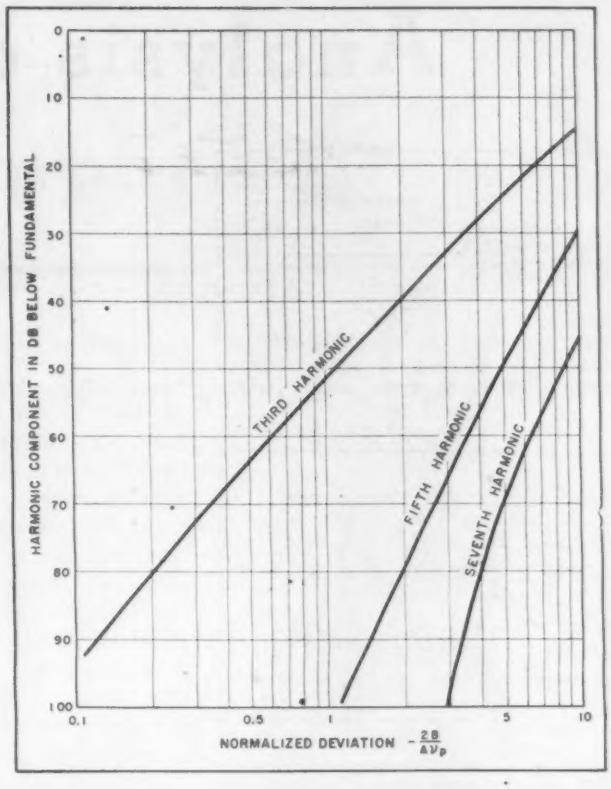

FIG. 4 — Distortion curves for cathode-driven discriminator, giving amplitudes of important harmonic components as function of ratio of frequency deviation to peak-to-peak bandwidth of discriminator

Figure 4 shows the amplitude of the various important harmonic components in the output of the cathode-driven discriminator as a function of the ratio of deviation of incoming frequency to peak-to-peak bandwidth of the discriminator. These curves are an extension of the ones given previously in an analysis of microwave discriminators,[3] the extension having been carried out by twelve-point schedule analysis. As is shown by these curves, the peak-to-peak discriminator bandwidth must be five times the peak-to-peak deviation for 1 percent third harmonic distortion or 2.5 times the peak-to-peak deviation for 5 percent third harmonic distortion. These figures compare with the 1.7 ratio ordinarily used in commercial frequency-modulation receivers. The curves given apply for square-law detectors. If linear detectors are used, the ratio of 2.5 may be dropped to 2.1 for critical coupling. Distortion is always less with linear than with square-law rectifiers in this discriminator, but is somewhat dependent on coupling at high deviations.

TABLE I—Effect of Load Resistance on Bandwith

Physical Load Resistor (ohms)	Actual Shunt Resistance* (ohms)	$\Delta\nu_p$ (mc)
∞	3,200	1.7
87,000	3,090	2.0
12,000	2,530	2.5
5,000	1,950	3.0
3,000	1,550	4.0
1,500	1,020	6.0
1,000	760	7.6

* Corrected for 3,200 ohms circuit resistance.

This paper is based in whole or in part on the work done for the Office of Scientific Research and Development under Contract OEMsr-262 with the Massachusettts Institute of Technology.

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Rectifier and Capacitive-Input Filter

Some simplifying assumptions make it possible to express results in terms of two general parameters. Data are presented for tube angles, output voltage, ripple, and peak tube currents. Filter resonance, pulsed current, and continuous current operation are included

The other special case might be called the non-cutoff or the continuous-current case in that each tube is conducting 180 degrees and the current entering the filter never cuts off or drops to zero. The current in each tube is stopped by conduction starting in the other tube, and a form of commutation between the tubes is set up in which the current switches from one tube to the other and back again.

The main problem for this second special case is not that of finding γ which is known to be 180 degrees, but that of finding the boundary or dividing line between the cut-off or pulse-current and non-cut-off or continuous-current cases. In the analysis it was found that the boundary occurred at approximately $\omega CR = 0.6$ with the continuous-current case occurring for smaller values of ωCR and the pulsed-current case for larger values of ωCR.

Characteristics

The characteristics of the circuit that are of the most use in the design of the circuit and predetermining its properties are the tube angles, the average output voltage, the ripple, the peak and average tube currents, and the peak inverse voltage on the tubes.

The angle γ during which the tube conducts is presented in Fig. 3 as a function of ω^2LC for various values of ωCR. For a fixed value of ω^2LC this angle varies from zero degrees at open circuit (ωCR infinitely large) to 180 degrees for the continuous-current case which includes values of ωCR from approximately 0.6 to zero (short circuit). With ωCR fixed, the angle γ varies little for values of ω^2LC above ten, but for ω^2LC less than ten the variation is considerable, especially as filter resonance is approached ($\omega^2LC = 0.5$).

The angle β at which the tube stops conducting is presented in Fig. 4 and the manner of variation is similar to that for angle γ. For a fixed value of ω^2LC, angle β varies from ninety degrees at open-circuit to 180 degrees for values of ωCR from approximately 0.6 to zero (short circuit). The angle α at which the tube starts conducting may be found by the use of the relation, $\alpha = \beta - \gamma$. Increasing the load of the rectifier circuit has the effect of increasing both angles β and γ, while α is decreased. The effect of approaching resonance in the filter is similar in that angles β and γ are increased, while α is decreased.

One of the most important characteristics of a rectifier circuit is the average output voltage. This voltage may be obtained from the ratio of the average output voltage to the maximum value of the sinusoidal voltage across one-half of the transformer secondary, E_{dc}/E_m. The ratio E_{dc}/E_m is given in Fig. 5 as a function of ω^2LC for various values of ωCR. For a fixed value of ω^2LC, the ratio decreased from a maximum of unity for ωCR infinitely large (open circuit) to 0.637 for ωCR equal approximately to 0.6. For values of ωCR from 0.6 to zero (short circuit) the continuous-current case applies, and $E_{dc}/E_m = 0.637$. The voltage ratio E_{dc}/E_m varies little for large values of ω^2LC, but changes considerably as ω^2LC approaches the resonance value of 0.5. Both increasing the load and approaching filter resonance have a similar effect, that of making the voltage ratio approach 0.637.

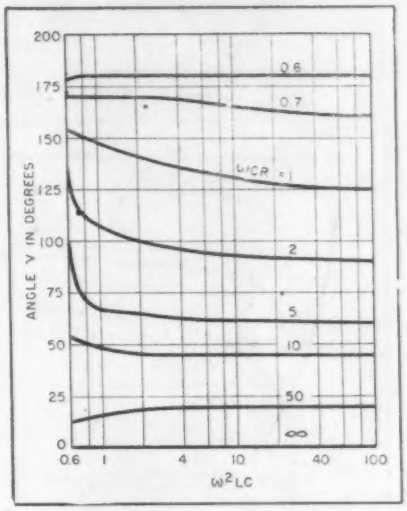

FIG. 3—Angle during which the tube conducts

The ripple voltage across the load resistor of the rectifier circuit is another important characteristic. It may be obtained from the percent ripple r which is defined as 100 times the ratio of the effective value of the fundamental ripple voltage appearing across the load resistor to the average voltage E_{dc} across the same resistor. In practice only the fundamental ripple voltage is sufficiently large to warrant consideration, and for a full-wave circuit, the fundamental ripple frequency is twice the supply frequency.

For large values of ω^2LC and ωCR it was found possible to give a partly empirical and partly theoretical expression for the percent

Analysis of Full-Wave

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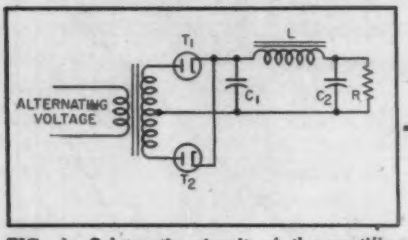

FIG. 1—Schematic circuit of the rectifier and filter

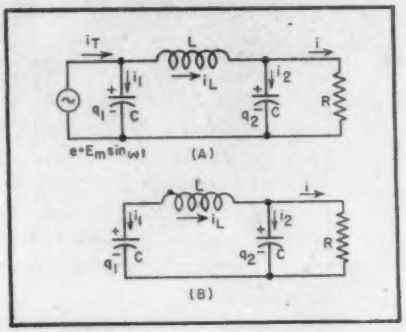

FIG. 2—Equivalent circuits of the rectifier and filter. (A) tube conducting; (B) tube not conducting

THE FULL-WAVE rectifier circuit with a capacitive-input filter is probably the most common of all rectifier circuits.

Freeman and Terman[1] have made an analysis of this circuit using as their chief assumption the idea that the filter inductance was infinitely large. They have presented curves which could be used in the design of the rectifier and filter circuit, but unfortunately the curves are valid only at lighter loads and at one frequency, that of sixty cycles per second. Stout[2] made an analysis which presented the waveforms in the circuit but gave no characteristic curves suitable for use in design, while Mitchell[3] assumed an infinitely large inductance in his analysis.

This paper presents the results of an investigation made to determine the effects of a finite filter inductance on the operation of the rectifier and filter circuit. The results may be used in the design of the rectifier and filter circuit for any frequency and with any load from open circuit to short circuit.

The analysis of the circuit was made using the newly developed steady-state operational calculus[4]. It was found that approximate values of the portion of a cycle that each tube was conducting would be helpful in the analysis, and these were obtained by setting up the circuit and observing the tube currents by means of a cathode-ray oscilloscope. With these approximate values as a starting point, the exact portion of each cycle that the tube was conducting could be determined from the analysis.

The schematic diagram of the rectifier and filter circuit is shown in Fig. 1. Tube T_1 conducts during the first half cycle of the applied alternating voltage and charges capacitor C_1 almost to the peak value of the alternating voltage.

Current flows from capacitor C_1 through the filter inductor L to charge capacitor C_2. The load resistor R has the same voltage as the second capacitor C_2. During the second half cycle of the alternating voltage, the second tube T_2 conducts, and capacitor C_1 is charged again to repeat the process. The filter may be thought of as a low-pass filter which passes direct current but greatly attenuates the alternating currents.

Assumptions

To make a mathematical analysis feasible, certain simplifying assumptions are necessary:

(1) The transformer and filter-inductor resistance is zero. The tube resistance is zero when it is conducting.

(2) The leakage reactance of the transformer is zero.

(3) Both filter capacitors have the same capacitance.

The errors caused by the first of these assumptions will be corrected for in a manner to be described.

Two equivalent circuits were used as shown in Fig. 2. When the tube is conducting Fig. 2A applies, while Fig. 2B should be used when the tube is not conducting. The equations for the currents and voltages in the circuits were set up and then solved by the use of the steady-state operational calculus. The solution showed that the tube conduction angle γ of one tube depends only on the two independent variables ω^2LC and ωCR where ω is 2π times the supply frequency, L is the inductance in henries of the filter inductor, C is the capacitance in farads of one filter capacitor, and R is the resistance in ohms of the load resistor. The variable ω^2LC may also be regarded as the ratio of the reactance of the filter inductor to the reactance of one filter capacitor, while ωCR is the ratio of the load resistance to the reactance of one filter capacitor. The solution showed furthermore that all characteristics of this rectifier circuit depend on these two variables alone.

Filter resonance seems to occur at $\omega^2LC = 0.5$, for then the sum of the reactances of the two filter capacitors and of the inductor is zero. This assumes that the reactances are calculated at the fundamental ripple frequency which is twice the supply frequency. The problems of practical importance at present always have ω^2LC greater than 0.5 and for this reason it was decided to use 0.6 as the lowest value of ω^2LC.

Two special cases of the general solution were considered separately because of the significance of the results. The first of these is that for ω^2LC infinitely large which implies that the filter inductance is infinitely large and that the current through the inductor has zero ripple.

The circuit was set up to obtain experimental verification of the foregoing calculated results. Measurements were made with a sixty-cycle source of alternating voltage and with a filter having $\omega^2LC = 2.0$. This value of ω^2LC was chosen so that a large range of ωCR could be obtained. Filters with other values of ω^2LC were also used with good agreement.

Experimental Results

Various quantities such as average output voltage, ripple voltage, and the tube angle γ were measured. Figure 7 shows the calculated and experimental results for the average output voltage ratio with the calculated results shown as a solid curve and the experimental points as circles. Agreement is good except for low values of ωCR which correspond to heavy loads.

For heavy loads the resistance of the rectifier tube, transformer, and filter choke become appreciable in comparison to the load resistance. These resistances were measured and were added to the load resistance to give an equivalent load resistance. The tube resistance was taken as the voltage across the tube at rated current divided by the rated load current of the tube. The average output voltage is then the product of the average output current and the equivalent load resistance.

The new value of ωCR is increased since the equivalent load resistance is larger than the load resistance alone. The resulting compensated experimental values are shown as crosses in Fig. 7, and the agreement between the experimental points and calculated curve is quite good even for heavy loads (small values of ωCR).

Similar results for the percent ripple are shown in Fig. 8. Since the percent ripple is inversely proportional to the average output voltage, the compensated values of percent ripple are smaller than the uncompensated ones. Again agreement between the calculated curve and the compensated experimental points is good.

The peak tube currents will be somewhat in error for high values of ωCR even though compensation is used. Equation 3 will give somewhat high values of peak currents in this case.

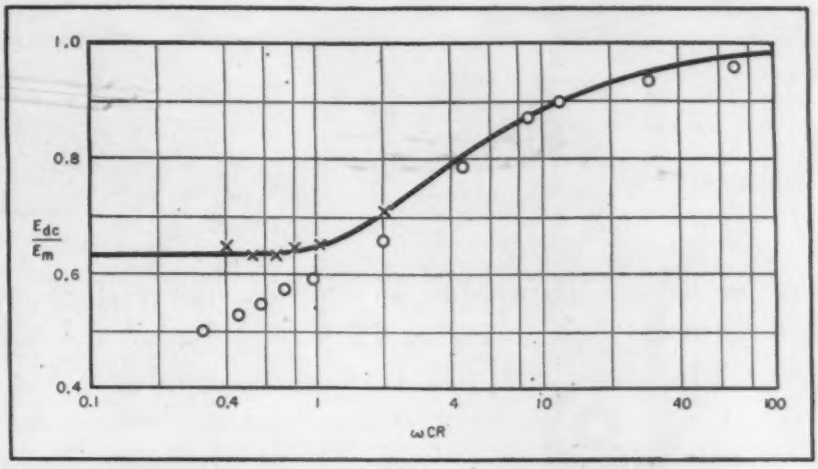

FIG. 7—Average output voltage for $\omega^2LC = 2.0$. The solid curve is calculated, the circles are uncompensated experimental points, and the crosses are compensated experimental points

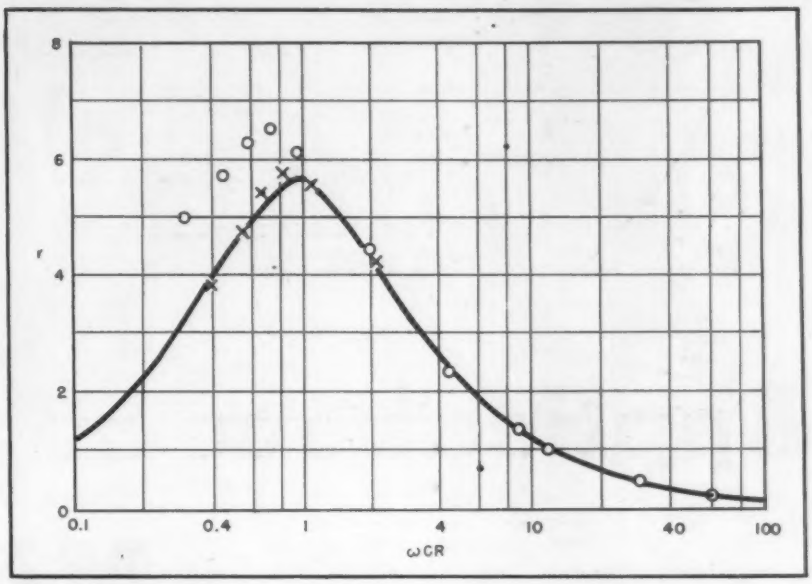

FIG. 8—Percent ripple voltage for $\omega^2LC = 2.0$. The other conditions are the same as illustrated in the graph of Fig. 7

One very common design problem is that in which the maximum ripple, the load current and voltage, and input alternating voltage and frequency are known and the filter components, tube and transformer are to be selected. If a size of filter capacitor is selected, ωCR may be calculated. From the ripple characteristics of Fig. 6 the value of ω^2LC may then be selected to give the maximum permissible ripple. The inductance necessary may then be obtained from ω^2LC. If an inductance of this size is not available, a different value of capacitance C should be selected and the whole procedure repeated.

Once the values of ωCR and ω^2LC have been obtained the secondary voltage of the transformer may be found from the average voltage characteristics of Fig. 5. The tube or tubes may be selected by the use of Eq. 3. If the tube, transformer, and choke resistances are appreciable, they should be taken into account.

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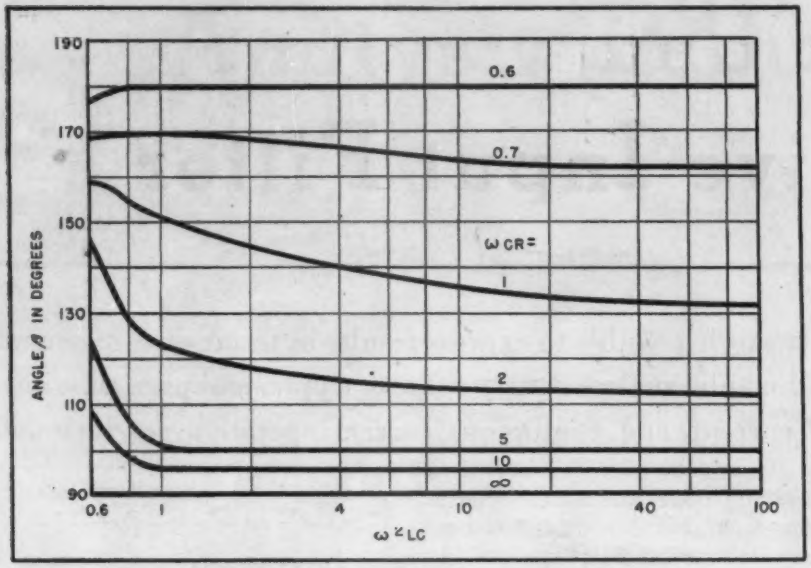


FIG. 4—Angle at which the tube stops conducting

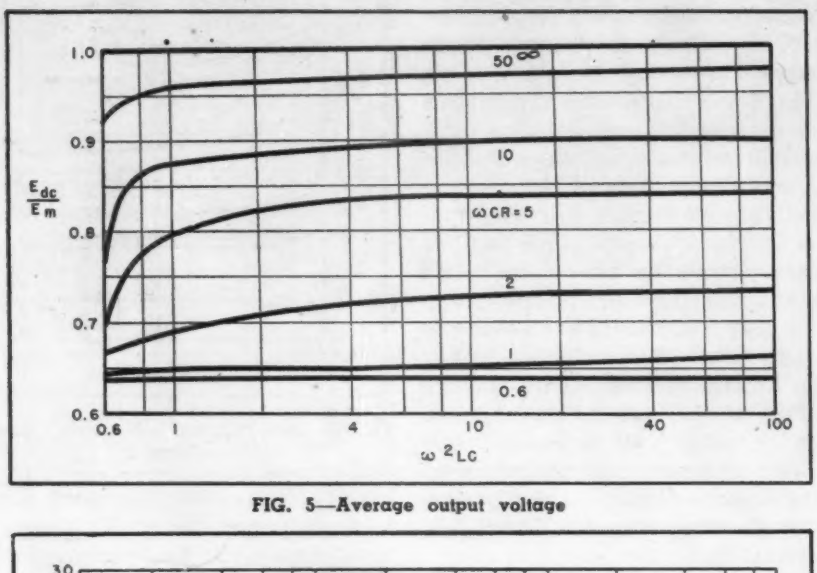


FIG. 5—Average output voltage

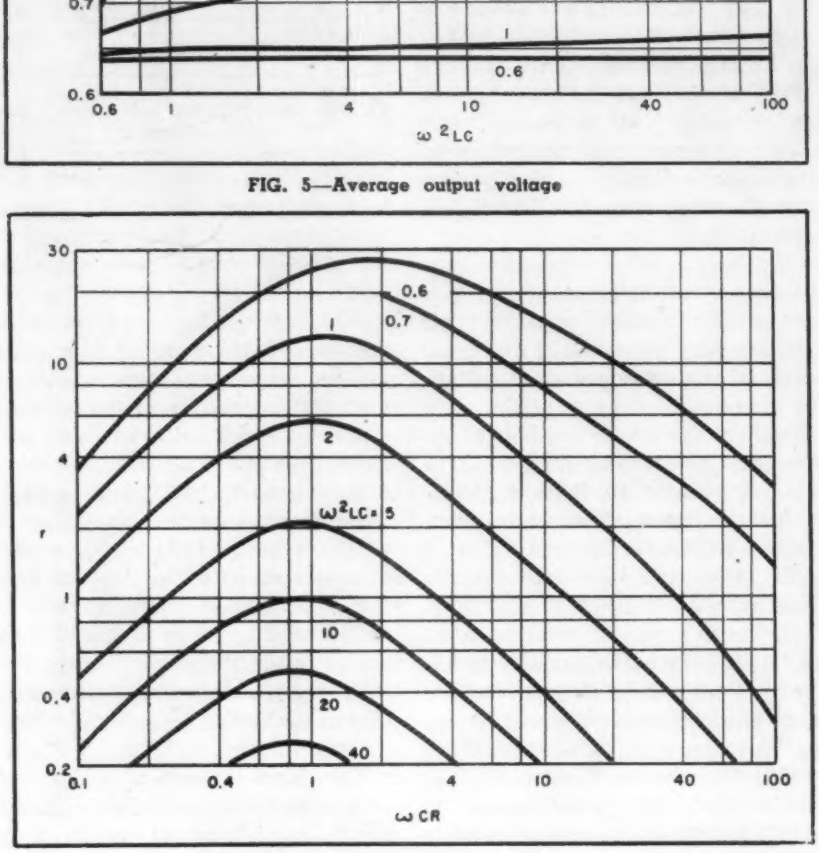


FIG. 6—Percent of ripple voltage

ripple across the load resistor. This is

$$r = \frac{31.5 + 67.25\ \omega^2LC}{(\omega^2LC)\ \sqrt{4\ (\omega^2LC)^2 + (\omega CR)^2\ (4\ \omega^2LC - 1)^2}} \quad (1)$$

and the error in the percent ripple will be less than five percent if $\omega^2LC \geqq 2.0$ and $\omega CR \geqq 5.0$.

Equation 1 will be applicable in many of the problems encountered in practice, particularly those in which the circuit is lightly loaded. It expresses the percent ripple for the region where each tube is firing only a short time, the output voltage is high, and the charges on the filter capacitors do not vary much during one cycle.

For heavier loads, the percent ripple is shown as curves in Fig. 6. For ω^2LC constant, the ripple increases from zero for ωCR approaching infinity (open circuit) to a maximum in the vicinity of ωCR equal to unity and then approaches zero again as ωCR approaches zero (short circuit). For a fixed value of ωCR, the percent ripple increases from zero for ω^2LC approaching infinity (infinitely large inductance) toward a maximum as ω^2LC approaches 0.5 (filter resonance). For very heavy loads ($\omega CR < 0.6$) the continuous-current case applies and the percent ripple r is given by

$$r = \frac{47.14\ (\omega\ CR)}{\sqrt{4\ (\omega^2LC)^2 + (\omega CR)^2\ (4\ \omega^2LC - 1)^2}} \quad (2)$$

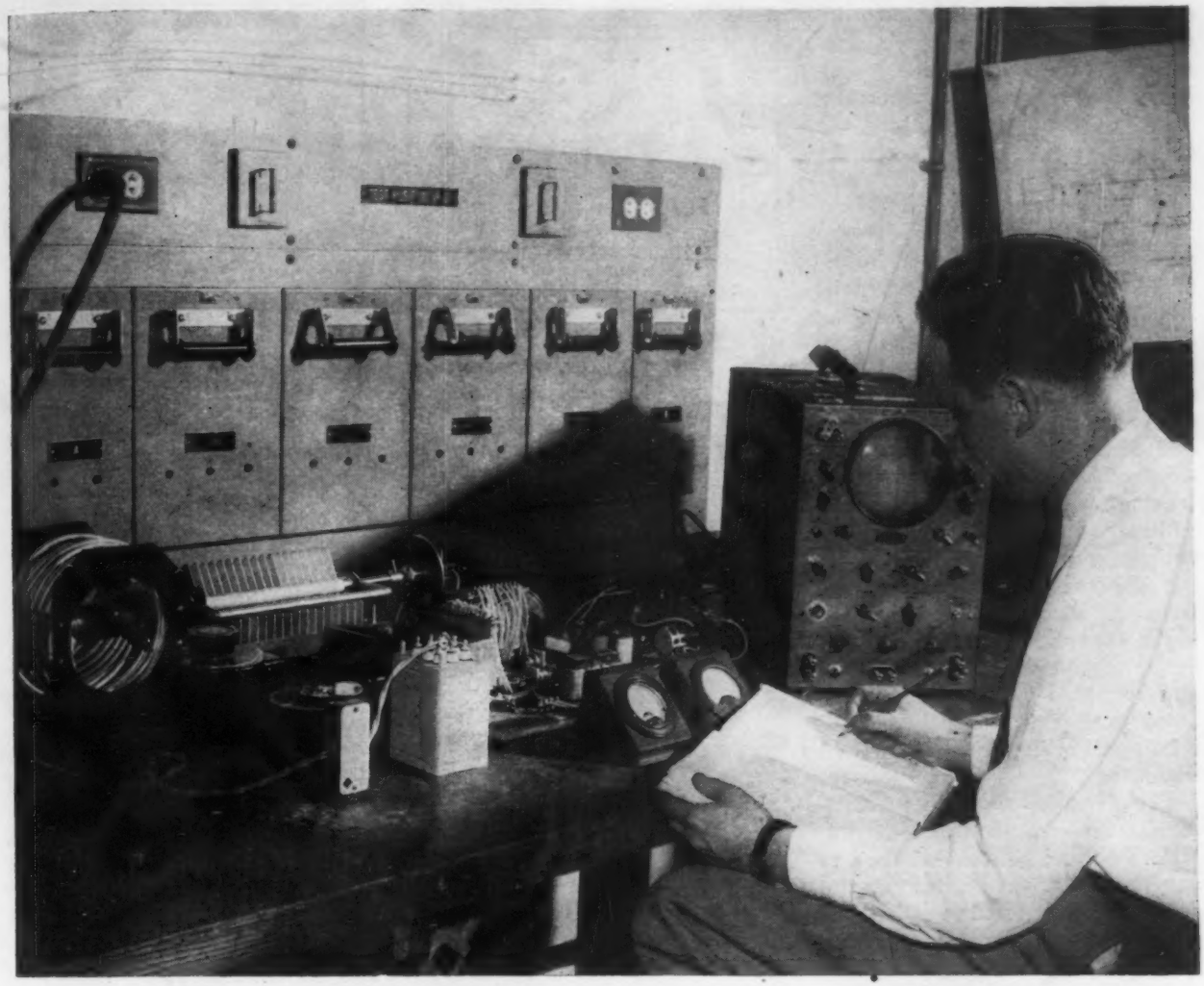
The percent ripple approaches zero as ωCR approaches zero (short circuit) because most of the ripple voltage appears across the inductor and very little of it appears across the output load resistor, while all of the average voltage appears across the load resistor.

The peak tube current may be obtained from the ratio P_T of the peak tube current to the average current for one tube. An empirical equation was fitted to the calculated values of P_T and this equation is

$$P_T = 5.016\ (\omega CR)^{0.5025} \quad (3)$$

where $\omega CR \geqq 0.6$. For values of ωCR from 0.6 to zero, P_T varies from four to a minimum of two.

The inverse peak voltage applied to each tube is always the maximum value of the transformer secondary voltage.



Measurements on modern magnetic amplifiers show that they have fast response and high gain

AMPLIFIERS

gotten and "amplifier" became synonymous with "vacuum tube amplifier". However, European scientists pursued its development for power applications.

With the ascendancy of the Nazi regime in Germany, scientific research became highly important for development of their war machine. The Germans investigated everything that could be of military use, including the magnetic amplifier. American engineers had failed to parallel development of the related components with improvements in the magnetic amplifier. The Germans, however, took the magnetic amplifier, added a more recent noteworthy U. S. development, Permalloy, a magnetic material having high-permeability characteristics, and also took advantage of the dry disk rectifier (copper oxide, copper sulfide, and especially selenium). The excellence of the German selenium rectifier was a great aid in their development of the magnetic amplifier.

Though Germany has been quite successful in the development of this device for military use, greater success was inhibited by lack of proper coordination between their Army, Navy, and Air Force. Each service conducted its individual program, disseminating a minimum of information to other services.

The German Navy employed magnetic amplifiers wherever possible on their naval vessels, some having been installed a decade before the war; they operated satisfactorily and without attention for the life of the vessel. They were used for stabilizing range finder mounts, gun mounts, and as preamplifiers to servosystems used to stabilize antiaircraft directors in space. The

Current in auxiliary winding on saturable core reactor controls current in main winding. High amplification and fast response are obtained. These magnetic amplifiers have been used in gun control servomechanisims, braking systems of electrical locomotives, and in automatic aircraft pilots. Research has been conducted toward their use in the steering control of V-2 rockets

Magnetic amplifiers, essentially saturable core reactors in which the degree of magnetization and flux change is controlled by auxiliary windings, were invented by E. F. W. Alexanderson for use with his alternator. However they had not been widely applied because of limitations of core materials and lack of a high-current rectifier. High permeability core materials have been alloyed in Japan, efficient selenium rectifiers have been developed in Germany and the theoretical analysis in Scandinavia so that this American device is now practical. The nonelectronic amplifier, being free from mechanical limitations of vacuum tubes, may well provide the means for wider industrial application of the techniques built around electronic amplifiers. Development of magnetic amplifiers dramatizes the international aspects and interrelations of basic sciences of modern technology, and also reminds us occasionally to review discarded techniques in the light of current materials to determine whether their shortcomings have been negated by subsequent developments.—Editors.

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Applications of MAGNETIC

INTERROGATION of German prisoners revealed that the German Navy had begun replacing some electronic servoamplifiers, which were part of their naval fire control equipment, with servoamplifiers that employed magnetic amplifiers not liable to damage by shock or vibration. Although antedating vacuum-tube amplifiers, magnetic amplifiers had been little used in America. This information of German use indicated however that these devices could compete with vacuum tubes in some applications, just as other nonlinear devices have been used in place of vacuum tubes for modulation and detection.

The magnetic amplifier was first disclosed in a United States Patent granted to E. F. W. Alexanderson of the General Electric Co. in 1916. Over a period of years this patent was followed by 75 to 80 others. The purpose of Dr. Alexanderson's development was to provide means for modulating high-frequency alternating currents from one of his alternators so that its current could be used for trans-Atlantic radio telephony. It is interesting to note that by 1916 the magnetic amplifier had been used to modulate as much as 72 kw while Dr. DeForest's three element vacuum tube had been developed to handle only a few watts.

The magnetic amplifier became of minor domestic importance because of growing popularity and technical advances in vacuum-tube techniques. Development of high-permeability magnetic core materials, and efficient dry disk rectifiers, both necessary to the development of magnetic amplifiers of high efficiency, came later. By the time these components were available, the magnetic amplifier had been for-

Germans were also developing all-electric computers in which magnetic amplifiers were used in a manner analogous to the way we employ cathode follower vacuum tube circuits.

As a control device, the Luftwaffe used it in connection with automatic pilots. The German Army also conducted extensive research at Peenemunde in applying the magnetic amplifier to the V-2 as a steering control. It was also developed as a frequency regulator for motor-generator sets capable of maintaining accuracy to within 1/1,000 of the desired frequency while using a d-c motor whose voltage supply was not constant. Commercial use appears to have begun about 1942 or 1943 with application to electrical braking systems for railroad locomotives. At Darmstadt, research had begun in applying the magnetic amplifier to high-voltage power lines; however, this was interrupted by allied bombings and never developed past the fundamental research stage.

The Japanese conducted fundamental research on magnetic amplifiers but did not develop them to any great extent. Fire control equipment, purchased from the Germans, included magnetic amplifiers, but it never reached Japan because of the allied shipping blockade. The Japanese, however, developed magnetic core materials, which in the future will play an extremely important role in magnetic amplifier progress.

Other Foreign Contributions

Although Japan has never been an outstanding nation in the development of electronic equipment, there is one important exception, namely, in the field of magnetism: in this Japan led the world. In most countries, research in magnetism has always been a somewhat secondary consideration in electronic development. This was not true, however, in Japan. For over 30 years there has been maintained at Sendai, a government sponsored Research Institute for Iron and Steel and Other Metals, under Tohuku Imperial University. As a result of the work of this and other magnetic laboratories, Japan is responsible for at least half of the new magnetic materials introduced since 1920. The most important of these are Alfer, Sendust, and special silicon-iron alloys having permeabilities nearly twice that of United States' best magnetic material. These are vital to efficient magnetic amplifier design, for it is the property of extremely high permeability at low values of magnetomotive force that allows the magnetic amplifier to achieve high amplification. Of equal, if not greater importance is the fact that these alloys contain no nickel, an element in which Japan, Germany, and the United States are strategically deficient.

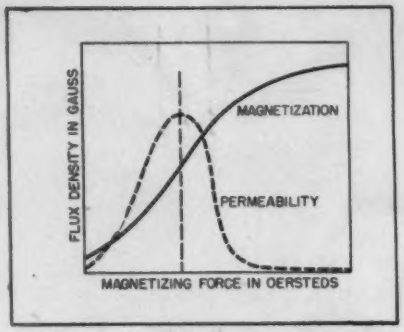

FIG. 1—Typical magnetic characteristics

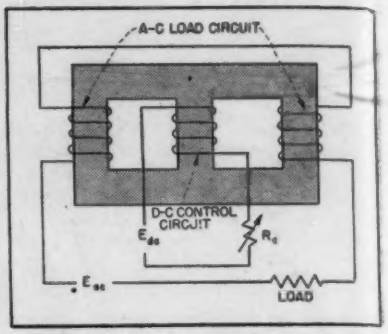

FIG. 2—Simplest amplifier connection

Swedish engineers are also interested in the magnetic amplifier. Dr. Uno Lamm, wrote a book in 1943 on the subject of magnetic amplifiers entitled "The Transductor." This work gives perhaps the most complete technical and theoretical treatise on the magnetic amplifier which has been published to date. The book discusses use of the magnetic amplifier for automatic regulating and relaying purposes.

The British are also conducting research on the magnetic amplifier. Reports show that they have been able to overcome to a certain extent a major fault of this device—the slowness of response to control signals.

Basis of Operation

Before discussing some of the mechanics of the saturable core reactor and the complete magnetic amplifier, it is advisable to review briefly the properties of iron-cored coil.

Any coil possesses the property of electrical inertia, called inductance. This inductance opposes a change in the current in the coil. The magnitude of this property is expressed in units called henrys. In an iron-cored coil, the inductance is dependent to a very great extent on the permeability of the iron in the core. Permeability is, in effect, a measure of the relative ease with which magnetism flows in a certain circuit.

The relationship between inductance and the permeability of the core is

$$L = \frac{1.256\ N^2 \mu\ a_i}{l_i} \times 10^{-8}\ \text{(henrys)}$$

where N is number of turns in the coil, μ is permeability, a_i is cross sectional area of the core in centimeters, and l_i is length of the magnetic path in centimeters. These constants are those used in the cgs system.

Because the permeability is not a constant, the inductance will vary with it. Therefore one must consider the permeability next.

Permeability of magnetic material is defined as the ratio B/H, where B is flux density (lines per square centimeter in the cgs system) and H is the magnetizing force (in oersteds, or in ampere turns since one ampere turn per centimeter is 1.256 oersteds). With these units, the permeability of air is unity.

Permeability, depending on flux density, is in general, small at low flux densities, maximum at moderate values and quite small when the core is saturated, as shown in Fig. 1.

A circuit containing inductance will, when subjected to an alternating voltage, have a current whose value is determined by $I_{ac} = E_{ac}/X_L$ where X_L is inductive reactance, which in turn is related to the inductance of the circuit by $X_L = 2\pi fL$ where f is frequency of the alternating voltage. So, as X_L increases, I_{ac} decreases and vice versa.

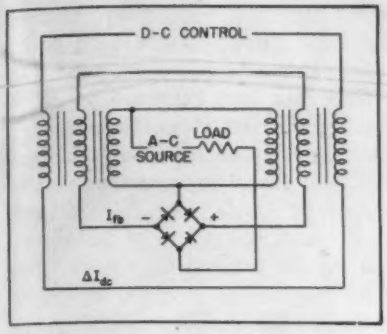

FIG. 3—Self-excitation can be used

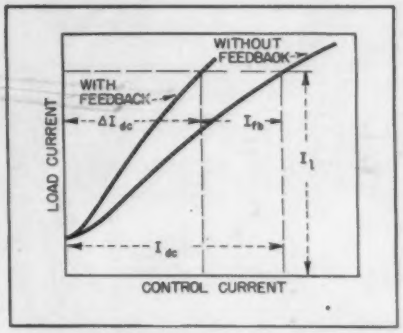

FIG. 4—Feedback increases sensitivity

For the purposes of this discussion, saturable core reactors are considered to be the actual transformer-like structures that are part of the complete magnetic amplifier circuit. A saturable core reactor consists of separate load and control windings with, perhaps in specialized cases, additional windings for feedback and stabilization circuits. Uses of these windings will be developed in the following discussion.

Because a given amount of direct current will control a larger amount of alternating current, the device is an amplifier and for many purposes can be considered analogous to a vacuum tube. A reactor used in this manner, together with selenium or dry rectifiers to provide the direct current, is called a transductor in Sweden.

Figure 2 shows a simplified circuit of a typical saturable core reactor. The reactor is designed so that with little or no current flowing in the d-c control winding, the reactance of the load windings is high and effectively limits the a-c to a small value. With a relatively large d-c flowing, the core becomes saturated, the permeability decreases, inductance and consequently reactance decreases and the a-c increases, allowing more a-c power to be dissipated in the load.

The amplification factor or more specifically in this case—the control ratio, can be defined as the ratio of ampere turns in the load windings to ampere turns in the control winding.

$$R_{\text{control ratio}} = \frac{NI_{\text{load}}}{NI_{\text{control}}}$$

Where NI_{load} is the number of turns of wire in the load windings multiplied by the current flowing through them; the same is true for NI_{control}.

Dr. D. W. Ver Planck of Carnegie Institute of Technology, working under sponsorship of the Office of Naval Research, has found that certain core and winding configurations have advantages over others as far as control ratio, speed of response, and insulation requirements are concerned.

The two reactors mentioned so far have been series wound. Of course it is possible and often desirable to use parallel wound reactors. It is known that as far as steady state performance is concerned, the series and parallel connections are of equal merit provided the impedance of the load is relatively small, as is usually the case.

Using Feedback

A German scientist, Dr. Theodor Buchhold, doing basic research on saturable core reactors and their applications in magnetic amplifiers, has demonstrated that addition of a certain amount of feedback current in the amplifier circuit reduces the necessary control power. A typical circuit utilizing feedback is as shown in Fig. 3; each reactor has primary, secondary, and tertiary windings.

Action of the feedback current can be shown with reference to the curves of Fig. 4. The one curve shows variation of load current as a function of the control current in a saturable reactor circuit such as that of Fig. 2. If a particular amount of feedback is added by a circuit such as that of Fig. 3, the load current varies more for a given change in control current than without feedback, as shown by the other curve. The reactor still requires the same total direct-current control as without feedback, however a large portion of the control current is supplied by the feedback winding; the control circuit need supply only ΔI_{dc}. Unless precautions are taken in feedback design, there is danger of instability of the system with resultant large currents.

Figure 5 shows a circuit used for frequency regulation of motor-generators by means of feedback through a magnetic amplifier. Normally changes in the d-c supply to the motor would cause changes in the speed and hence the frequency of the alternator. With the control circuit, however, if the frequency of the alternator changes from the intended value of 400 cps, increased current will flow through one or the other resonant circuit producing more current in one of the control windings of the saturable core reactor. In the case of falling frequency of the alternator, more current will flow through L_1. The magnetism so caused opposes that produced by the rectified alternator current that flows in L_2 and is also used to excite the motor. The decreased magnetism resulting from this opposition results in increased reactance of L_2, so that the current in the motor field is reduced, the motor accelerates, and the alternator frequency is restored to normal, with both resonant circuits drawing equal currents. By this system the alternator frequency can be controlled to one part in 1,000.

Speed of Response

In all of the presently known magnetic amplifier circuits there is one common fault: the slowness of response to control, which is inherent in any circuit containing a large amount of inductance. In most commercial applications this sluggishness of response is not an important drawback, but in certain military and commercial applications which are foreseen, a very short response time is necessary.

In preliminary laboratory work in this country a response time of approximately 6 cycles of the 60 cycle a-c supply has been obtained. The German naval magnetic amplifiers on the other hand had a response time of over 16 cycles of the supply frequency, which was, however, 333 cycles per second. The actual power gain of these German units was rather low, about 100. It

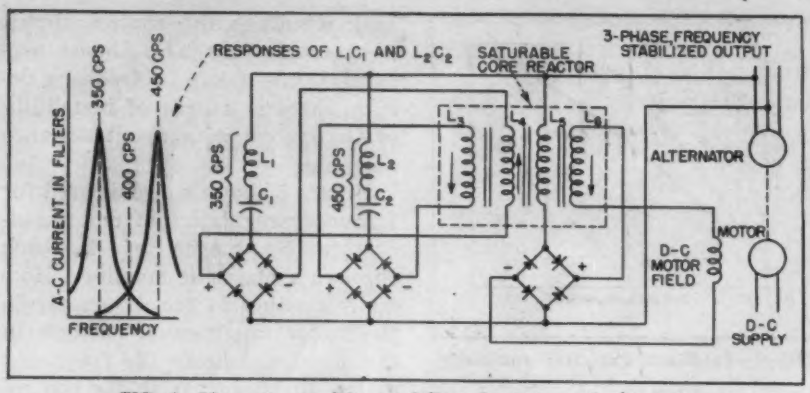

FIG. 5—Magnetic amplifier stabilizes frequency of alternator

would be desirable for naval applications, to have a response time of 1 cycle or less at 60 cycles supply frequency.

An obvious way to circumvent this problem and reduce the response time would be to increase the supply frequency. In Alexanderson's work, where he was using a supply in the neighborhood of 25 to 30 kc, response time was not a problem and his amplifier (or modulator in his case) was able to follow variations in audio-frequency control currents.

Changing the supply frequency of large power installations to any great extent seems to be impractical at this time. Therefore, it is important to approach the problem from the standpoint of the magnetic amplifier itself. The British have achieved some success in the solution of the response time problem by using a feedback or phase advance circuit, the nomenclature depending on one's point of view.

Future Potentialities

In conclusion it may be said that the magnetic amplifier has been up to now a relatively forgotten American invention which was taken by German engineers and given the prominence necessary to assure its proper development and application. Successful use of the magnetic amplifier throughout the German Navy for ten years has demonstrated beyond doubt its efficiency and practicability in use where equipment is subject to shock or vibration. As a matter of fact, this ruggedness was demonstrated by an instance when a mine, utilizing a magnetic amplifier, was dropped from a considerable height by an aircraft. It struck the ground instead of the water, was practically demolished, and scattered over the terrain. The magnetic amplifier portion, however, when recovered and tested, functioned normally.

Unlike the vacuum tube whose impedance is ohmic and therefore introduces considerable internal power loss, the impedance of the magnetic amplifier is an inductive reactance, which dissipates but little energy as heat, giving back most of it to the circuit, like any other inductance. It can be made in any size depending upon its application and can be extremely sensitive. Magnetic amplifiers have been made which gave a power gain of 100,000. There is no limit to the kilovolt-ampere capacity, except size, which makes it ideal for controlling high power. Designs in Germany ranged up to 50,000 kva.

The magnetic amplifier will undoubtedly have extensive use, not only in the electrical control of military and naval equipment but especially in commerce. For control of many manufacturing processes and automatic devices for private use, it cannot be surpassed. It would be excellent if developed for underground and transoceanic cable use for, containing no moving parts and not requiring servicing, it can be installed in such a cable system and forgotten.

The magnetic amplifier is, at the present time, far from being a perfectly developed device. Its two major faults are that it has an inherently slow response time, and it is an efficient amplifier only if it has a core material of extremely high permeability. The latter of these two deficiencies can be improved with the development of core material made of noncritical materials which are sufficiently cheap so that magnetic amplifiers can be used as expendable devices.

Germany has returned to us a much improved magnetic amplifier together with an excellent selenium rectifier; Japan has provided vitally important basic research on efficient nonstrategic core materials. The magnetic amplifier today is at a developmental stage comparable to that of the vacuum tube 20 years ago. Its future capabilities are limited only by the degree of attention it receives from research agencies in this country. The Office of Naval Research, realizing the potentialities of this device, is coordinating a concerted program of fundamental and applied research on the magnetic amplifier and its applications in coordination with the Bureau of Ships, the Bureau of Aeronautics, the Bureau of Ordnance, and the Army. The improvement and utilization of this device may well prove to be a most important technical development.

Since this paper was written, Dr. Uno Lamm has visited the United States and lectured under sponsorship of the United States Navy on the theory and applications of the magnetic amplifier. He brought several pieces of demonstration apparatus with him and it was learned that Sweden is probably farther ahead in the techniques of utilizing the magnetic amplifier for power control and measurement than any other country at this time.

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*Although definitely not as widely known or applied domestically as electronic tubes, magnetic elements have, as this bibliography indicates, been used in various ways. If readers know of other literature in which controlled reactors are described, we would appreciate their calling it to our attention. THE EDITORS.

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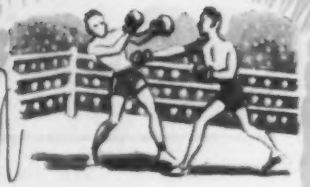

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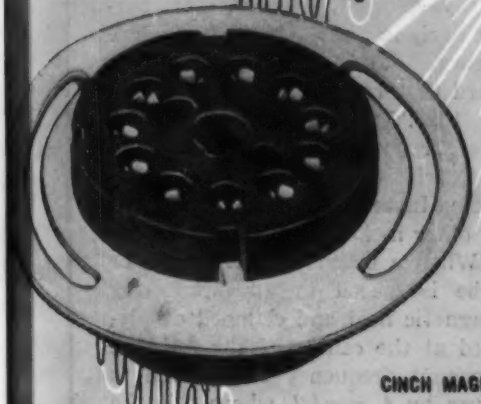

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CINCH DIHEPTAL SOCKET—Full floating contacts designed to insure easy insertion of tube and yet provide excellent electrical connections. Quick to wire. Mounting ring permits easy orientation of tube. Available in two sizes.

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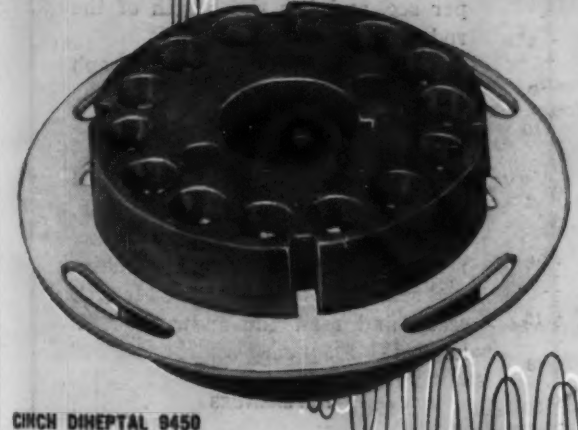
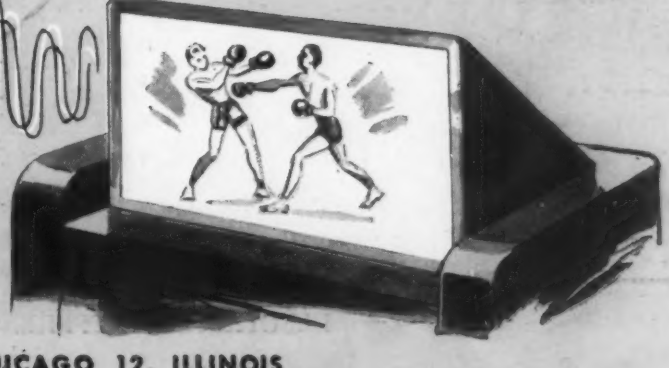


CINCH DIHEPTAL 9450



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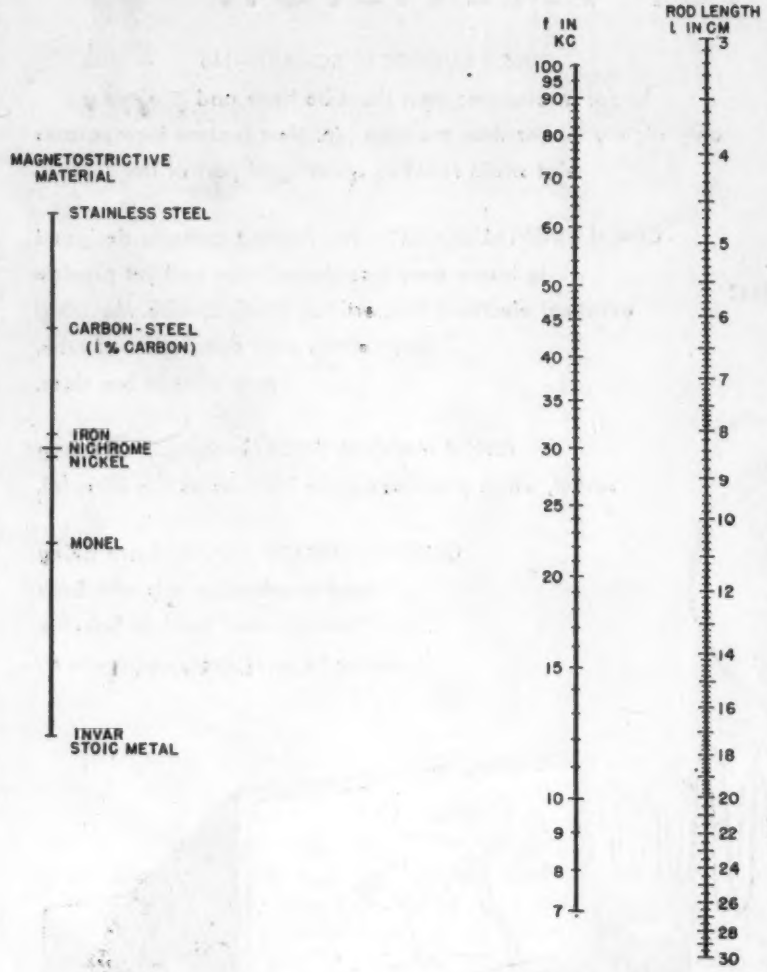
Magnetostriction Resonant Frequencies

Nomograph for ultrasonic and sonic applications gives natural vibrating frequency of oscillator when rod length and magnetostrictive material are known

By R. C. COILE
Colton & Foss, Inc.
Washington, D. C.



MAGNETOSTRICTION OSCILLATORS have been widely used in many electronic applications, particularly in underwater sound, since George Washington Pierce of Harvard first reported on his research in 1927.[1,2] These oscillators have their frequency controlled by a magnetostrictive rod of nickel, Monel metal, Invar, Nichrome, Stoic metal, or other nickel alloys.

When a magnetostrictive rod or tube is placed in an alternating magnetic field and clamped or pivoted at the center of the rod, the resonant frequency of the rod is given by $f = v/2l$, where f is the resonant frequency in kc, v is the velocity of sound[3] in the rod in cm per sec, and l is the length of the rod in cm.

The accompanying nomograph permits simple calculations of the resonant frequencies. Thus, to find the resonant frequency of a magnetostriction oscillator using a Monel metal rod 5 centimeters long, lay a straightedge between Monel metal on the magnetostrictive material scale and 5 cm on the rod length scale, and read the answer of 47 kc on the frequency scale.

References

(1) Patent 1,750,124, March 11, 1930, Vibratory System and Method. Application filed January 3, 1927.

(2) G. W. Pierce, *Proc. Amer. Acad. Arts & Sci.*, Volume 63, April 1928. Reprinted in *Proc. I.R.E.*, Volume 17, p 42, 1929.

(3) Bertha W. Henvis, Wavelengths of Sound, ELECTRONICS, p 134, March 1947.

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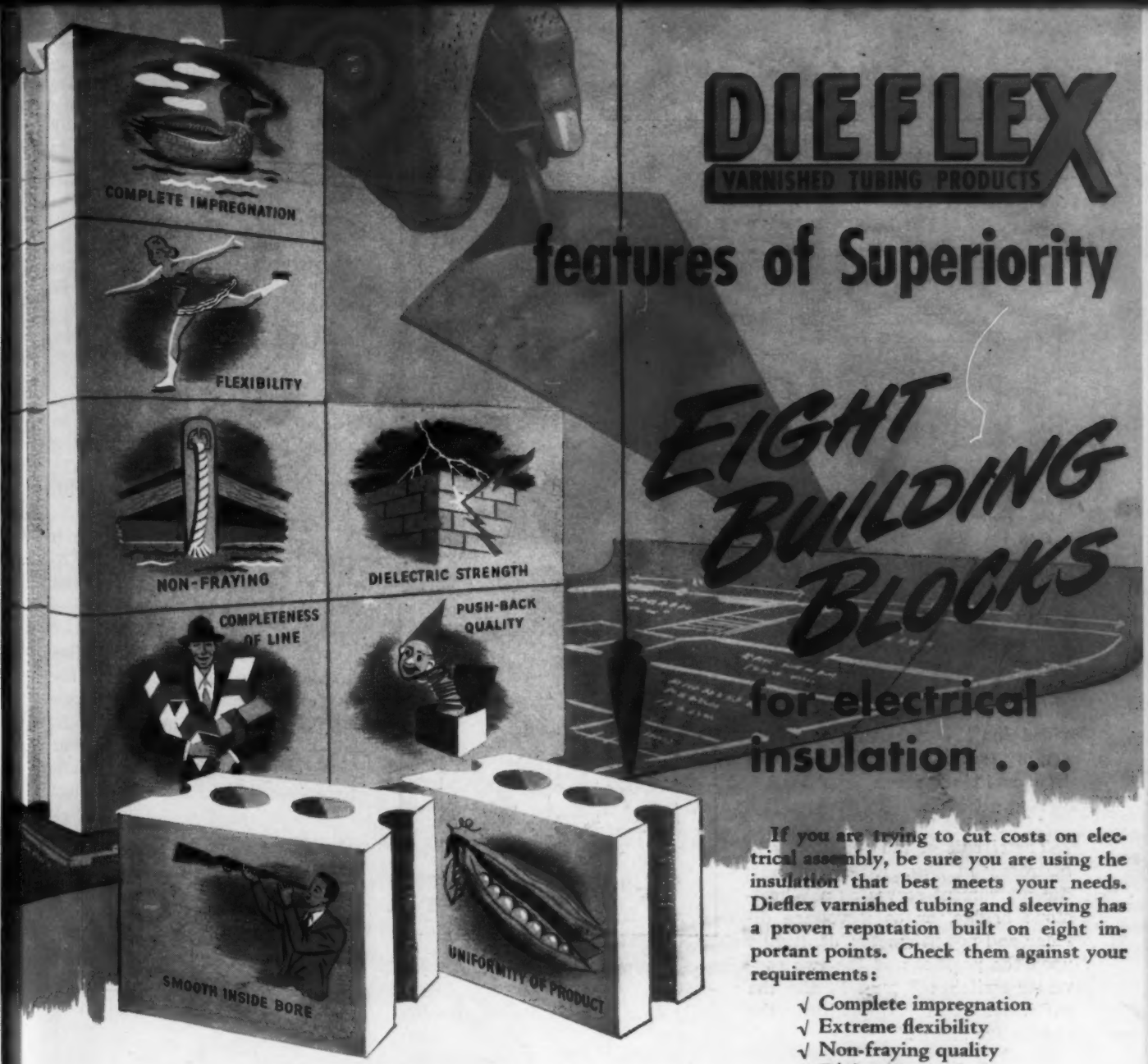

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TUBES AT WORK

Including INDUSTRIAL CONTROL

Edited by VIN ZELUFF

Sprayed Metal for Wiring 132
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Cathode-Coupled Half-Shot Multivibrator 150
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Hardening Shear Blades 164
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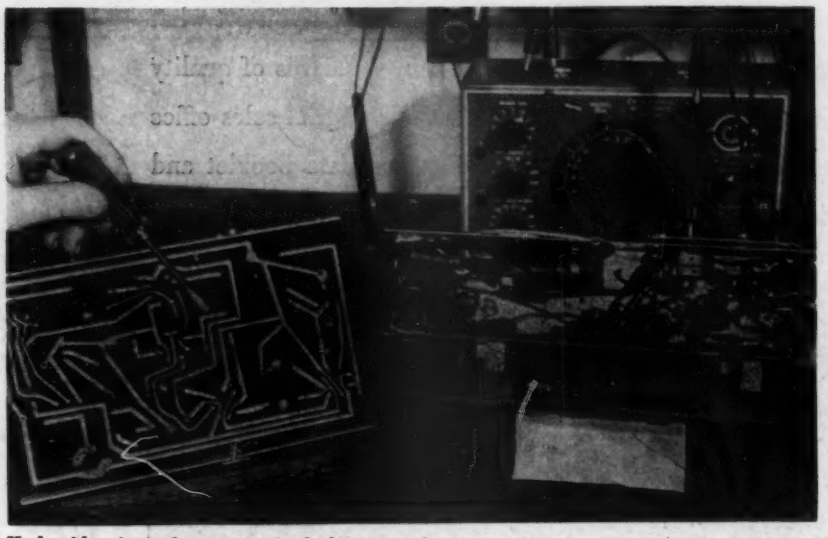
Sprayed Metal for Wiring

A NEW method of forming several electronic circuits at once consists of spraying metal into grooves in dielectric panels. A stencil is cut to coincide with the wiring diagram and the panel is sand blasted through the stencil to form grooves. Metal, copper, silver, or aluminum, is then sprayed through the stencil into the grooves. The panels may be formed of plastic, ceramic, glass, or wood.

Tube sockets, capacitors and other parts are assembled on the panel prior to sandblasting or metallizing, with their contacts so placed that they will be in the proper grooves when the latter are formed. The contacts are spray-wired into permanent connection with the circuit when the panel is sprayed. Both sides of a penal can embody sprayed conductors, and the conductors on opposing sides can be connected by metal eyelets prior to sandblasting or metallizing.

The thickness of the metal deposited is at least 0.005 inch. If insulation of a conductor is required, spraying, dipping, or brushing on suitable insulation material can be done. According to Spraywire Laboratories, Inc. of Minneapolis, Minn., the Spraywiring method is adaptable to the manufacture of many types of products which embody wiring. It can be used in producing the circuits of a tiny hearing aid, or the basic wiring of a prefabricated house. The metallizing equipment used in the spraying process is made by Metallizing Engineering Co., Inc.

In the Spraywiring technique, a stencil is applied to a blank insulating panel, left, which is then sandblasted and sprayed with metal. It then appears as at center, with stencil still in position. A finished two tube amplifier is shown at right with stencil removed



Underside view of a spraywired chassis and a conventional receiver having wires and soldered connections

Circuit Analyzer for The Blind

BY CLIFFORD M. WITCHER
Haskins Laboratories, New York City

SEVERAL MODELS of a multitester, capable of rapid and accurate use by the blind and essentially equivalent to moderately priced commercial circuit testers, were built and have proved to be quite satisfactory. The instrument, is considerably more rugged both electrically and mechanically than most commercial units, and measurements can be made with it at a comparable speed.

The analyzer embodies the equivalent of a multirange precision bridge-type ohmmeter, a multirange 2,000 ohms per volt d-c voltmeter, and a multirange d-c milliammeter. Measurements are obtained as nuls in an audible signal from a pair of headphones, and readings are easily taken from a single linear raised scale on the dial of an accurately linear potentiometer.

Addition of an external crystal rectifier or diode with suitable bias permits use as a peak reading a-c volt or milliammeter. Provision is also made for the insertion of external series resistors to provide any number of additional resistance

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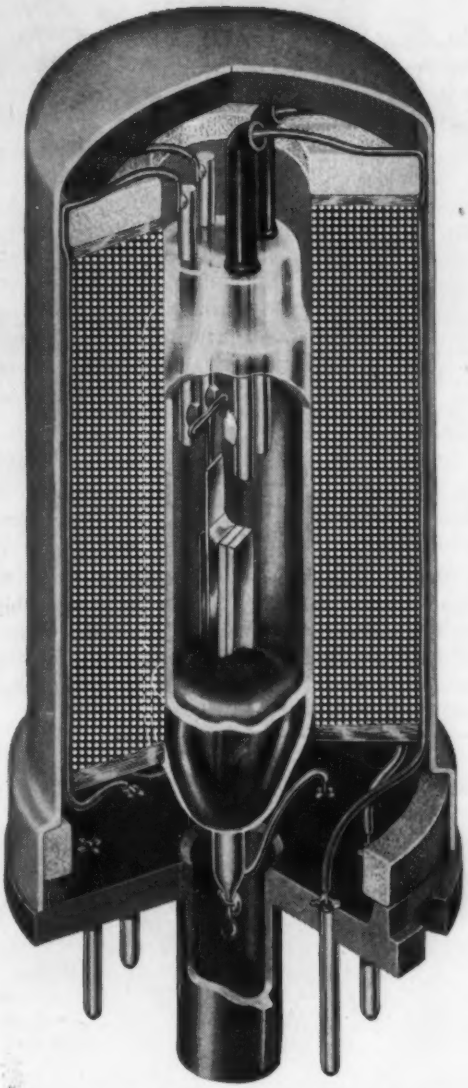

This cutaway view shows the glass sealed, mercury wetted contact switch surrounded by the operating coil and encased in a metal housing mounted on an octal base.

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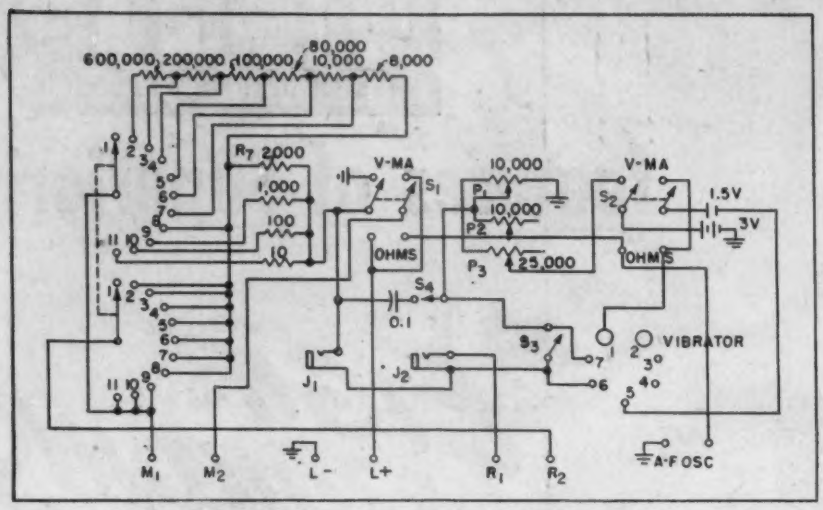

FIG. 1—Circuit of the multitester that provides aural indication of measurements

or voltage ranges, and for the insertion of shunts to provide higher current ranges. The instrument may also be used as an a-c bridge by connecting an external audio oscillator to the binding posts provided for this purpose and by inserting suitable standard capacitors or inductances.

Theory and Operation

The circuit of the analyzer is illustrated in Fig. 1. Essentially, the instrument serves as a Wheatstone bridge, or as a voltage divider and potentiometer, depending upon the positions of the two converter switches S_1 and S_2 and the particular jack into which the phones are plugged.

Consider the action of switch S_1 if binding posts M_1 and M_2 are connected together. When the blades of this switch are down and an unknown resistance X is connected across binding posts $L+$ and $L-$, it can be seen that a bridge circuit is formed by (1) the portion of the measuring potentiometer P_1 between the grounded end and the movable contact; (2) the calibrating potentiometer P_2; (3) the unknown resistance X; and (4) a portion of the resistance network associated with the range selector switch. With switch S_2 in the position correspond to that of S_1, three volts are applied across one diagonal of this bridge, and a pair of phones inserted in the other diagonal, jack J_1. The 2-volt vibrator contacts, shunted by a microswitch push button (both normally open) are in series with the phones.

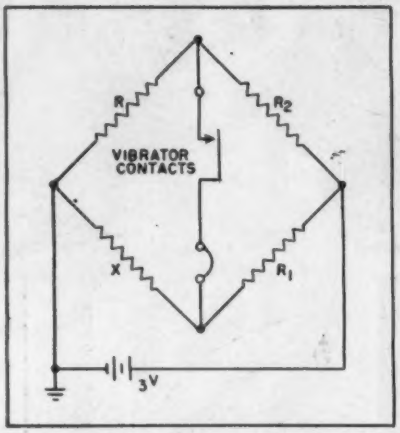

FIG. 2—Bridge circuit employed in the analyzer when resistance is being measured

Referring to Fig. 2, let R_1 be the resistance selected by the selector switch, and R_2 the resistance of the calibrating potentiometer P_2. When the instrument is in proper calibration, R_2 is set equal to the resistance of the portion of the measuring potentiometer P_1 between ground and full-scale position of the movable contact, a scale reading of 10.0. If the bridge is balanced for an unknown resistance X, by rotating the pointer of the measuring potentiometer until a null point is found, and if R is the resistance of the portion of the measuring potentiometer between ground and the movable contact, then $R/R_2=X/R_1$. But calling the scale reading Y, and remembering the definition of R_2, we have $R/R_2=Y/10$ therefore $X=(Y/10)\ R_1$. Since the values used for R_1 are always multiples of 10, the value of X can be readily obtained mentally from the scale reading. Since the scale only extends to 11.0, X must be less than or equal to $1.10R_1$.

Voltage and Current

Consider now the action of the instrument when switches S_1 and S_2 are in the V-MA position, and the phones are in jack J_2. Assume that binding posts R_1 and R_2 are connected together, as they are except when an external rectifier is to be used. For positions 2 through 8 on the range-selector switch, the resistance network associated with this switch is connected as a voltage divider, with the voltage drop across the 2,000-ohm resistor R_7 being opposed (through phones and vibrator) by the voltage between ground and the movable contact of the measuring potentiometer (See Fig. 3). The voltage across the measuring potentiometer is, in this case, obtained from the same flashlight cells mentioned above, but this time through the voltage calibration potentiometer P_3. The latter is so adjusted that the voltage drop across the first 10 scale divisions of the measuring potentiometer is 1.0 volt. In practice it is found best to check the setting of P_3 by comparison with a reasonably

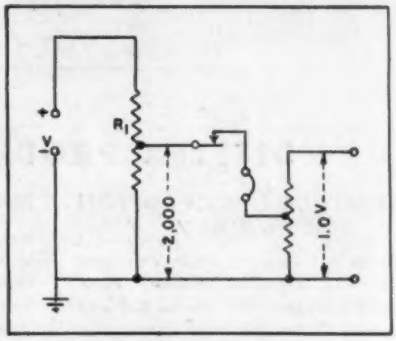

FIG. 3—Voltage divider and potentiometer arrangement for current and voltage measurements

accurate voltage source every few weeks.

With this circuit arrangement the voltage V corresponding to a given scale reading Y at which a balance has been obtained is $v = (y/10)\ (R_1/2000)$. In practice, one remembers the range selector switch settings as voltage ranges

(continued on p 148)

THE ELECTRON ART

Edited by FRANK ROCKETT

Designing Automatic Controllers by Electrical Analogues 136
Steepness of Pulse Fronts 136
Voltmeter for Pulses 178
Pulse Rise and Decay Time Measurement 180
Survey of New Techniques 190

Operator observes effect of variations in simulated process control

Designing Automatic Controllers by Electrical Analogues

DYNAMIC STABILITY and other performance criteria of automatic industrial controllers can be determined automatically using an electronic model of the system. Using this technique, the design of controllers is simplified by the Analaut (automatic analogue) designed by George A. Philbrick, G. A. Philbrick Researchers, Inc. (Boston, Mass.).

Electrical Analogue

The components of the controlled process are reduced to their electrical equivalents. The complete system, with either the actual controller or an electrical analogue of it, is then periodically shock excited and the recovery transients at appropriate points synchronously displayed on an oscilloscope. The behavior of the system for various adjustments of the process and controller is observed by manipulation of the electrical circuit in direct correspondance with variations of the actual physical constants. For example, effects of lag times or valve limits can be analyzed.

Various electrical circuits are used to represent plant equipment. A series of tanks so connected that the fluid cannot back up, as where one tank overflows into another, is represented by resistance-capacitance networks isolated by vacuum tubes; if the fluid can back up, the tubes are omitted. The inverse of the technique is used to synthesize practical processes and plants from the required physical reactions. In addition, because the electrical analogue compresses into fractions of a second plant reactions occupying hours, it provides a means of quickly and graphically teaching automatic control techniques to plant operators. A classroom model has recently been successfully used at Massachusetts Institute of Technology. Special models have been used in designing and improving industrial controllers, and others are being used in designing nonreciprocating aircraft engine controls.

Steepness of Pulse Fronts

BY MILTON D. RUBIN

Engineer, Raytheon Mfg. Co.
Dorchester, Mass.

MOST ANALYSES of pulsed circuits assume that the wavefront of the pulse is a step function, such as

ELECTRONS RADIATE LIGHT

Electrons traveling at extremely high speed give off electromagnetic radiation directly. This electronic light has just been observed by Floyd Haber, laboratory assistant, working on the synchrotron in the General Electric research laboratory. The radiation, predicted in 1944 by two Russian physicists, D. Iwanenko and I. Pomeranchuk, is given off tangentially from the free speeding electrons, and appears as a bluish-white spot; the cross bar in the photograph is produced by refraction at the glass wall of the synchrotron doughnut. Calculations by Dr. J. S. Schwinger of Harvard University indicate that maximum energy of the radiation is in the visible or infrared portion of the spectrum, with only slight energy in the microwave region. Dr. H. C. Pollock, physicists in charge of the synchrotron, and Drs. R. V. Langmuir and F. R. Elder, his associates, are preparing a flat quartz window in the doughnut wall so that measurements of the spectral content of the radiation can be made. Comparison of the measurements with calculations will provide further insight into the nature of the electron

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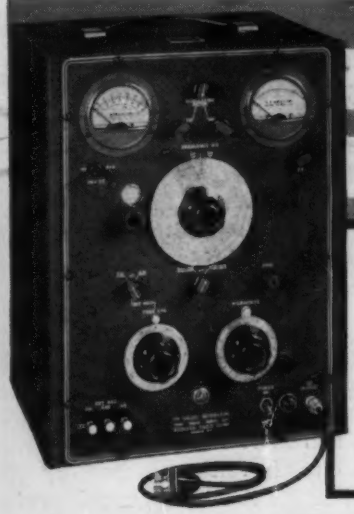
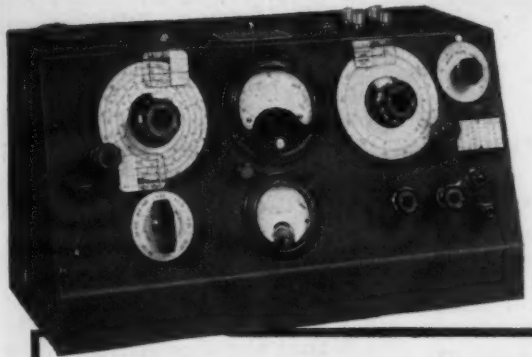


FM SIGNAL GENERATOR *TYPE 202-B*

RF COVERAGE: 54—216 megacycles in two ranges; accuracy ±0.5%.
FM DEVIATION: Two ranges, 0—80 kc., 0—240 kc., continuously adjustable.
FM DISTORTION: Less than 2% at 75 kc. deviation.
AMPLITUDE MODULATION: One range, 0—50%; calibrations at 30% and 50% modulation.
INTERNAL MODULATING FREQUENCIES: Eight frequencies from 50 cycles to 15 kc.
RF OUTPUT: 0.1 microvolt to 0.2 volt, continuously adjustable by means of 26.5 ohm piston type attenuator.

UNIVERTER *TYPE 203-B*

An FM-AM frequency converter accessory with unity gain for use with 202-B FM Signal Generator.
RF RANGE: 0.4 mc. to 25 mc.
RF INCREMENT DIAL: ±250 kc. in 10 kc. increments.
RF OUTPUT: 0.1 microvolt to 0.1 volt, calibrated output (26.5 ohms); 2 volts max. uncalibrated output (approx. 470 ohms).

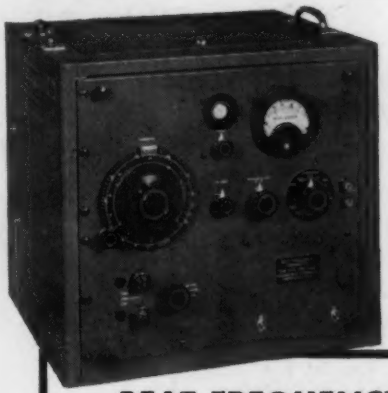


Q-METER *TYPE 160-A*

FREQUENCY RANGE: 50 kc. to 75 mc. (±1%, 50 kc.—50 mc.; ±3%, 50 mc.—75 mc.). May be extended down to 1 kc. with external accessory oscillator.
RANGE OF Q MEASUREMENTS, COILS: 50—625.
ACCURACY OF Q MEASUREMENT: In general, ±5%.
RANGE OF Q TUNING CAPACITOR: Main Section, 30—450 mmf. (±1% or ±1 mmf., whichever is greater). Vernier Section, plus 3, zero, minus 3 mmf., (±0.1 mmf.).

Q-METER *TYPE 170-A*

FREQUENCY RANGE: 30 mc. to 200 mc. (±1%).
RANGE OF Q MEASUREMENT: 80—1200.
ACCURACY OF Q MEASUREMENT: In general, ±10%.
RANGE OF Q TUNING CAPACITOR: 11—60 mmf. (±1% or ±0.5 mmf., whichever is greater).



QX-CHECKER *TYPE 110-A*

FREQUENCY RANGE: 100 kc. to 25 mc. using accessory plug-in inductors. A calibration sheet is supplied with each inductor, accurate to within ±3%.
RANGE OF COIL CHECKS: Inductors having values between 10 microhenries and 10 millihenries may be checked against a standard to an accuracy of about ±0.2% provided the Q of the inductor is 100 or greater.
CAPACITANCE RANGE: Capacitance values between about 2—1000 mmf. may be checked against standard by direct substitution, with an accuracy of a few tenths of one mmf., provided Q of capacitors is high.

BEAT FREQUENCY GENERATOR *TYPE 140-A*

FREQUENCY RANGE: 20 cycles to 5 megacycles in two ranges; LOW RANGE: 20 to 30,000 cycles. HIGH RANGE: 30 kilocycles to 5 megacycles. Accuracy ±2 cycles up to 100 cycles, ±2% above 100 cycles.
OUTPUT POWER: One watt, available from a variety of output impedances.
ATTENUATOR: 5 steps; X1.0, X0.1, X.01, X.001, X.0001.
DISTORTION: 5% or less.

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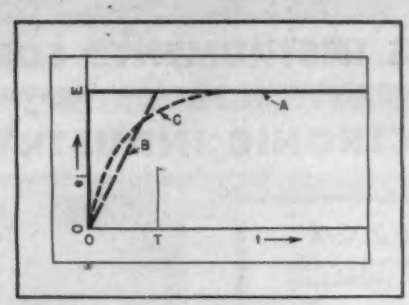


FIG. 1—Pulse front can be approximated by a step function (curve A), a linearly rising front (curve B), or an exponentially rising front (curve C)

curve A in Fig. 1. Actually it is physically impossible for a voltage or current to change instantaneously from one value to another. Therefore a wavefront cannot have infinite slope as implied by this function. There are cases in which this assumption of a perfectly square pulsefront leads to erroneous results. In these cases the following analysis gives better results with no greater complexity.

Linearly Rising Pulse Front

Often it is appropriate to assume an infinite slope as a first approximation, and to interpret the results accordingly. However when this approximation gives incorrect results (One of the principal cases being an analysis of pulse differentiation), a second approximation to the solution, a linearly rising pulsefront such as curve B in Fig. 1, which is close enough to be of great practical help, can be assumed. A fairly complete analysis of this problem has assumed exponential rise of the pulse, such as Curve C in Fig. 1, which might be considered as a third and even closer approximation[1].

Nevertheless, assumption of a linear rather than exponential rise may, at times, be more accurate, as the rise is often linear until just before flattening out, particularly when peaking circuits are used.

What the practical electronic designer prefers is a few rules of thumb for quick application to various cases, especially when numerous stray effects need to be considered and the circuit values are not accurately known.

Assume a linear rise of the pulse. By way of illustrating the approach, a simple R-C circuit, shown in Fig. 2, is considered because of its simplicity and wide application. Likewise, the essentials of the mathematical analysis are given because of their simplicity and because the method can be used for other circuits.

The voltage rises from 0 to E in time T, therefore the rate of rise is E/T, and the input voltage is $e_i = Et/T$. The current can be found by dividing the operational transform expression for the voltage by

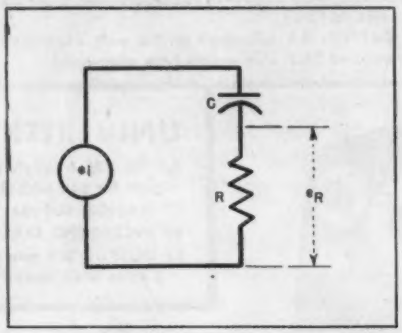


FIG. 2—Curves in Fig. 3 pertain to this simple but common circuit

the operational transform for the impedance, and solving for the current by known equivalences (The operational transform method, giving direct solutions of pulse transmission problems, should be at every electronic engineer's fingertips.), or by contour integration in the complex plane. The expression for the voltage by the operational transform method[2,3], for the rise period, is therefore E/Tp, where p is the well known operator. The impedance is $R + 1/Cp$, thus

$$i :: \frac{E/Tp}{R + 1/Cp} \qquad (1)$$

where :: denotes operational transform equivalence. The solution is

$$i = (CE/T)(1 - \epsilon^{-t/RC}) \qquad (2)$$

Usually we are interested in the voltage across the resistor. This voltage is

$$e_R = (RCE/T)(1 - \epsilon^{-t/RC}) \qquad (3)$$

which is of the form of curve C in Fig. 1.

Interpretation of Results

To interpret this equation into practical rules of thumb, let us consider several cases, realizing that for the present we are considering only the rise period.

First, for t considerably less than RC

$$e_R = Et/T \qquad (4)$$

Thus the voltage follows and is equal to the input voltage, as it also has this equation. If T is considerably less than RC, then the whole wavefront is transferred faithfully across the resistor.

For the second case, consider T approximately equal to RC. Then $t = T = RC$, substituting in Eq. 3.

$$e_R = (RCE/RC)(1 - \epsilon^{-1}) = 0.6321E \qquad (5)$$

or e_R is approximately $(2/3)E$. The voltage across the resistor thus rises exponentially to about $(2/3)E$, and there is some differentiation and loss in voltage.

For the third case, consider when the wavefront rises to its full value in time $t = T = 3RC$. Here again the voltage rises exponentially to its final value, which is, from Eq. 3

$$e_R = (E/3)(1 - \epsilon^{-3}) = (0.95/3)E \qquad (6)$$

or e_R is approximately $(1/3)E$. In this case we have lost 2/3 of the voltage, and are now in the region in which many differentiating circuits work. The analysis indicates that, unless the rise time of the input voltage is about the same as or less than the time constant of the differentiating circuit into which it is working, much of the voltage is lost in passing through the circuit.

For the fourth case, consider a rise time very much larger than RC. For $t = T = aRC$, where a

(continued on p 172)

FRENCH ATOMIC PILE

Electronic instrumentation is being used extensively at plants outside Paris, France in developing nuclear piles, first step in manufacture of radioactive materials

Put This Formula To Work For You

$$I = \frac{E}{IRC}$$

The formula $I=\frac{E}{IRC}$ is a favorite and easily-remembered solution to resistance problems. Radio and electronic Engineers know that IRC offers the most complete line of resistance products in the industry . . . a fixed or variable resistor for most every requirement . . . with uniform dependability proved by years of rigorous laboratory and field tests. Purchasing Agents and material control executives like IRC's service . . . "on-time" deliveries . . . factory stock-piles of the most popular types and ranges from which they can draw in emergency . . . IRC's distributor network, providing speedy, 'round-the-corner service for small order requirements.

Put this formula to work for you . . . check below the catalog bulletins in which you are interested, and write for them today on your letterhead, giving your name and title. International Resistance Company, 401 N. Broad Street, Philadelphia 8, Pennsylvania. In Canada: International Resistance Company, Ltd., Toronto, Licensee.

INSULATED COMPOSITION INSULATED WIRE WOUND RESISTORS
IRC types BT & BW ☐

FINGER-TIP CONTROLS AND SWITCHES
IRC types H & SH ☐

WIRE WOUND RESISTORS
IRC type MW ☐

MATCHED PAIR RESISTORS ☐

POWER RESISTORS ☐

FLAT POWER RESISTORS
IRC type FRW ☐

VOLUME CONTROLS
IRC type DS ☐

WIRE WOUND POTENTIOMETERS
IRC type W ☐

VOLTMETER MULTIPLIERS
IRC types MFA & MFB ☐

POWER RHEOSTATS
IRC types PR & PRT ☐

WATER-COOLED RESISTORS
IRC type LP ☐

PRECISION RESISTORS ☐

HIGH VOLTAGE RESISTORS
IRC type MV ☐

HIGH POWER RESISTORS
IRC type MP ☐

INTERNATIONAL RESISTANCE COMPANY

Wherever the circuit says -WV-

Power Resistors • Precisions • Insulated Composition Resistors • Low Wattage Wire Wounds • Rheostats • Controls • Voltmeter Multipliers • Voltage Dividers • HF and High Voltage Resistors

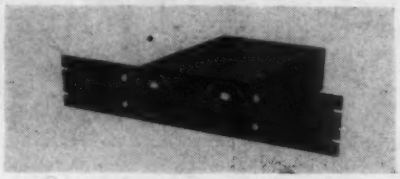
NEW PRODUCTS

Edited by A. A. McKENZIE

New equipment, components, packaged units, allied products; new tubes. Catalogs and manufacturers' publications reviewed.

Audio Compensator (1)

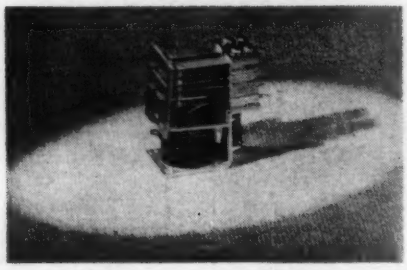
ARLINGTON ELECTRICAL PRODUCTS, INC., 18 West 25th St., New York 10, N. Y. The model EA-2 audio compensator is particularly applicable for use in audio equalization



of film and disc recording and broadcast studio work. The rack-mounted equipment is completely described in curves and a specification sheet available from the manufacturer.

Small Multipole Relays (2)

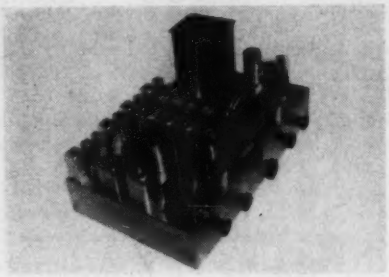
LEACH RELAY CO., 5915 Avalon Blvd., Los Angeles 3, Calif. A new line of small multipole relays is now available for circuits not exceeding



12.5 amperes, noninductive, in combinations up to 4-pole double-throw. Coils can be furnished for a-c or d-c operation. Full details are given in Bulletin 202.

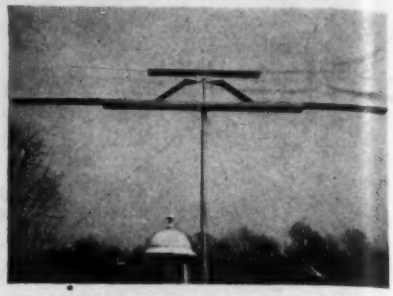
F-M/A-M Tuner (3)

COLLINS AUDIO Products Co., 126 Park St., Westfield, N. J., announces the new f-m/a-m tuner which provides the finest possible reproduction from the present broadcast band. Frequencies up to 10,000 cycles are passed without loss. The combination of two separate tuners, having only the audio tube common to both, eliminates dead spots due to interaction. Other features are the new tuning eye (6AL7-GT) to provide the accurate tuning necessary on the f-m band, and the heavy duty power supply which is an integral part of the chassis making for compactness. A total of 17 tubes is used in the tuner.



Broad Band Antenna (4)

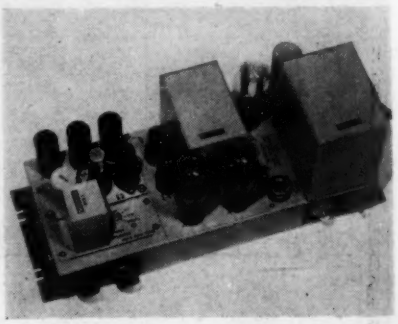
F. A. KOLSTER, VIENNA, VA. The broad band antenna illustrated can be used for both f-m bands and the



television channels without readjustment. Bidirectional characteristics of the antenna are essentially the same over the range 40.9 to 210 mc. Graphs showing gain relative to a dipole, standing wave ratios and measured field patterns can be obtained from the manufacturer.

Wired Music Amplifier (5)

THE LANGEVIN CO., INC., 37 West 65th St., New York 23, N. Y. The model 610 amplifier designed for wired music has a power output of 20 watts with less than 2 percent rms total harmonic distortion from



50 to 15,000 cycles. Input and output circuits are particularly arranged for connection between telephone line and speakers or distribution system. Full engineering details are available in a 4-page two-color brochure.

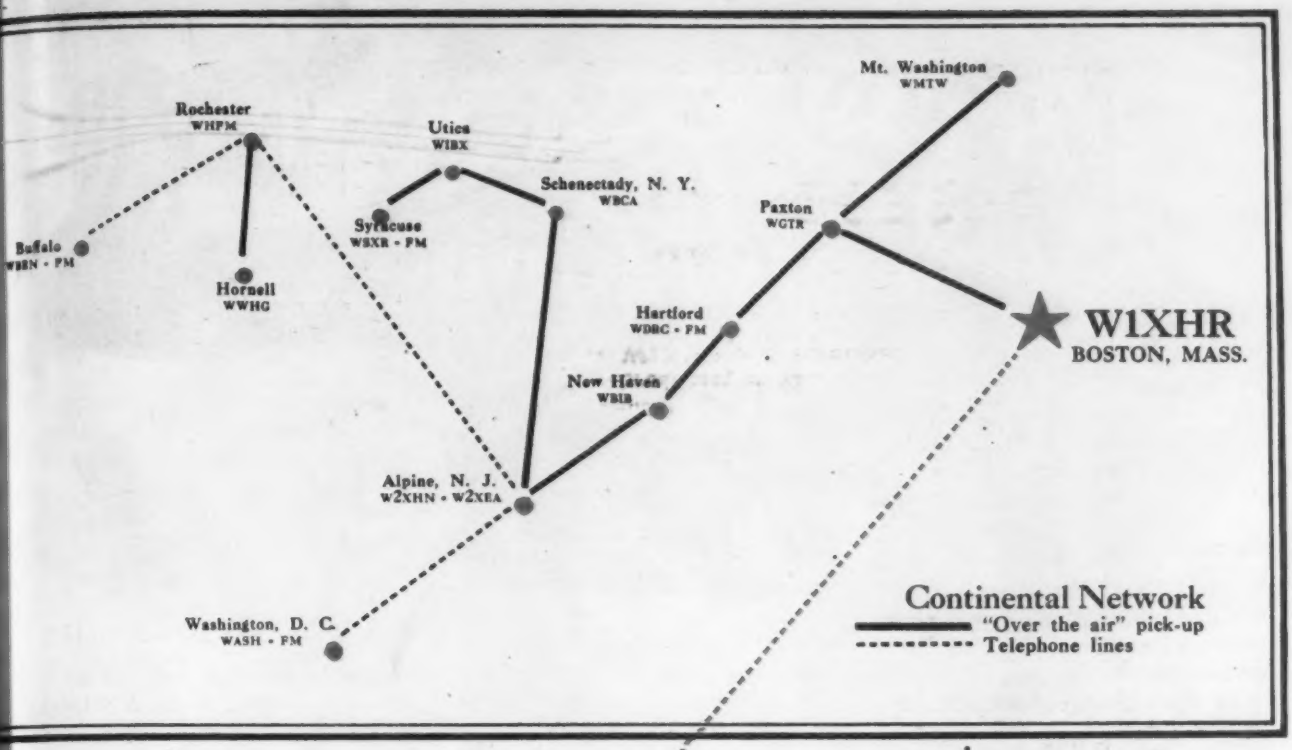
Tuning Fork Standard (6)

AMERICAN TIME PRODUCTS, 580 Fifth Ave., New York 19, N. Y. Type 2001 tuning fork frequency standard is available for any frequency in the range from 200 to 1,000 cycles. The fork, of bimetallic construction, is housed in a hermetically sealed, shock-mounted container. Output is 5 volts at 150,000 ohms, approximately a sine wave. Frequency accuracy is 0.001 percent and the temperature coefficient

USING THE NUMBERS

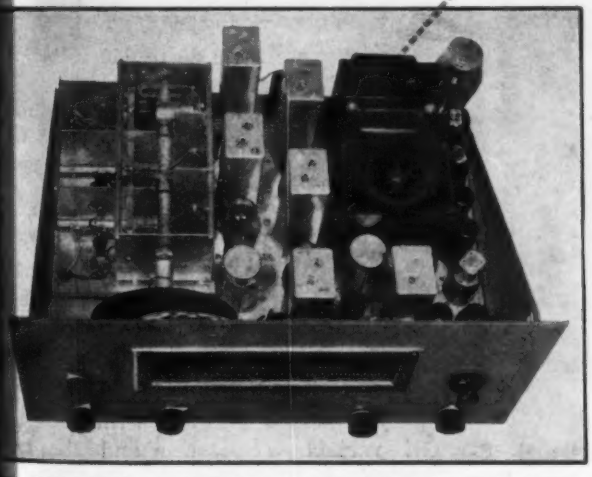
Readers desiring further details concerning any item listed in the New Products department can obtain the information by using the cards furnished as a stiff, colored insert elsewhere in this department.

Place the number (appearing to the right of the heading) of one item in which you are interested in a circle and then fill out the balance of the card according to directions appearing on the colored sheet. Unnumbered items listed at the end should be procured direct from the manufacturer or publisher upon payment of the fee noted.



The FM Receiver with Network Experience

In the recently formed Continental Network, W1XHR, Harvey's High Fidelity FM station, serves as the FM outlet for Eastern Massachusetts. Programs originating in Washington, D. C. are received from WGTR in Paxton, Mass. on 103.1 Mc. and rebroadcast on 92.1 Mc. Harvey's 193-R High Fidelity FM receiver is used as the Relay-Link receiver at W1XHR and furnishes quality audio signals to its transmitter for rebroadcast purposes.



The Model 193-R High Fidelity FM Receiver (top view) employs a double conversion superheterodyne circuit using two I.F. channels 10.7 Mc. and 4.6 Mc.

Price $490 f.o.b. Cambridge, Massachusetts

CHARACTERISTICS:

Frequency Range: 85 to 115 Mc.

Audio: an extremely high fidelity audio system is used, providing flat response from 10 to 40,000 cycles.

Sensitivity: RMS quieting signal of 1 microvolt.

Bandwidth: 250 Kc., adequate to pass the sidebands produced by 15 Kc. modulation, without appreciable band-edge attenuation.

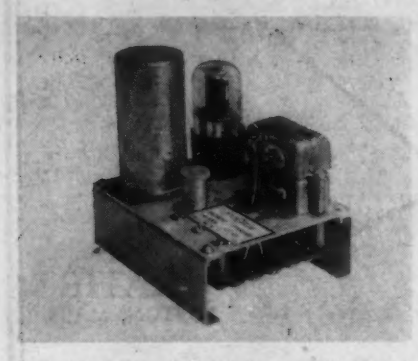
Squelch Circuit: smooth, fast acting and adjustable from .1 microvolt to 1 microvolt.

Size: 21½" x 10½" x 15", available in an attractive gray wrinkle-finish cabinet.

Write for HAR-CAM *further information*

HARVEY RADIO LABORATORIES, INC.

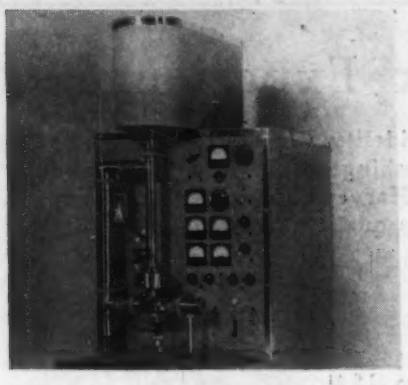
439 CONCORD AVENUE • CAMBRIDGE 38, MASSACHUSETTS



is better than 1 part in a million per degree Centigrade from 0 to 65 C. Other types are also available.

Microoscillograph (7)

CENTRAL RESEARCH LABORATORIES, Inc., Red Wing, Minn. A new high-speed microoscillograph now in production extends the range of application of single-sweep recording by



a factor of 10 in frequency over previous limits. Recordings of three simultaneous phenomena at frequencies up to 10,000 megacycles have been made.

Reproducers (8)

RADIO-MUSIC CORP., East Port Chester, Conn. Vertical, lateral or universal heads with linear response from 40 to beyond 11,000 cycles, together with arms and an equalizer are now in production under the name Para-Flux. Bulletin PR6 gives

full details of interest to broadcasters, recording studios, theaters and other users of recordings.

High Frequency Converter (9)

COLUMBUS ELECTRONICS INC., 229 So. Waverly St., Yonkers, N. Y. The model HFC 610 converter is

available in two ranges, 27 to 30 mc and 50 to 54 mc. When connected to a communications receiver it provides sharp tuning and image-free reception. It is further described in Bulletin C5.

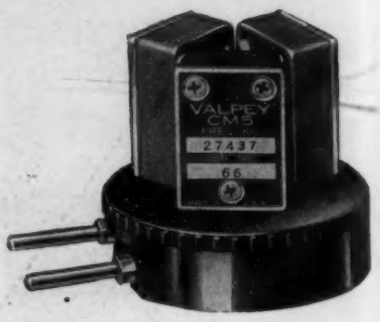
Studio Recorder (10)

FAIRCHILD CAMERA & INSTRUMENT CORP., 88-06 Van Wyck Blvd., Jamaica 1, N. Y. A new studio recorder of console type accommo-

dates all sizes of acetate and 18-inch flowed wax masters. The model 523 unit operates at 33⅓ rpm and will cut either outside-in or inside-out at any pitch between 80 and 160 lines.

Multicrystal Holder (11)

VALPEY CRYSTAL CORP., Holliston, Mass. Two or three transmitting crystals in a common holder can be used in sequence by turning a



knurled rim. The unit accommodates crystals mounted in holders having 0.094 diameter pins with half-inch spacing. The Xtalector can be supplied in two types to fit either ½ or ¾-inch sockets.

Neon-Glow Voltmeter (12)

INDUSTRIAL DEVICES, INC., 22 State Road, Edgewater, N. J. A midget neon-glow voltmeter is calibrated

for use on a-c from 65 to 660 volts with an impedance of 0.5 megohm. Voltage is read directly from the scale. When connected to a d-c source, only one electrode glows and the reading is multiplied by 1.15.

Pickup Arm (13)

VIBRO-MASTER CO., 144 West 54th St., New York 19, N. Y. A new pickup arm, type K, has been particularly designed for use with the GE variable reluctance and the

(continued on p 194)

NEWS OF THE INDUSTRY

Edited by JOHN MARKUS

Technical programs of NAB Engineering Conference, Instrument Conference, and Western IRE Conference; status of facsimile; radar

Tracking Stations for Guided Missiles

INSTRUMENTATION capable of showing the speed, trajectory, range, and other characteristics of guided missiles for approximately the full 90-mile length of the White Sands Proving Grounds, Las Cruces, New Mexico, is to be installed by the Army Corps of Engineers in cooperation with the U. S. Navy.

The construction program calls for the erection of approximately 75 additional instrumentation stations of various types and sizes, dust-proofed as necessary to house doppler and radar equipment, tracking cameras, cine-theodolites and other paraphernalia.

Rapid developments in the guided missiles program have resulted in greater range of the missiles, and additional stations are necessary to enable observations and plotting of trajectories in the entire flight of such missiles.

As part of the coordinated program, the Army Air Forces also is installing a network of radar and other equipment to supplement the installations by Army Ordnance and the Navy. Both Army and Navy have firing stations at the White Sands Proving Grounds.

Electronics Conference In Chicago Nov. 3-5

THE COMPLETE program for the National Electronics Conference at the Edgewater Beach Hotel in Chicago Nov. 3-5 includes 20 technical sessions with a total of about 50 papers, two luncheons, and the main banquet, in addition to the general session Monday morning. Technical details of the new dynamic noise suppressor circuit invented by Herman Hosmer Scott will be released, and manufacturers will stage educational exhibits.

Major emphasis in the technical program is placed on industrial electronics, having a total of three sessions, with one of these arranged by the AIEE in connection with their national convention in Chicago the same week.

One session will be devoted to commercial, f-m, and television broadcasting, and another to color television and oscillography. Other papers will deal with new types of antennas and with instrumentation, and a panel discussion on electronic research will be led by experts in the field. Subjects of general interest to all will include guided missiles, electronic computers, ultrasonics, infrared, microwaves, audio circuits, and detection of particles.

PUTTING SOUND ON INDUSTRIAL MOVIES

This ultramodern audio console in the projection room of Reeves Sound Studios, New York City, is capable of mixing eight channels to a single output, as required for putting narration, music, and sound effects on sound tracks of industrial films with precise synchronization. Operator records footage at which each sound source comes in, along with all required data on volume level changes and fading schedules. Many practice runs are made before final sound channel is photographically recorded on the film. New Reeves studio provides over five floors of sound stages, shooting stages, recording rooms, and sound laboratories for recording sound for film, records, radio, and television

Pay-As-You-See Television

A METHOD of charging a fee for a particular television program, its amount varying with the nature of the entertainment, was recently announced by Zenith Radio Corporation. Known as Phone Vision, it operates by sending a vital portion of the picture signal over a telephone or electric power line to the user. Without this signal the picture is a confused and jumpy blur.

In operation, the subscriber would call the telephone operator when he desired to see a special program, and would receive it by

SPRING LOCK NUT

REVOLUTIONARY NEW FASTENER

COSTS LESS THAN NUT

ELIMINATES ONE OPERATION

SURE LOCK FOR BOLT

SPEEDS PRODUCTION

EASILY ASSEMBLED

CUTS TIME . . . MATERIAL . . . COSTS

Available in Two Types—Spring Nut and Spring Lock Nut

PROVED BY TESTS and hard usage on all kinds of products this new, revolutionary fastening device is already saving time, material and money for the nation's progressive manufacturers.

TWO TYPES of this new Diamond G Product, manufactured by the Garrett Company, are available. One replaces the conventional nut and is called the Diamond G Spring Nut. The other is made with spring action tension in it in such a way that it supplies action similar to that of a lock nut—the threads inside of the Diamond G Spring Lock Nut maintain a strong grip on the bolt, or screw and the tighter it is drawn up the greater the locking action.

AVAILABLE IN TWO AND THREE TURNS —depending on the requirements. The two turn type is for use on light assemblies and the three turn type is recommended when it is necessary to use heavy tightening torques on assembly.

FOR DRIVING OR SETTING—adaptable socket wrenches are now standard items as manufactured by Apex Machine and Tool Company. These new Diamond G Products—Spring Nut and Spring Lock Nut—have all of the advantages of regular nuts and are adaptable for use in hopper feeders. Both types are priced to offer a considerable saving over the conventional type of nut, or fastening device. These products are available in rust resisting metal as well as high carbon steel.

START YOUR PLANT on greater efficiency today . . . write for the complete data sheet on these new Diamond G time and money saving products.

GEORGE K. GARRETT CO., INC.

1421 CHESTNUT STREET, PHILADELPHIA 2, PA.

Manufacturers of

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LOCK WASHERS • • FLAT WASHERS • • STAMPINGS • • SPRINGS • • HOSE CLAMPS • • SNAP AND RETAINER RINGS

IF IT'S UHF EQUIPMENT

WE CAN

- **DESIGN**
- **DEVELOP**
- **PRODUCE IT!**

LAVOIE LABORATORIES are specialists in high frequency work. We can begin with your own ideas — design, develop and produce; or we can manufacture (in any quantity) from your blueprints.

In either event, we can do a business-like job because we have the personnel and equipment for precision work plus a background of practical experience.

A few typical examples of LAVOIE-produced equipment are shown, including Frequency Standards—Frequency Meters—Receivers, etc.

Complete information and detailed estimates of LAVOIE service are available promptly without obligation.

Lavoie Laboratories

RADIO ENGINEERS AND MANUFACTURERS

MORGANVILLE, N. J.

Specialists in the Development and Manufacture of UHF Equipment

MEETINGS

AUG. 26-29: AIEE 1947 Pacific General Meeting, Hotel San Diego, San Diego, Calif.

SEPT. 8-12: Second Annual Conference and Exhibit of The Instrument Society of America, at Stevens Hotel, Chicago.

SEPT. 12-13: FMA convention and exhibits, Roosevelt Hotel, NYC.

SEPT. 15-18: Annual convention and exhibit, National Association of Broadcasters, Convention Hall, Atlantic City, New Jersey.

SEPT. 23-25: AIEE Middle Eastern District Meeting, Dayton, Ohio.

SEPT. 24-26: West Coast IRE Convention, Palace Hotel, San Francisco, Calif., Prof. Karl Spangenburg, general chairman, Stanford University, Palo Alto, Calif.

SEPT. 26-28: Third annual Electronics Trade Show, Hotel Whitcomb, San Francisco, sponsored by West Coast Electronic Manufacturers' Association, W. Noel Eldred, secretary, care of Hewlett-Packard Co., Palo Alto, Calif.

SEPT. 26-28: Hudson Division ARRL Convention, Convention Hall, Asbury Park, N. J.; technical papers and exhibits of electronic equipment for radio amateurs.

SEPT. 29-OCT. 2: Annual meeting, International Municipal Signal Association, Inc., Pantlind Hotel, Grand Rapids, Michigan; technical papers and exhibits of police and fire radio equipment.

SEPT. 30-OCT. 11: National Radio Exhibition, Olympia Hall, London, sponsored by British Radio Industry Council and featuring new British radio, electronic control, radar, and television equipment.

OCT. 20-24: SMPE Theater Engineering Conference, Hotel Pennsylvania, New York.

OCT. 23-25: Annual Meeting, Optical Society of America, Netherland Plaza, Cincinnati, Ohio; symposium on microwave optics.

NOV. 3-5: National Electronics Conference, Edgewater Beach Hotel, Chicago.

NOV. 3-7: AIEE Midwest General Meeting, Chicago, Ill.

NOV. 7-8: Conference on X-ray and Electronic Diffraction, Mellon Institute of Industrial Research, Pittsburgh, Pa.

NOV. 17-19: Rochester Fall Meeting, Sheraton Hotel, Rochester, N. Y.; technical papers and exhibits.

MARCH 22-25: IRE Convention and Radio Engineering Show, Hotel Commodore and Grand Central Palace, New York City.

means of an attachment on the telephone line. He would be billed at the end of each month for the special programs actually received, much as he is now billed for long-distance calls. All standard television broadcasts would be free, and the key signal would be used only for top-flight entertainment features such as new superpictures, Broadway plays, current newsreels, and special sporting events. These features are not available for ordinary television because their high production costs make them far too expensive for an advertising sponsor.

In making the announcement, Commander E. F. McDonald, Jr., Zenith president, stated that up to a year would be required to get the system in operation in a few key cities now having television, that Phone Vision sets would soon be in production, and that existing television transmitters can easily be altered.

NEMA Establishes Electronic Heating Group

ELEVEN manufacturers of induction and dielectric heating equipment have formed a new Induction and Dielectric Heating Apparatus Section within the National Electrical Manufacturers Association. Chairman is Dr. H. B. Osborn, Jr., sales manager of the Tocco Division of The Ohio Crankshaft Co., and vice-chairman is C. W. Miller, sales manager of Westinghouse's Industrial Electronics Division in Baltimore.

The section was formed to insure greater benefits to users of high-frequency heating equipment by such means as establishing equipment standards and ratings, prescribing uniform safety requirements, and studying other factors affecting efficiency, production, and costs of heating.

Other companies participating are Allis-Chalmers Mfg. Co., Cutler-Hammer, Inc., Federal Telephone & Radio Corp., General Electric Co., The Girdler Corp., Induction Heating Corp., Lepel High Frequency Laboratories, Inc., RCA Victor Division, and Weltronic Co.

NAB Engineering Conference

ENGINEERING progress in television, f-m, and standard a-m broadcasting will be stressed at the day-long engineering conference that opens the annual convention of the National Association of Broadcasters in Convention Hall, Atlantic City, N. J. Sept. 15-18. The agenda, prepared under the direction of Royal V. Howard, NAB director of engineering, is as follows:

Sept. 15, morning session, presided over by Orrin W. Towner, chairman of NAB engineering executive committee and technical director of WHAS, Louisville, Ky.:

Recent Television Developments, by O. B. Hanson, vice-president and chief engineer of National Broadcasting Co.; emphasis on photography of kinescope images; illustrated with motion pictures.

Frequency Modulation Broadcast Station Construction, by Paul A. de Mars of Raymond M. Wilmotte, Inc.

Audio Considerations for Broadcast Stations, by John D. Colvin, audio facilities engineer at American Broadcasting Co.

Sept. 15, afternoon session, presided over by Royal V. Howard:

Transmitter Maintenance for Small and Medium-Sized Stations — symposium, con-

R. V. Howard

O. W. Towne

O. B. Hanson

P. A. de Mars

(continued on p 241)

Now available

REL **TECHNICAL PURPOSE - DUAL BAND FM RECEIVER**

Here is a custom built, top quality instrument—designed and produced by the leading manufacturer of FM transmitters.

This receiver is proving most useful to broadcasters for: monitoring, relaying, qualitative field strength measurements, checking directive antennas, demonstrations and many other professional uses.

Used in combination with a quality reproducer, this instrument will give the discriminating listener the highest performance in FM reception.

Specifications of Model 646 — illustrated above — includes standard receiver, 115 volt A. C. power supply and 10 watt amplifier in a single cabinet:

Frequency Range: 88 to 108 and 44 to 50 megacycles.

Sensitivity: Receiver limits on input circuit noise at any frequency in the 88 to 108 and 44 to 50 megacycle bands. For all input signals of four microvolts or more applied to the input terminals, the receiver output signal to noise ratio in a band from 50 to 15,000 cycles is within three db of the optimum obtainable at the present state of the art where the limiting noise is random.

Frequency Response: Plus or minus 1 db 30 to 15,000 cycles, including de-emphasis of 75 microsecond time constant.

Noise Suppression: Receiver noise is 70 db or more below output of ten watts for any quieting carrier.

Waveform Distortion: Receiver distortion up through the detector is less than 1% for 100% modulation. The ten watt audio amplifier associated with the Model 646 Receiver has less than 1.5% distortion at full output for any fundamental from 50 to 7500 cycles.

Image Rejection: Better than 45 db.

I.F. Rejection: Better than 65 db.

Power Supply: 115 volts, 60 cycles, single phase. Consumption 125 watts.

Output: 10 watts, 500 and 8 ohms.

Controls: Tuning, Band change switch, Audio frequency gain, Radio frequency gain, Power switch.

Meters: Signal strength.
Tuning indicator.

Dimensions: 19" wide x 16⅝" deep x 10⅜" high.
Weight: 54 lbs. net.

Net price: $340 F.O.B. factory — plus tax. Now in production for early delivery.

Other models available include #647 — designed for schools and other applications where a receiver but no audio amplifier is needed. Incorporates receiver chassis and AC power supply.

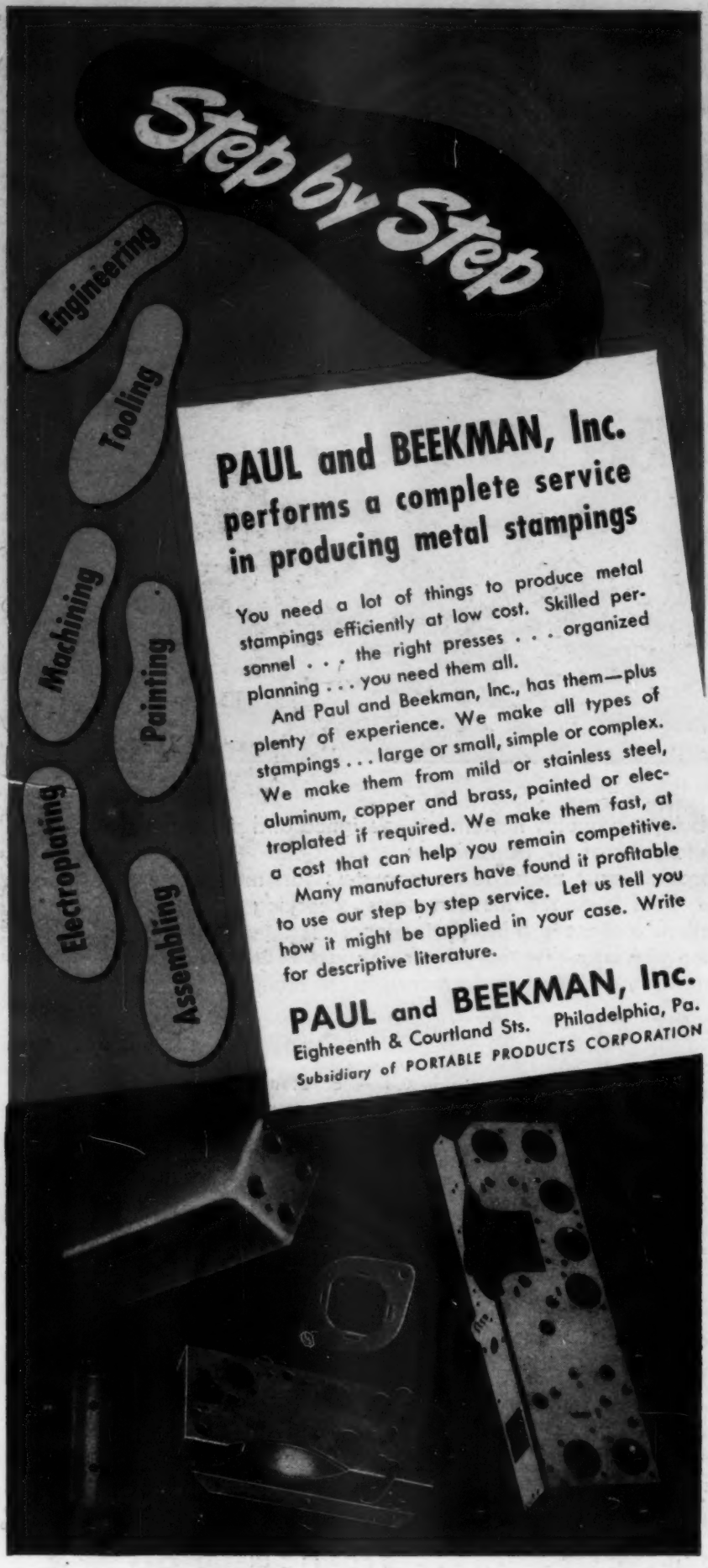
#648 for mobile use — includes receiver chassis, 6 volt DC power supply combined with 4 watt amplifier. Operates from 6 volt storage battery.

Inquiries for descriptive bulletin or further details will be promptly answered.

FM THE ARMSTRONG SYSTEM
LICENSED UNDER ARMSTRONG PATENTS

RADIO ENGINEERING LABORATORIES, INC.

35-54 36th STREET **LONG ISLAND CITY 1, N. Y.**


Step by Step
Engineering
Tooling
Machining
Painting
Electroplating
Assembling
PAUL and BEEKMAN, Inc.
performs a complete service
in producing metal stampings
You need a lot of things to produce metal stampings efficiently at low cost. Skilled personnel . . . the right presses . . . organized planning . . . you need them all.
And Paul and Beekman, Inc., has them—plus plenty of experience. We make all types of stampings . . . large or small, simple or complex. We make them from mild or stainless steel, aluminum, copper and brass, painted or electroplated if required. We make them fast, at a cost that can help you remain competitive.
Many manufacturers have found it profitable to use our step by step service. Let us tell you how it might be applied in your case. Write for descriptive literature.
PAUL and BEEKMAN, Inc.
Eighteenth & Courtland Sts. Philadelphia, Pa.
Subsidiary of PORTABLE PRODUCTS CORPORATION

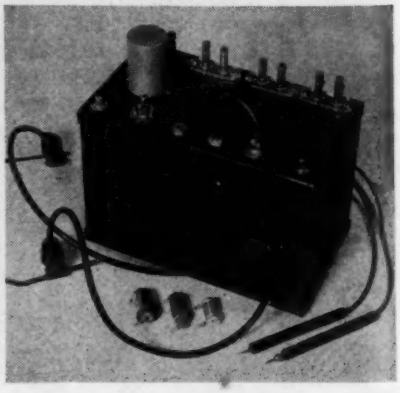
and simply multiplies these ranges mentally by $Y/10$.

For positions 9, 10, and 11 of the range selector switch, 1,000, 100, and 10 ohms respectively are shunted directly across the test-lead binding posts, and the voltage drop across the particular one of these resistors being used is opposed through phones and vibrator by the calibrated voltage drop across the measuring potentiometer. The instrument then acts as a current meter with a range of 1.0, 10, or 100 ma. The 1.0-volt range (position 8) can also be regarded as a 500-μa range.

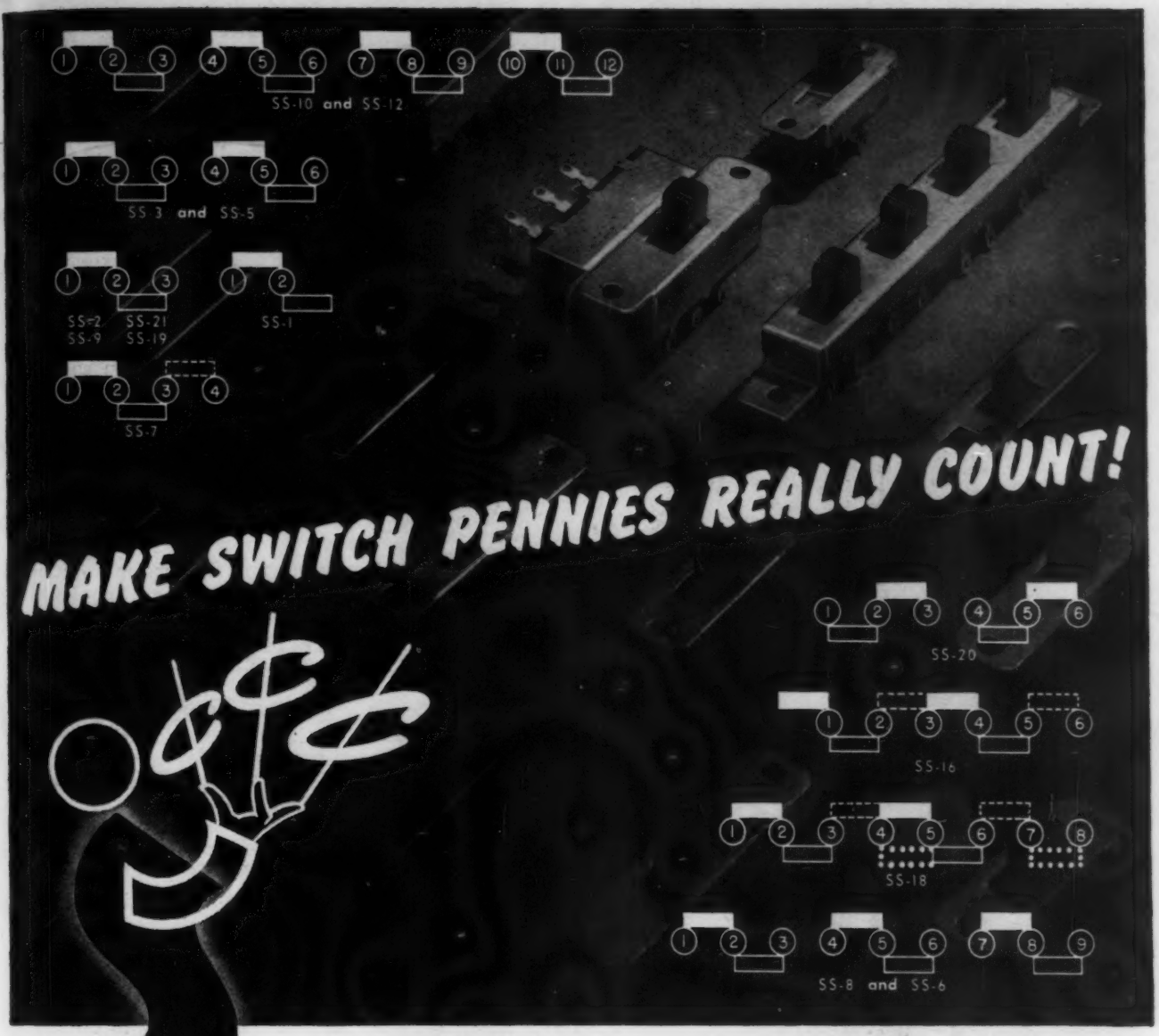
• *Ranges*

The following resistance ranges are obtained for positions 2 through 11 of the range switch respectively: 1,000,000, 400,000, 200,000, 100,000, 20,000, 10,000, 2,000, 1,000, 100, and 10 ohms. The corresponding voltage and current ranges are: 500, 200, 100, 50, 10, 5, and 1 volt; and 1, 10, and 100 ma.

By adding external resistance between binding posts M_1 and M_2, higher ranges of voltage and resistance can be readily obtained. The 0.1 μf capacitor together with switch S_4 and pushbutton S_2, makes it possible to measure quite high resistances provided accurate range multipliers of the same value are available. To make such high resistance measurements, the vibrator is removed from the instrument, and the push button is used as a hand operated interruptor. This gives the capacitor time to charge between successive applications of the push button, and when



Complete instrument that permits a blind person to measure resistance, voltage, and current



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CONTACT CODE

POSITION 1
POSITION 2
POSITION 3
POSITION 4

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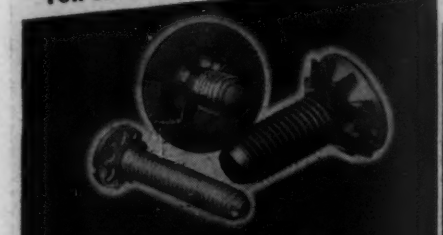

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the latter is depressed, the charge produces a click in the phones. These clicks disappear when the measuring potentiometer is rotated to the point of balance. The vibrator should also be removed during all a-c measurements.

The tolerance for the resistors used is one percent. Experience has shown that a raised scale of small dots in which division points are placed only at intervals of 5 percent of full scale, can easily be read by touch to an accuracy of better than 1 percent. If the distance between division points is between approximately 3/16 and 3/8 inch, these dimensions are commensurate with the dimensions of the very sensitive area on the tip of the index finger. Under the worst conditions, readings should not be subject to an error greater than 2 percent of full scale. Usually the errors observed have been less than 1 percent.

Cathode-Coupled Half-Shot Multivibrator

By Robert K. F. Seal
New York, N. Y.

In connection with the construction of synchronization equipment for the Stanford University multi-frequency ionosphere recorder, the author developed a frequency divider which is simple in construction, dependable in operation with either continuous, intermittent series or individual actuating pulses, and which has already proved valuable in several other applications. Due to its original application as a pulse-frequency halver it was referred to as a pulse-halver. However, it is more broadly referred to as a half-shot multivibrator, from the fact that it is de-

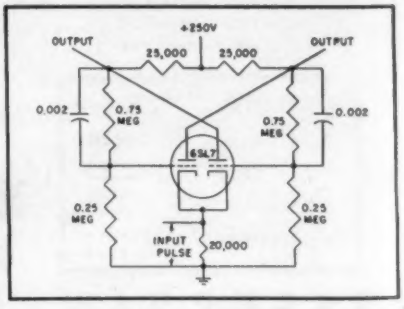
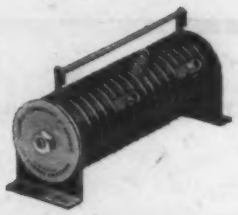


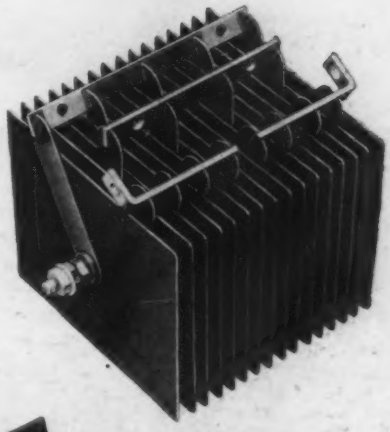
FIG. 1—Circuit of the half-shot multivibrator

This recently completed modern structure, providing more than 32,000 square feet of floor space, is our new home.

Production volume has been considerably increased by the installation of latest equipment for highly specialized operations. Engineering, inspection and testing facilities have been greatly expanded to insure excellence of products with the maximum of efficiency.

These greatly expanded overall plant facilities, plus the recognized dependability of *S C A* products, make it possible for us to offer the most complete line of Selenium Rectifiers and self-generating Photoelectric Cells.





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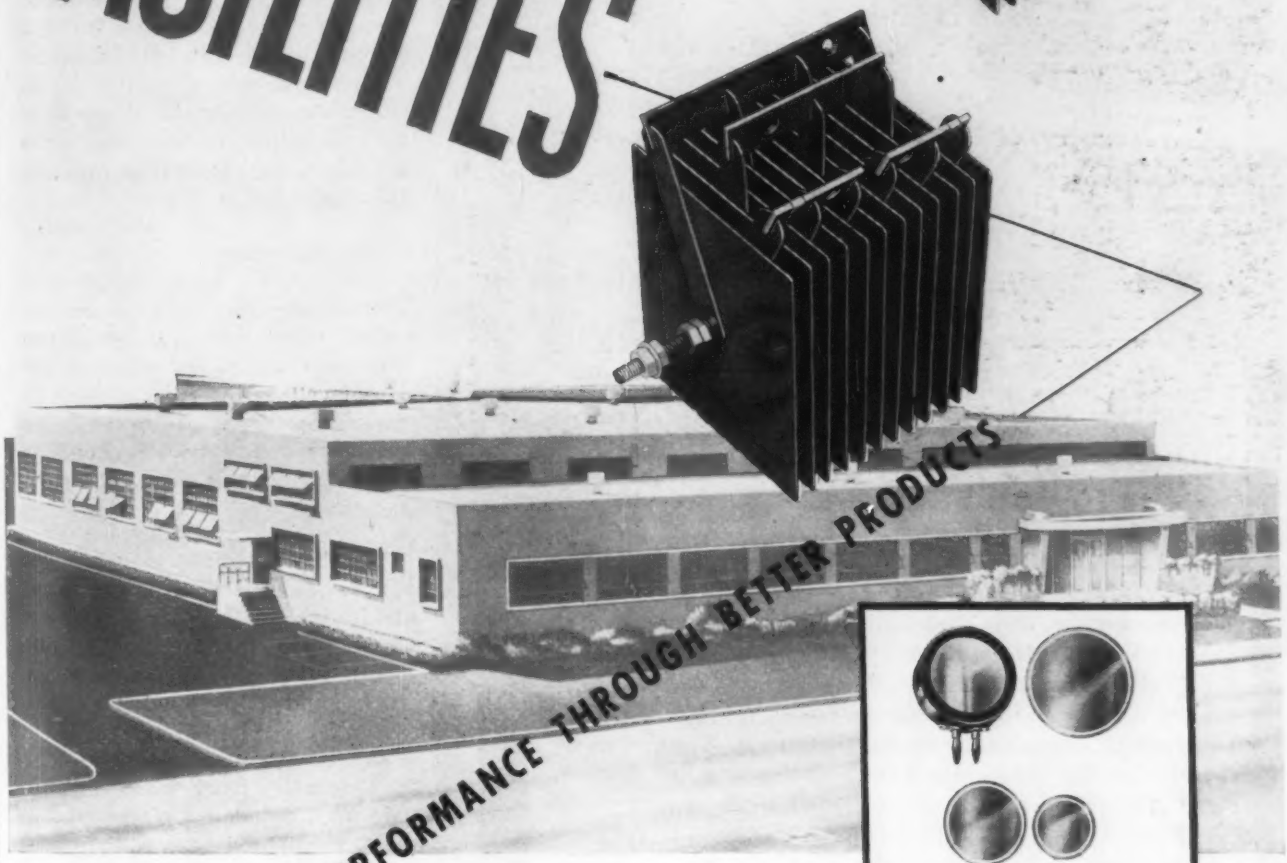

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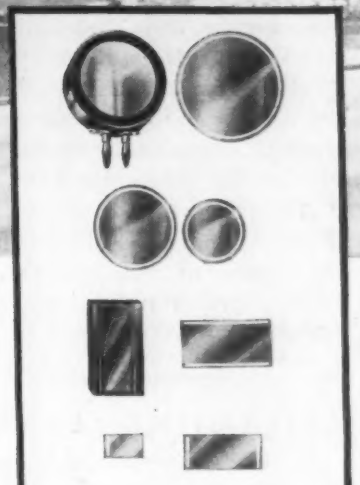

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Descriptive Bulletin No. 10 Available

BALLANTINE LABORATORIES, INC.

BOONTON, NEW JERSEY, U.S.A.

veloped from the basic Eccles-Jordan multivibrator circuit and that one-half an output cycle is obtained for each complete input cycle.

The circuit developed by the author is simple in construction and does not require a large negative voltage from a separate power supply. It operates with any plate supply voltage from 125 to 325 volts and the cathode potential will not exceed eighty volts with respect to ground or heater.

Uses

One application of this circuit is for frequency division, where any frequency (from practically zero on up into the lower r-f range) may be divided by any power of two simply by using the required number of stages in series, the output of each stage being applied to the input of the next through a simple R-C differentiating circuit. By means of a single-pole selector switch a versatile circuit may be constructed to give either rectangular waves or pulses, by connecting the taps either ahead of or following the differentiators, or both, at a number of different frequencies.

The half-shot multivibrator may also be employed as an electronic switch, which operates each of two external circuits by means of the output voltage of either or both of the plates. For this purpose the circuit supplies two rectangular waves 180 degrees out of phase.

Circuit Operation

The basic circuit is shown in Fig. 1. This circuit is symmetrical and is stable with either triode conducting, and the other one cut off. This means that it is possible for a single input pulse to actuate it for one-half cycle.

As this circuit is activated only by a negative pulse whose magnitude is limited to a reasonable value, it may be triggered by a differentiated rectangular wave, each cycle of which gives a single operating pulse and thus one-half of an output rectangular wave cycle. In this way, two input rectangular wave cycles (pulses) give a single output rectangular wave cycle. If alternate negative input pulses are equally spaced, the output will be a square wave, but by controlling the spacing of the input pulses, rec-

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THE HART MANUFACTURING COMPANY
202 BARTHOLOMEW AVENUE, HARTFORD, CONN.

tangular waves of various proportions may be obtained from the output.

Action With Positive Pulse

In either stable condition, one triode is conducting and the other is cut-off. If a positive pulse is applied to the cathodes the conducting tube will tend to conduct less current. However, this tendency is counteracted by the fact that any decrease in this grid-cathode potential causes less grid current to flow, and the drop in the cathode, the grid-to-plate, and the plate-to-plate supply resistors causes the grid voltage to return to nearly zero with respect to the cathode.

As the cathode and grid go positive as a unit, the plate does likewise, and due to the coupling capacitor the grid of the cut-off tube is forced to go more positive. However, the grid-cathode potential of the cut-off tube remains about the same and this tube is held beyond cut-off even with a positive pulse on the cathode. Therefore the circuit is not actuated by positive pulses as long as the coupling capacitor's discharge time constant is large compared to the pulse width, and the pulse is of reasonable magnitude.

With a positive input pulse, a very small positive pip is visible on the plate waveform during both the cutoff and conducting portions, since the drop in grid current of the conducting tube also causes the latter's plate voltage to rise. These pips are visible only with large vertical amplification of the plate voltage waveforms.

Effect of Negative Pulse

If a negative pulse is applied to the cathodes, the tube that is conducting will continue to do so, but the other tube will start to conduct, causing its plate voltage to be decreased. Due to the coupling capacitor, the grid potential of the conducting tube will be slightly decreased also, so that it starts to cut off, causing its plate voltage to rise, which in turn raises the grid-cathode potential of the formerly conducting tube still more through the other coupling capacitor.

This action is cumulative and instantaneous, amounting to a complete switching of tubes. Another

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Smooth, cool-running operation is assured by an efficient lubricating system and *two* cooling fans.

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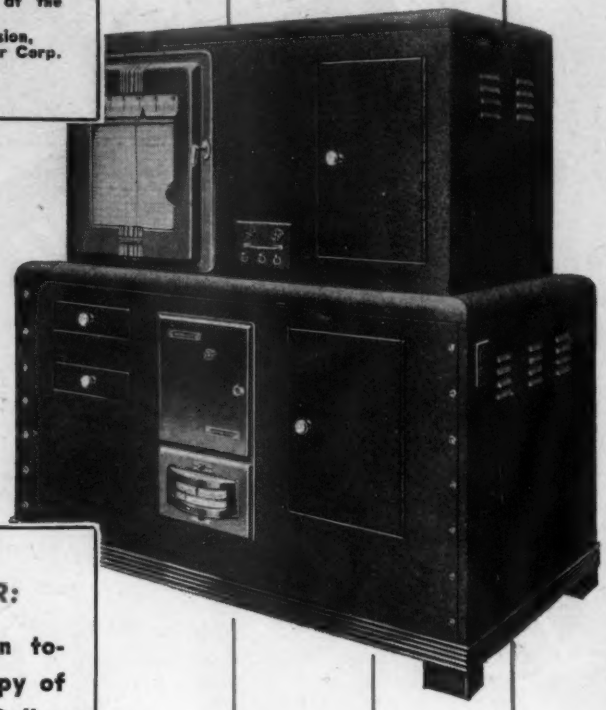

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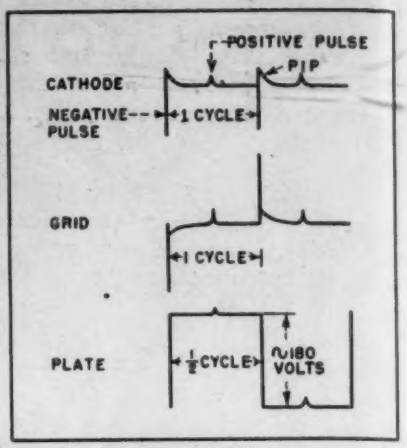

FIG. 2—Waveforms at elements of one triode. The pip at the cathode is due to charging effects of an isolating capacitor in the input pulse circuit

negative pulse applied to the cathodes will switch the tubes back to their original condition, in a similar manner, and repeated pulses result in rectangular waveforms on each plate of one-half the frequency of the negative triggering pulse. The output of either or both plates may be used, as desired.

The resistance values are calculated with the aid of the tube characteristic curves to give a steady-state condition which can be easily switched by the available cathode pulse so that when one tube has zero grid-cathode potential the other will be held just beyond cut-off.

The limitations on the operating frequency range are due to the charging and discharging of the coupling capacitors. Above the upper frequency limit the circuit acts like a conventional frequency divider whose operation depends on the charging rate of a capacitor, and thus its factor of frequency division is a function of the input pulse width and R-C time constants.

For normal operation, the size of the coupling capacitors is not critical. The capacitance must be great enough so that little charging and discharging can take place while the pulse is applied to the cathodes. Otherwise the circuit becomes unstable and does not know which way to go because each triode will be conducting an equal amount of current and it will be a matter of chance which tube gains control and causes the other tube to be cut off. The capacitors must be small enough so that they can

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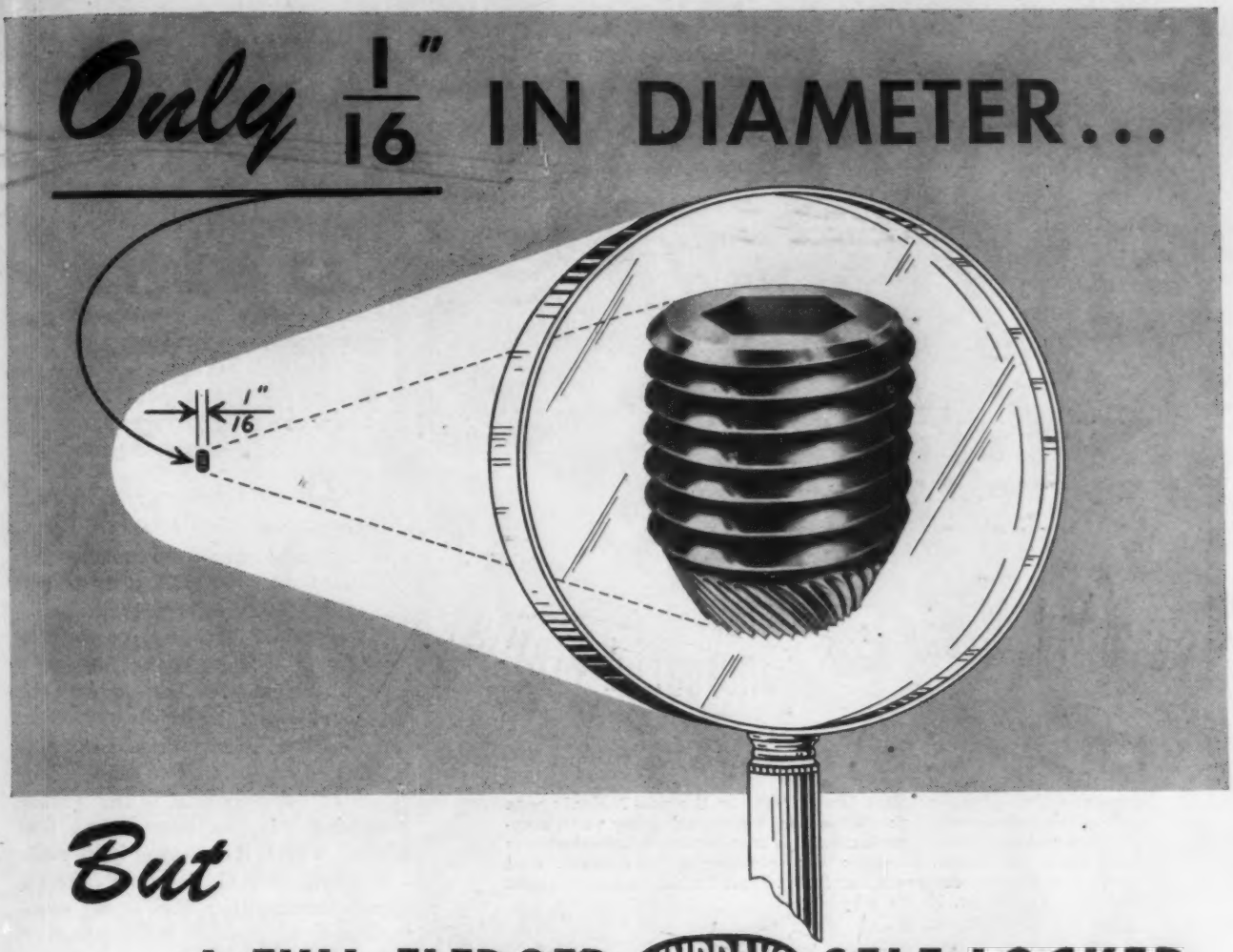

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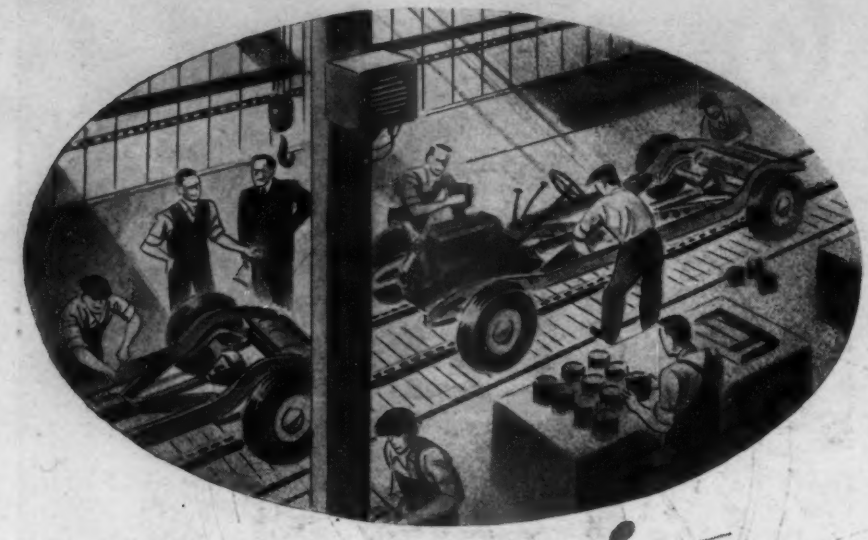

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Of principal interest are their excellent electrical characteristics, extreme simplicity of design to avoid trouble, and the "hum-bucking" characteristics, which reduce the effect of extraneous magnetic fields. When required, the head cartridge alone (pole piece and coil unit) may be supplied for incorporation into manufacturers' own head structure.

These latest developments in magnetic recording equipment can now be obtained for radio combinations and other uses. Brush engineers are ready to assist you in your particular use of magnetic recording components.

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charge and discharge sufficiently between pulses to be prepared for the next pulse. Fig. 2 shows the waveforms of the circuit when actuated by a 60-cycle input pulse.

Tubes Critical

Either a high-mu (6SL7) or a medium-mu (6SN7) twin triode may be used. The wider the frequency range over which a particular circuit is to operate, the more constant the plate supply voltage must be, and for best results in such applications a regulated power supply is indicated.

Particular tubes of the same type will not always operate in this circuit, and sometimes it is not possible to obtain operation by adjustment of the resistor network alone. However, the circuit can always be made to work by connecting small external capacitances across the tube elements by trial and error. Adjustment of circuit resistance or external capacitance to make a particular tube operate is not recommended. It has been found that when a circuit has once been made to work properly, and no external capacitance other than stray capacitance is present, approximately 60 percent of the dual triodes selected at random will operate. The use of individual triodes may be preferred to facilitate obtaining properly matched pairs of triodes.

The very large rectangular wave output is capable of producing so-called r-f burns, even with a repetition frequency as low as 15 cps.

C-R Tube on Power Lines

McGraw-Hill World News

USE of the cathode-ray tube as a power-frequency synchroscope has been successfully developed and extensively tested by the State Electricity Commission of Victoria, Australia.

To preserve continuity of power supply, synchronizing facilities are required at each switching station, so that reconnections can be made at points of circuit interruption without de-energizing sections of line back to the generating station.

Dispensing with the large number of potential transformers which must be provided for connection on

Capable of realizing the full resources of the new professional FM tuners

. . . and linking their superior performance to the industry's first-choice speaker*

The final link in the chain of high-quality reproduction has now been forged: a power amplifier which brings out the final degree of excellence of performance of the new professional-quality tuners, and of the two-way multicellular speaker which has become the standard of the broadcasting, recording, and motion picture industries, the Altec Lansing Duplex speaker.

Salient characteristics of the New Altec Lansing A-323B Amplifier are:

1. Flat frequency response—within 1 db from 20 cycles to 20,000 cycles.
2. Full 15 watt power output within 1db from 35 cycles to 12,000 cycles.
3. Two inputs. One low gain from radio tuner, and one high gain with built-in equalization to operate direct from the new General Electric Variable Reluctance or the new Pickering Cartridge Pickup.
4. For record reproduction, stepped low pass filter which gives sharp cut-off of noise frequencies yet allows full reproduction of useable high frequencies
5. Adjustable base boost.
6. Hum balancing potentiometer to eliminate the necessity of careful selecting of tubes for quiet operation.

The Altec Lansing A-323B Amplifier was designed with a particular view to its use in high-quality music reproduction systems in which the electrical elements—tuner, amplifier, record player, and speaker—are either wholly concealed in the interior structure of the room, or partly concealed in furniture already in harmony with the interior scheme.

The noteworthy trend, among fastidious, musical-minded home lovers, toward built-in music systems which eliminate non-functional radio "furniture"—and thus eliminate the inevitable problem of assimilating "big sets" into modern decorator-influenced interiors, is becoming increasingly apparent; this trend highlights the importance of the new Altec Lansing amplifier and the Duplex multicellular speaker in the advanced planning in the FM field.

*The Altec Lansing Duplex Loudspeaker

The Altec Lansing loudspeaker is the only speaker which in actual performance has been found capable of reproducing the full frequency and dynamic ranges, and without distortion, all the characteristics of sound of which the frequency modulation method of radio transmission is capable. The high efficiency, and the resultant damping, causes the Altec Lansing Duplex loudspeaker to follow accurately every variation in the signal wave form, without contributing the spurious sounds that result from intermodulation and poor transit characteristics common to ordinary speakers.

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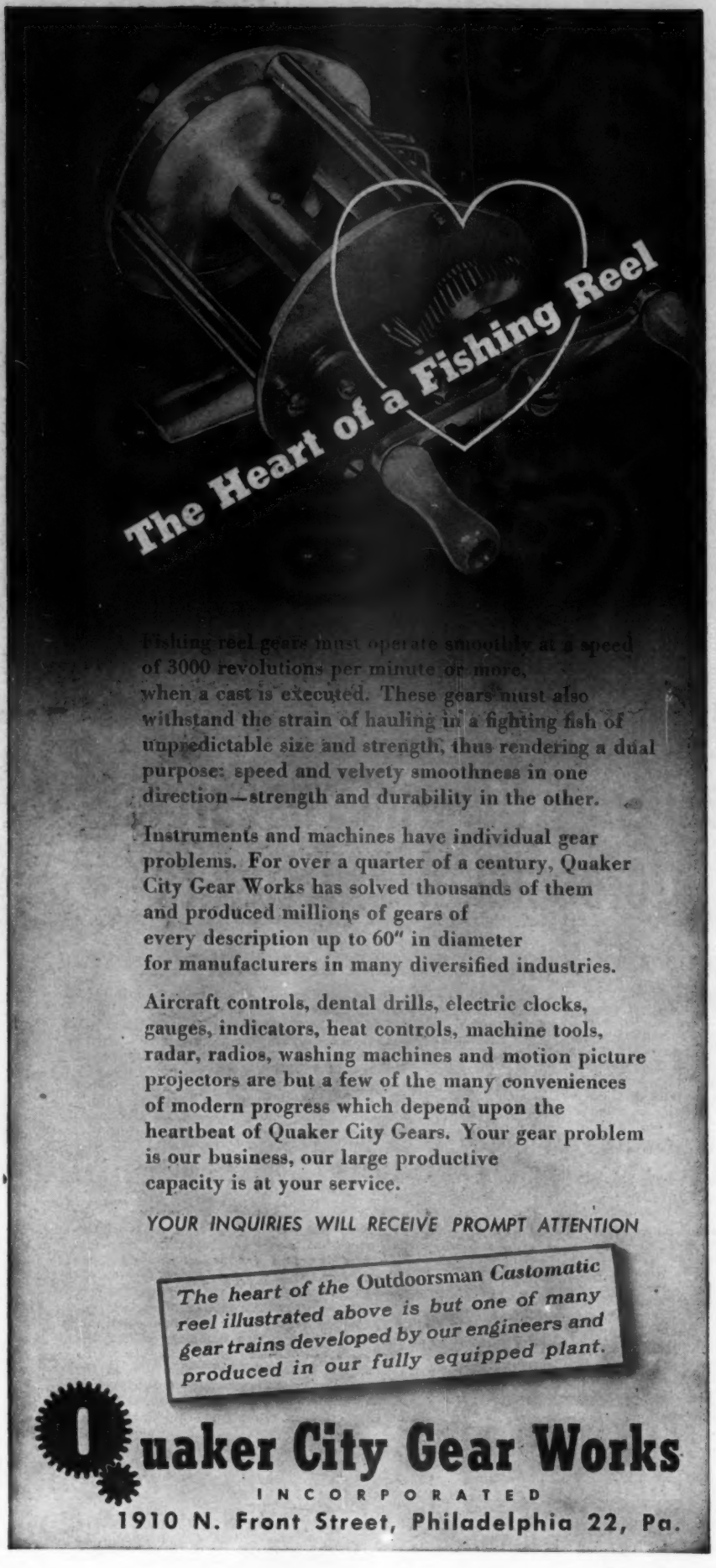

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1910 N. Front Street, Philadelphia 22, Pa.

each side of each circuit breaker across which synchronizing indications are required, use is made of the electrostatic potential in aerial conductors suspended adjacent to transmission lines. While these have been used to indicate whether the lines are energized, the electronic synchroscope relies on them for reflection of the phase relation of the potentials of the transmission lines.

Coax Coil for Heating Small Holes

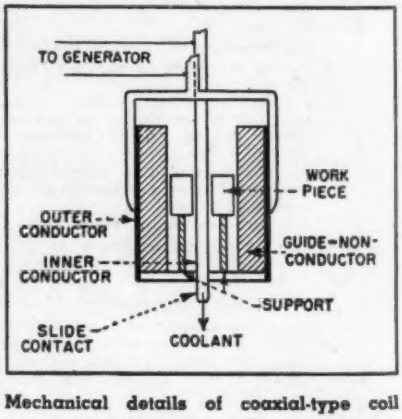
By David Baumoel
Electronic Process Heat Treating Co.
New York, N. Y.

THE APPLICATION of induction heating to a small diameter hole, roughly between $\frac{1}{4}$ and $\frac{3}{8}$ inch, has presented unusual difficulties due to the almost impossible task of building a work coil small enough and efficient enough to fit into the confined quarters and properly do its job.

A means toward overcoming this problem is found in abandoning the usual multiturn type of coil, in conventional use for heating of larger diameter holes, and using instead, a coaxial-type coil.

The accompanying sketch shows the basic design of the coil. In this coil the inner conductor acts as the energy transfer area. Ordinary copper tubing, water or otherwise cooled, is used, whose diameter is determined by the dimensions of the hole in the metal part to be treated.

The inner conductor is run through the hole to be heated and connected to the rest of the coil



Mechanical details of coaxial-type coil jig for internal heating. The shaded portions are ceramic or other insulating materials

Gates Radio Company
QUINCY, ILLINOIS, U. S. A.

Electronic Equipment
June 10, 1947

American Lava Corporation,
Chattanooga 5, Tennessee.

Gentlemen:

We would like to report to you the excellent performance and electrical characteristics we are obtaining in the field with your part E-16898 transmission line insulator. You will recall that after we covered the field quite thoroughly about two years ago we decided to use this particular design and insulator because of superior performance that we expected.

This particular insulator is used as the center support on broadcast frequency tranamission lines. At times these insulators are subject to considerable electrical and mechanical overload due to lightning hits, ice formations and a number of other unpredictable factors. To this date we have not had one single field failure and the performance both electrically and mechanically has been up to and exceeding expectations.

We wish to thank your company for the close cooperation in the design of such an outstanding electrically and mechanically product and assure you that on future developments of this type we will first request your valued engineering and production assistance.

Yours very truly,

GATES RADIO COMPANY

Fred O. Grimwood

Fred O. Grimwood, Ch. Engr.

FOG/nn

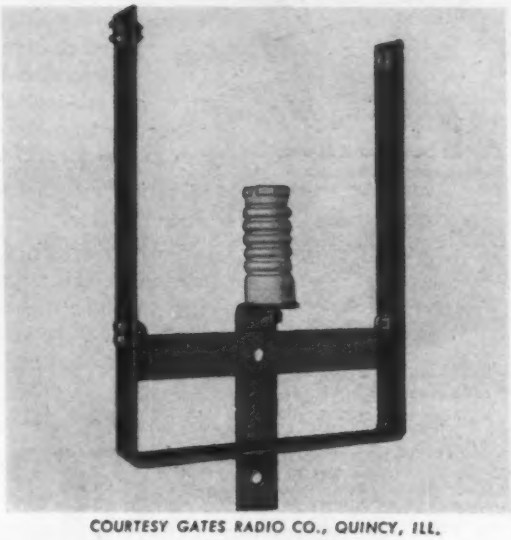
ALSIMAG TRANSMISSION LINE INSULATORS WITH METAL STUDS

Low loss AlSiMag Insulators with metal studs solve many installation problems . . . permit a variety of transmission line characteristics.

The Feed Through insulator illustrated is mounted in the center of an open wire transmission line bracket by means of a 2" metal stud sunk 1" into the insulator. This stud is accurately placed and is engineered to be a strong joint resistant to all types of weather.

The letter reproduced above outlines the experience of one user of this type of AlSiMag steatite insulator.

The physical characteristics of the principal AlSiMag technical ceramic compositions are shown on the AlSiMag Property Chart, sent free on request. Our engineers will gladly cooperate with you in developing the design most advantageous to your requirements.



COURTESY GATES RADIO CO., QUINCY, ILL.

ALSIMAG

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AMERICAN LAVA CORPORATION

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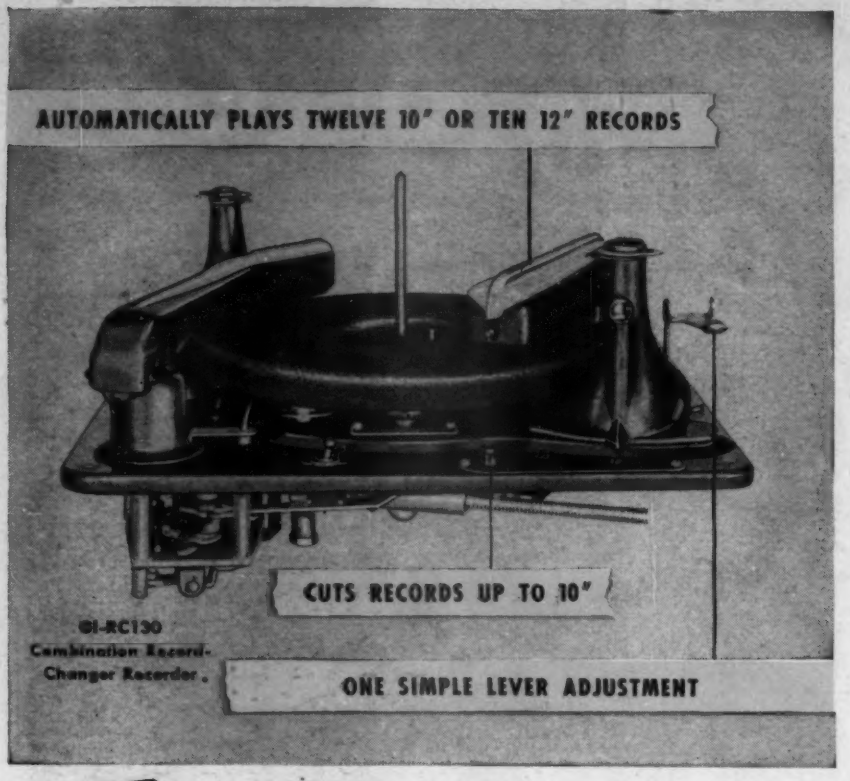

AUTOMATICALLY PLAYS TWELVE 10" OR TEN 12" RECORDS
CUTS RECORDS UP TO 10"
GI-RC130
Combination Record-Changer Recorder.
ONE SIMPLE LEVER ADJUSTMENT

Smooth Power COMBINATION
RECORD-CHANGER • RECORDER

GI-RM4 Rim Drive, Heavy-Duty Electric Recording Motor

GI-R90 Dual-Speed, Home Recording and Phonograph Assembly

LX Rim Drive, Constant-Speed Electric Phonograph Motor

Your customers will like the simplicity and fine performance of this unique combination *Smooth Power* unit.

They'll enjoy the ease of operation with one simple lever for quick changing from one size record to another, to remove records or to set for manual operation and recording. They'll appreciate the smoothness and quietness of the record-changer. They'll admire the brown iridescent finish and streamline plastic trim on self-indicating "Reproducer" and "Recorder" arms.

And, of course, they'll value the quiet, vibration-free operation of the *Smooth Power* Motor.

Send for details. Ask us for complete information on this popularity-building combination that can add new sales appeal to your products . . . and on the complete line of *Smooth Power* Phonomotors and Recorders.

THE GENERAL INDUSTRIES CO.

DEPT. ME • ELYRIA, OHIO

which acts as the outer conductor of the coaxial coil. Since the transmitted energy is of much higher density around the inner conductor, the area in the small hole of the metal part will heat at a much faster rate than the rest of the part.

The part can be drop or spray quenched for hardening purposes, as for example in the heat treating of dies, or merely removed and allowed to cool, in other applications such as brazing, soldering, or annealling. Many mechanical arrangements are possible for feeding work pieces into the jig. For example, the one illustrated might use a movable bottom plate that would contact the outer and inner conductor when the work is inserted. Shutoff of the flow of coolant could be controlled by a valve during loading, or continuous flow could be permitted in some applications.

Compact Power Supply for High Current

BY LAWRENCE FLEMING
Naval Ordnance Laboratory
Washington, D. C.

THE POWER SUPPLY shown in Fig. 1 delivers 235 volts d-c at one ampere and permits an unusually large ratio of power output to physical size by employing a three-phase rectifier circuit[1].

This unit is a convenient source of plate power for large numbers of receiving tubes in multichannel oscillograph amplifiers and similar installations. The single-phase sup-

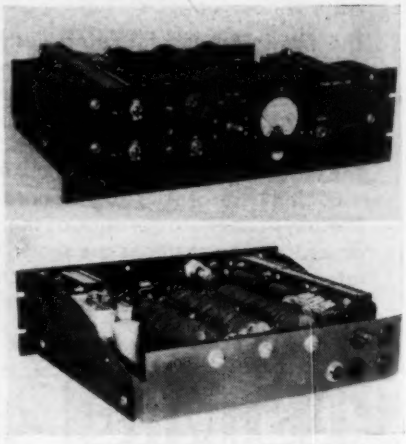


FIG. 1—Front and rear views of the high-current power supply

ROUND

MAXIMUM OD: .100"

Cathode shown—.100" OD x 36 mm long—single beaded.

MINIMUM OD: .040"

Cathode shown—.040" OD x 27 mm long—single beaded.

PLAIN
(without bead or emboss)

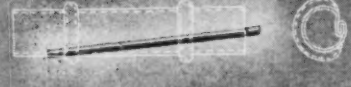
Cathode shown—.081" OD x 25 mm long—no bead.

SINGLE BEADING (any position, .040" min. from end)

Cathode shown—.045" OD x 28 mm long—beaded 10.25 mm from one end.

DOUBLE BEADING (same distance from each end)

Cathode shown—.045" OD x 28.5 mm long, beaded .118" from each end.



DOUBLE BEADING (different distances from each end)

Cathode shown—.040" OD x 29.5 mm long, double beaded .039" from one end, distance between beads .699"

ROUND

MINIMUM LENGTH
(Round) — 11.5 mm

MAXIMUM LENGTH
(Round) — 42 mm

Cathode shown—.045" OD x 11.5 mm long, single beaded.

REFERENCE DATA
on All Types of
SUPERIOR LOCKSEAM* CATHODE SLEEVES

ROUND AND OVAL—
Plain or Beaded (horizontal & vertical)
Showing maximum and minimum dimensions, bead position, etc.

BASIC DIMENSIONAL DATA

Maximum OD —.100"
Minimum OD —.040"
Minimum length —11.5 mm
Maximum length — 42 mm

STANDARD WALL THICKNESSES:

.0021" ± .0001"
.0025" ± .0001"

BEAD LOCATION: .040" min.

BEAD DIAMETER: Depends on OD. Write for Spec. EM-2

BEAD WIDTH: .017" + .003" − .002"

For tolerance data write for Specification EM-2.

For information on cathode materials write for Data Memo #5.

*Superior Tube Company Patents

VERTICALLY EMBOSSED

This new Lockseam Cathode has wide acceptance for use in certain types of special radio tubes.

Cathode shown—.045" OD x 26.5 mm long—vertically beaded.

FLAT OVAL

BEADING ON MAJOR ONLY

Cathode shown—.035" x .060" OD x 29 mm long—single bead on two sides only (major).

BEADING ON BOTH MAJOR AND MINOR

Cathode shown—.045" x .155" OD x 33 mm long—single beaded on three sides —2 on minor and 1 on major.

BEADING ON MINOR ONLY

Cathode shown—.034" x .084" OD x 28.5 mm long—beaded on minor only.

•DOUBLE BEADING

Cathode shown—.060" x .124" OD x 30 mm long—double beaded on major only.

MINIMUM LENGTH
(OVAL) — 11.5 mm

MAXIMUM LENGTH
(OVAL) — 42 mm

Cathode shown—.034" x .084" x 42 mm long—beaded one end on major only.

THE BIGGER NAME IN SMALL TUBING

Superior

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Coating of front surface mirrors—another wonder product of low pressure processing—is done automatically in this National Research Corporation installation using a KINNEY Vacuum Pump. The KINNEY Single Stage Vacuum Pumps, used in conjunction with diffusion pumps, maintain the low absolute pressures essential for this high vacuum evaporation process. In all parts of the world the uses of KINNEY Pumps are countless—from dehydrating foods to producing penicillin; from exhausting lamps and tubes to sintering alloy metals. Wherever low absolute pressures must be created *and maintained*, KINNEY Pumps are giving reliable service. KINNEY Single Stage Pumps, available in 8 sizes with displacements of 13 to 702 cu. ft. per min., maintain low absolute pressures to 10 microns; Compound Pumps in 2 sizes, 15 and 46 cu. ft. per min., maintain low pressures to 0.5 micron.

KINNEY
Single Stage
Vacuum Pump

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WE ALSO MANUFACTURE LIQUID PUMPS, CLUTCHES AND BITUMINOUS DISTRIBUTORS

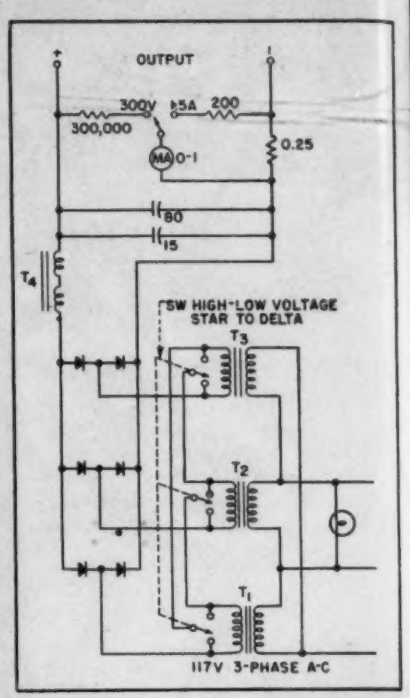


FIG. 2—The selenium rectifiers are Federal 5D16AV1. Choke T_4 is an isolation transformer, Stancor P-6160, with primary and secondary in series. The other transformers are the same type, used conventionally

ply which would ordinarily be considered for such applications would not only be much larger but would require special transformers and chokes.

Ripple output of the unit at full load is approximately 20 millivolts at 360 cycles. Filtering is not difficult because the three-phase full-wave rectifier circuit employed produces an output before filtering which contains only 4.2 percent ripple at 360 cycles.

A switch is arranged to connect the secondaries of the three transformers in either star or delta connection. The former connection gives 235 volts at full load, the latter 120 volts.

If the unit is to be operated continuously, adequate provision must be made for ventilation of the selenium rectifier stacks, since load-carrying capacity of the stacks is primarily limited by temperature.

REFERENCE

(1) Polyphase Rectification, *Proc. I.R.E.*, 19, Jan. 1931, p 78

Hardening Shear Blades

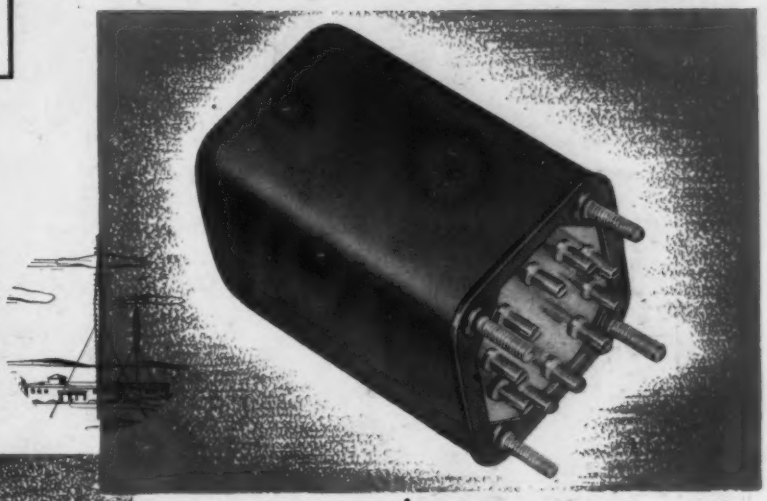
LESS THAN two seconds is required to harden shear blades by induction heating to a higher degree than

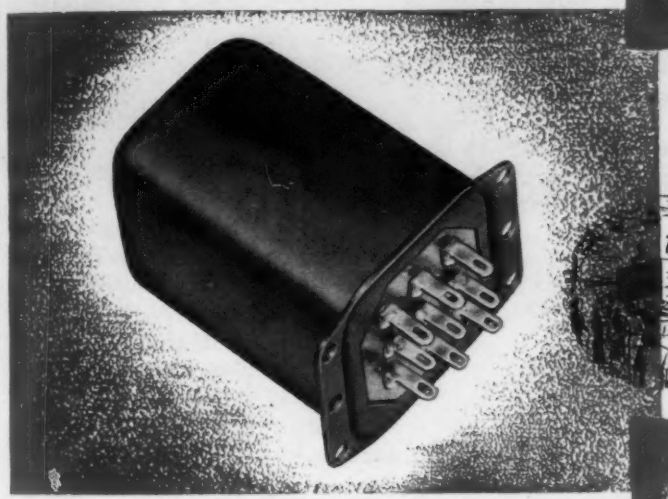
AUDIO TRANSFORMERS *for* [UNIFORM RESPONSE LOW DISTORTION IN 3 FREQUENCY RANGES

Write for Catalog showing complete new stock line

Full Frequency Range

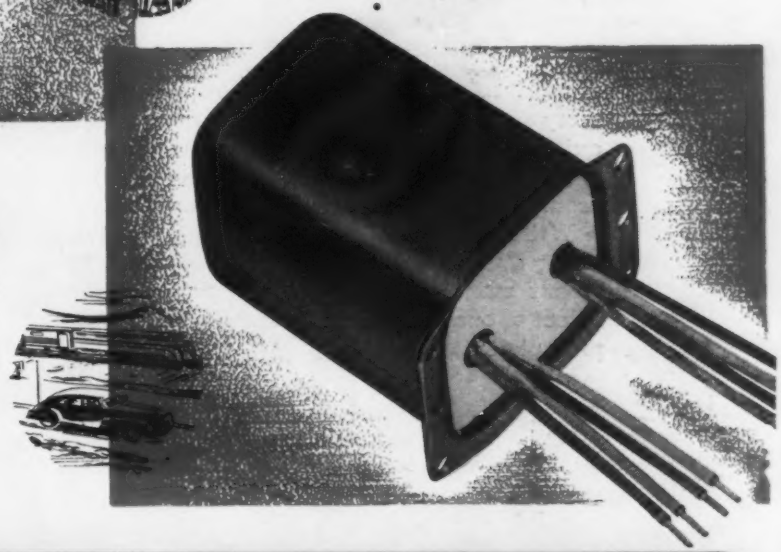
30 to 15,000 Cycles, provides uniform response over this entire band with ± ½ db up to 10 watts of audio power, within ± 1 db over 10 watts. Standard RMA impedances. Hum balancing coil structures and nickel alloy shielding. Included are Input, Output, Driver, and Modulation Transformers; Modulation Reactors. *Sealed in Steel* construction, stud mounting, with pin-type terminals.





Public Address Range

50 to 10,000 Cycles, frequency response within ± ½ db up to 10 watts of power, within ± 1db over 10 watts, throughout this range. Secondary impedances match 600 and 150-ohm lines, 16, 8 and 4-ohm reproducing systems. Listed are Driver and Output Transformers. *Sealed in Steel* construction, flange mounting, with solder lugs or wire leads.



Communications Range

200 to 3,500 Cycles, affords response with variations not exceeding ± 1 db over the range of voice frequencies. For use with 600 or 150-ohm lines. Input, Output, Driver and Modulation Transformers offered. *Sealed in Steel* construction, flange mounting, with wire leads or solder lugs.

CHICAGO TRANSFORMER

DIVISION OF ESSEX WIRE CORPORATION

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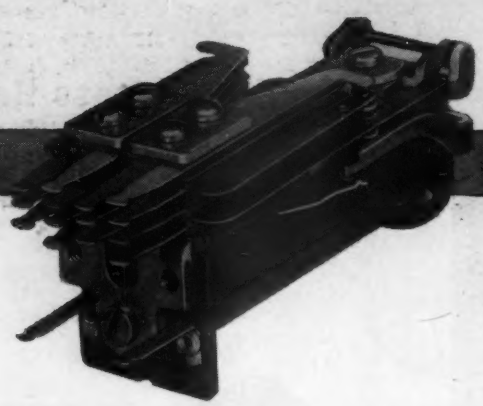
AMRECON
Telephone Type
RELAYS



Amrecon's technical knowledge of practical relay application simplifies your own control requirement problems. Our new modern plant is now in full operation and we can offer you prompt delivery and even greater service.

Write for our informative illustrated Bulletin listing our many products, describing the scope of our manufacturing facilities, and the wide extent of our services. Your inquiry will receive prompt attention.

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Shear blades are hardened in this jig of the 20-kw induction-heating generator

government specifications for this hardness, Rockwell C55 to 60. Such shears of definitely better quality, with blades electronically hardened on the cutting edges, are now being produced by Acme Shear Co. of Bridgeport, Conn. Blades are a solid forging of carbon steel, edge-hardened to the high limit, and the back of the blade left malleable to prevent breaking.

In bending tests, the electronic heat-treated blade with the hardest cutting edge withstood twice the bending pressure that fractured blades made by other methods.

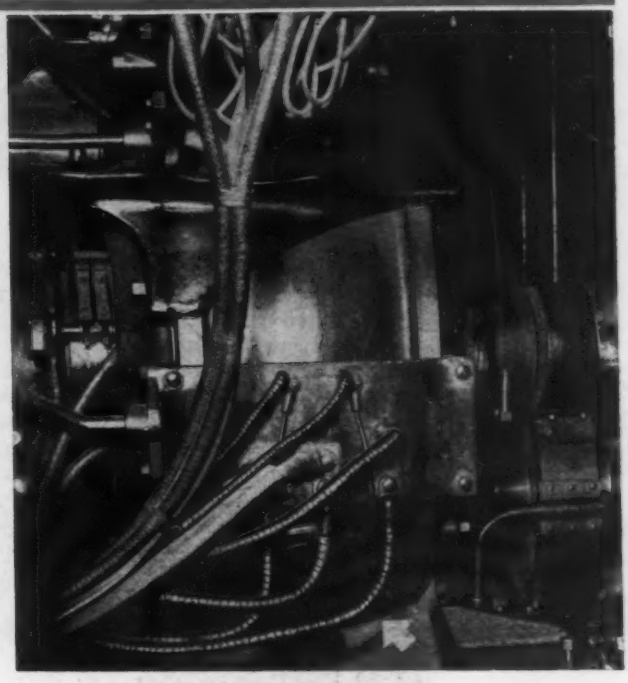
Further savings accomplished as a result of the use of high frequency heating include elimination of other operations due to the fact that with induction heat treating, control is so exact that the heat-treat operation can be done as one of the later operations. Thus, time is saved by doing certain cuts while the steel is soft, rather than after the hardening operation. The capacity of the Acme generator is 20 kw.

River-Mapping by Radar

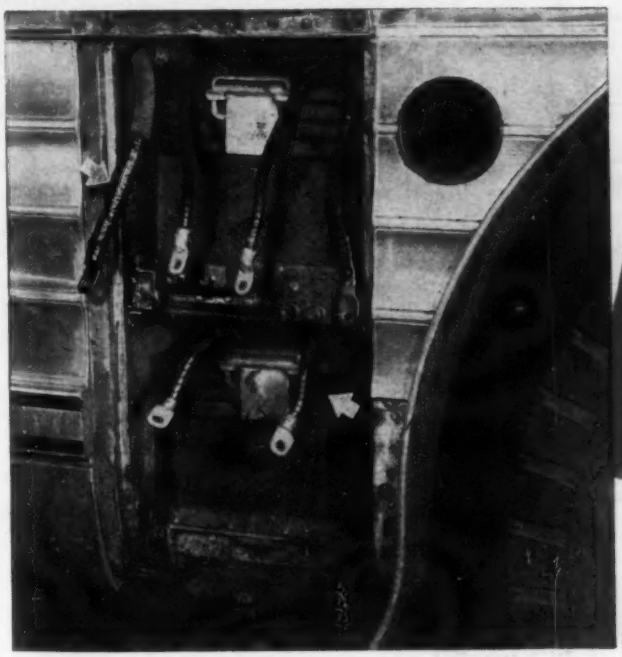
RIVER navigation charts are being made from radar photographs by the Army Corps of Engineers, Ohio River Division, using 3.2 centimeter radar equipment manufac-

FOR THE MOST EXACTING ELECTRICAL INSTALLATIONS

AUTOMOTIVE IGNITION HARNESSES. Applied with a Segur taper adapted to the work with the cooperation of the National Scientific Products Co. Applied in this manner "SCOTCH" Electrical Tape with Vinyl Plastic Backing makes a securely bound but flexible harness; high resistance to oil, water, alcohol, heat, abrasion; high tensile strength.



MACHINE TOOLS, PRINTING PRESSES. Where power wiring is subjected to oil, acids, etc., "SCOTCH" Electrical Tape with Vinyl Plastic Backing provides dependable resistance to oil; water, fresh or salt; acids; alkalis; alcohol; hydrocarbons.



PROTECTIVE SHEATH FOR BATTERY CABLES. Above picture shows heavy battery cables protected with a tight spiral-wound sheath of "SCOTCH" Electrical Tape with Vinyl Plastic Backing. The acid and oil resistant qualities of this tape equip it to give fullest electrical and mechanical protection.

REG. U. S. PAT. OFF.

SCOTCH BRAND *Electrical* TAPE

WITH VINYL PLASTIC BACKING

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"SCOTCH" is the registered trademark for the more than 100 varieties of adhesive tapes made in U. S. A. by

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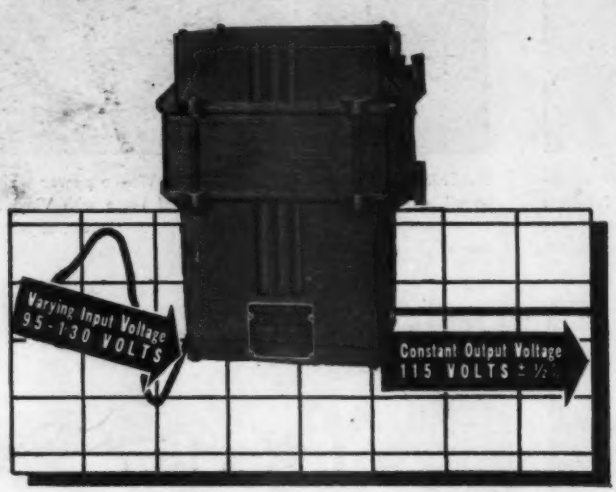
RAYTHEON Voltage Stabilizers

*Varying Power Supply need *not* prevent efficient, reliable and accurate performance of electrically operated equipment. A Raytheon Voltage Stabilizer of correct characteristics and capacity built into new equipment, or installed on old, provides constant output from varying input. Assure reliable and accurate performance of *your* equipment through *positive control* of power supply. Ask about Raytheon Voltage Stabilizers to meet *your* requirements.

Write for illustrated Bulletin DL-47-537



- OUTPUT VOLTAGE CONTROLLED to within ± ½%.
- STABILIZATION AT ANY LOAD within rated capacities.
- QUICK RESPONSE... varying input voltage stabilized within 1/20 second.
- ENTIRELY AUTOMATIC... no adjustments, no moving parts, no maintenance.

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tured by Radiomarine Corporation of America.

An automatic camera is attached to the radar instrument and adjusted to photograph images on the scope at regular intervals. The prints are then matched to produce a continuous strip-map reduced to the scale of standard navigation charts. On the assembled charts, engineers add dotted lines indicating the channel to be followed, together with figures showing the depth of water, and brief identifying descriptions of important landmarks.

By using radar and radar charts, it is possible for a pilot to know exactly where he is and see the obstacles before him regardless of fog, darkness, or storm. Each image on the radarscope, whether showing an island, shoreline cliff, a building or a cluster of gas storage tanks, will be recognizable as though seeing the landmark itself.

The cost of charting a river by radar would be low in comparison to possible savings through speeding-up of river traffic and elimination of delays due to weather conditions.

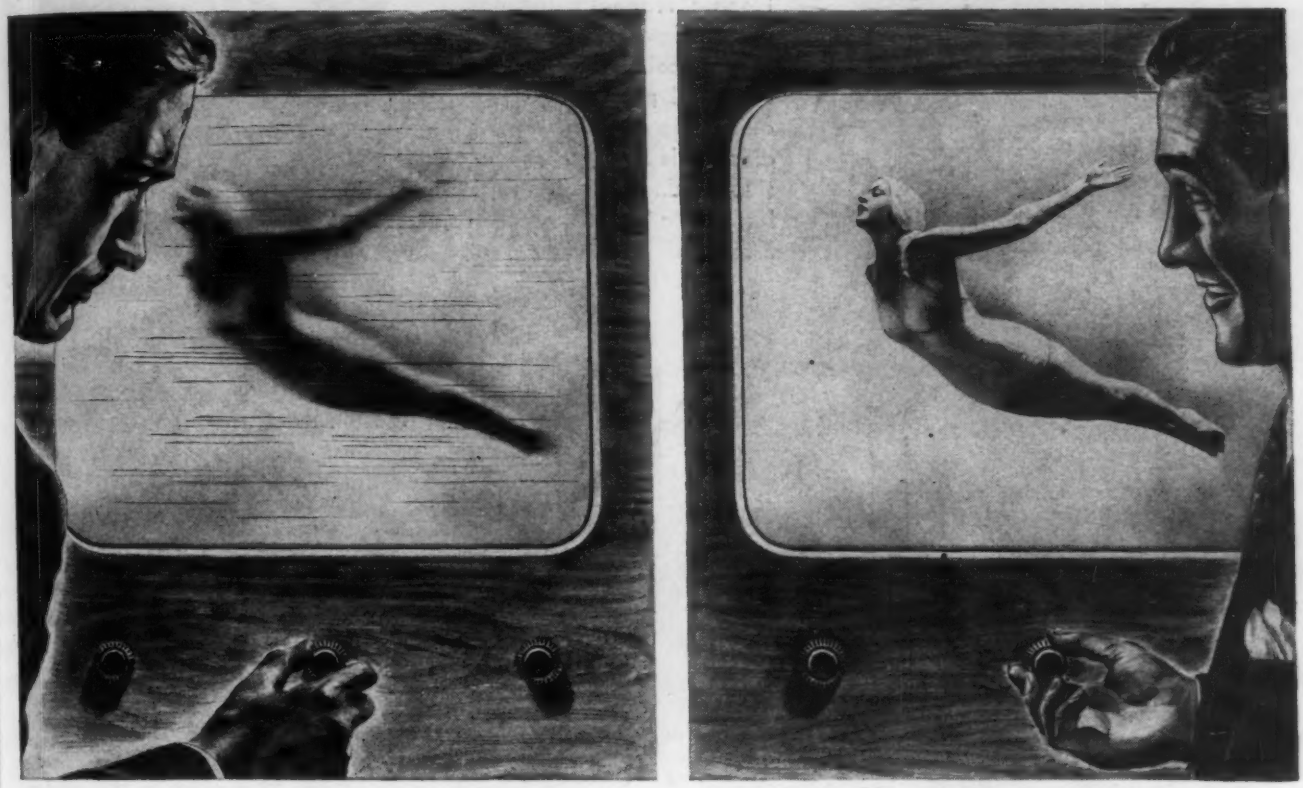
Use of the 3.2-centimeter band makes possible sharp, clearly-defined images that are easily photographed from the radar's 12-inch scope. The high-frequency beam hugs the surface of the water and picks up buoys or other small objects over distances twice as great as those afforded by lower frequency radars of the type used in wartime. Two or more closely spaced objects appear as separate, distinct targets in their true relationship to each other. Locks, bridges, shorelines, channel markers and approaching tows can be picked up by the radar.

The narrow beam and short pulse length used provide a high order of bearing and range resolution, with objects discernible as close as 80 yards and out to 1½ miles from the antenna. The radar can also be switched to longer ranges.

Brazing Lawn Mower Rotors

AN INTERESTING metal-joining operation by induction heating is the joining of three steel parts—a spider, drive shaft and bearing retainer—in one operation. This is

Nerves? OR Curves!



A Type ATV* Television Lead-In Line Can Make All the Difference

ANACONDA Type ATV lead-in lines are playing an important part in helping to give television buyers the kind of reception they want.

The well balanced design of conductors and dielectric in ATV lead-in lines minimizes the effects of attenuation and impedance mismatch. Satin-smooth polyethylene insulation sheds water readily — thus avoiding subsequent impedance discontinuities.

ATV line fulfills the most exacting requirements of wide-band reception — providing maximum freedom from distortion. Television buyers expect a lot. See that an ATV lead-in line helps your set to deliver!

A Type ATV Lead-In Line for Every Need

Anaconda offers a complete selection of Type ATV lead-in lines for 75, 125, 150 and 300 ohms impedance unshielded and 150 ohms shielded. For an electrical and physical characteristics bulletin, write to Anaconda Wire and Cable Company, 25 Broadway, New York 4, New York. Also, write for the characteristics of the complete line of Anaconda coaxial cables.

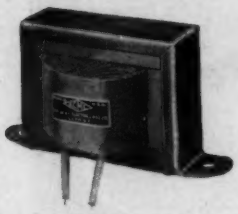
ANACONDA from mine to consumer

ANACONDA WIRE AND CABLE COMPANY

*An Anaconda Trade-Mark

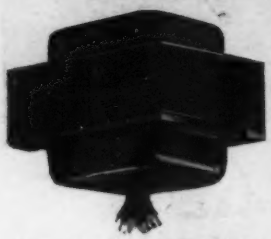
HOW TO BUY SPECIAL TRANSFORMERS *from* STANDARD PARTS

The special physical design or special electrical characteristic transformers you may need can probably be engineered from Acme Electric standard laminations and parts to the exact mechanical dimensions and electrical performance required. No need for special dies, tools or other expensive production materials. Here are a few of the Acme Electric designs available.

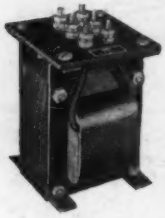


Mounting type 100. Available in ratings from 3 to 40 VA.

Mounting type 110. Available in ratings from 35 to 2500 VA.

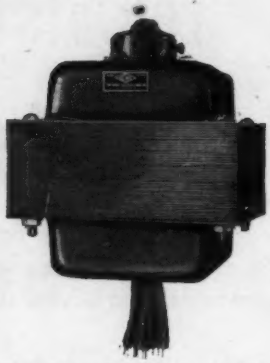


Mounting type 130, 2 hole horizontal mounting lead holes on bottom or side of shell. Ratings 15 to 100 VA.



Mounting type 121. Available in ratings from 35 to 2500 VA. Taps to suit your needs.

Mounting type 131 with end bells. Leads out bottom or side. Ratings 35 to 500 VA.



Mounting type 150. 4 hole horizontal mounting. Lead holes on bottom or side of shell. Primary tap changer on top. Ratings 35 to 500 VA.

Write for Specification Transformer Bulletin 168.

ACME ELECTRIC CORPORATION
31 WATER STREET + + + CUBA, N. Y.

Acme ACME Electric

Inserting lawn mower rotor assembly in heating coil for brazing operation

being done by Reading Hardware Co. in fast time with alignment of parts held to close tolerances.

At the brazing station, a girl assembler places four spiders in a ceramic fixture and brushes the center hole of each with Handy Flux. A ring of Easy-Flo brazing alloy wire is then slipped on the drive shaft and the end of the shaft and the brazing alloy ring are liberally brushed with flux. A bearing retainer is placed over the shaft end and the two parts are then inserted in place on the spider. The fixture with four such assemblies is ready for heating.

The induction heating unit has two brazing stations equipped with jigs for aligning and maintaining close tolerances between parts, especially between the drive shaft and bearing retainer. The fixture holding the four assemblies is placed over the heating coils and all heating is done from the bottom. Heating coils do not interfere with the operation of the jigs which are lowered over the parts while heating. Heat can be concentrated on the heavier parts—the drive shaft and the spider—thus preventing warping in the lighter gauge bearing retainer. With coils so arranged the brazing alloy is drawn through the joint bonding the parts together accurately and permanently.

Four assemblies are brazed at one time. It takes but 30 seconds, an average of 7.5 seconds each. With two brazing stations, one heating while the other is loaded, a production of 3,840 units can be maintained per eight-hour day.

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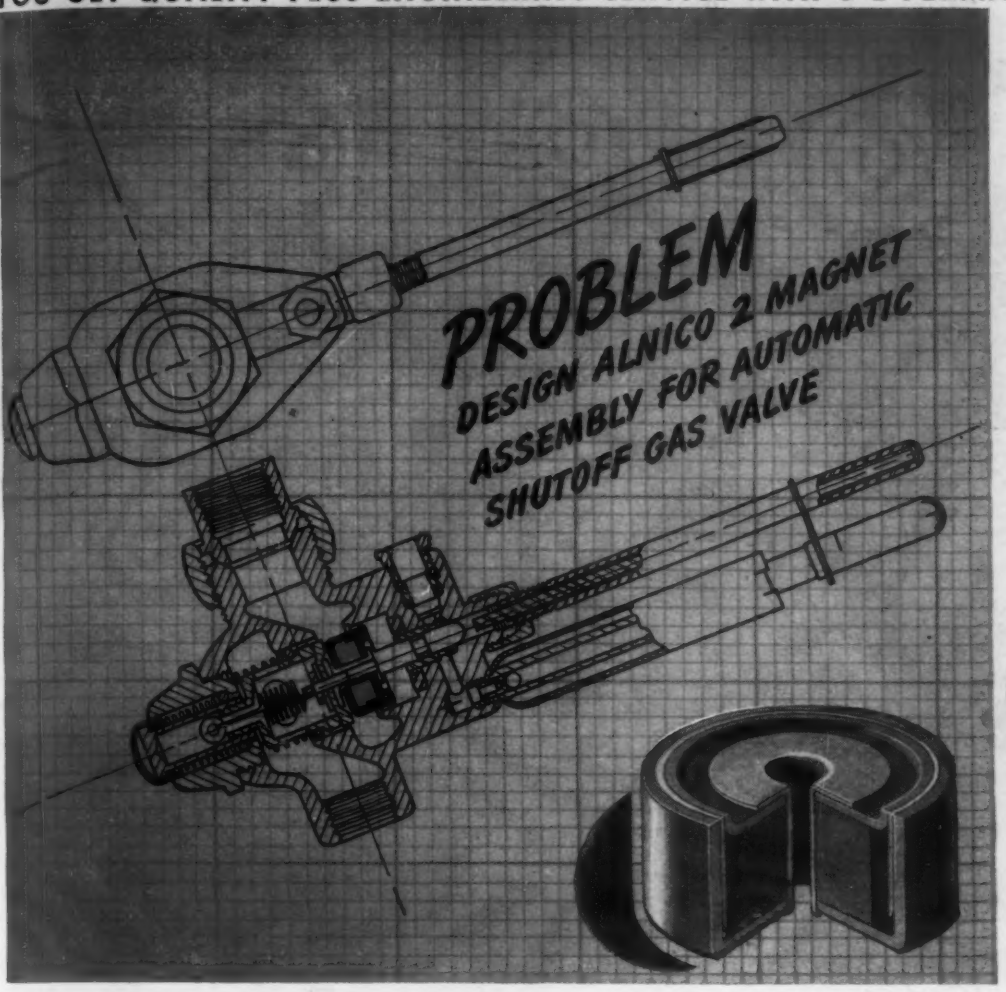
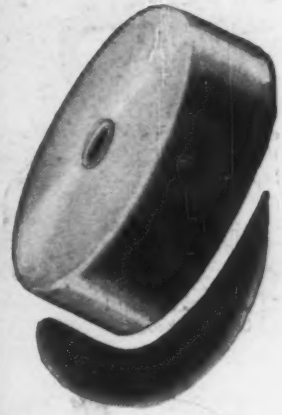


Designed by G-E engineers, this ALNICO permanent magnet holding assembly insures positive snap action in the Titan 100% Shutoff Automatic Safety Pilot manufactured by the Titan Valve and Mfg. Co.



The Magnet that Safeguards your Gas Heater

Here's a magnetically controlled pilot light for your gas heater that's 100% safe. While the pilot flame is burning, the ALNICO permanent magnet assembly grips a steel ring, keeping the gas line open. When the flame goes out, a thermal device and retaining spring force the holding magnet assembly away from the ring, shutting off the gas supply.

We shall be glad to help you also with your magnet application problems. Backed by years of research and magnet design experience, General Electric engineers are always at your service.

The precise quality control methods used throughout G-E magnet production plus accurate testing and rigid inspection, assure you of receiving magnets of the highest uniform quality for your application.

Moreover, greater flexibility of magnet design is possible with the many G-E permanent magnet materials now available. These include cast and sintered ALNICO, the ductile alloys CUNICO, CUNIFE and SILMANAL and lightweight nonmetallic VECTOLITE. *Metallurgy Division, Chemical Department, General Electric Company, Pittsfield, Mass.*

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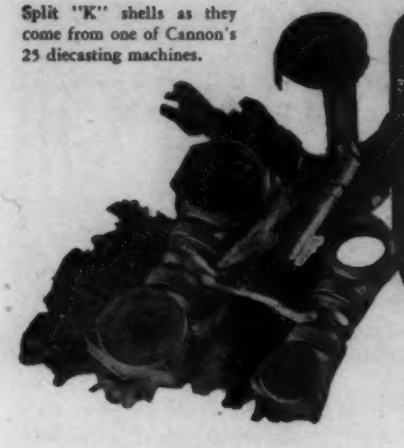
GENERAL ELECTRIC

CD47-JA6

Split "K" shells as they come from one of Cannon's 25 diecasting machines.

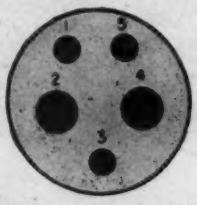


FROM DIECASTING TO COMPLETED PLUG All Under One Roof

WK-M2 insert, measuring ⅝" diameter, carries two No. 16 contacts (7/64" clearance). This phenolic insert may be used in any WK size shell.

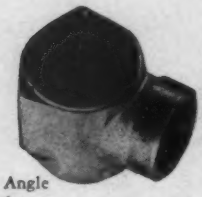
GK-P6 insert carries 3 No. 14 and 3 No. 10 contacts (3/64" clearance). Insert diameter ⅞".

"K" Angle 90° end bell after the "flash" has been removed.

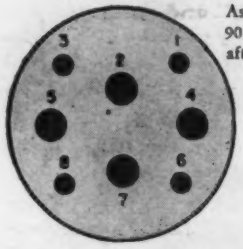


Largest insert diameter is "AK" with a maximum of 82 contacts.

"K" Angle 90° cap after excess metal has been roughed off.

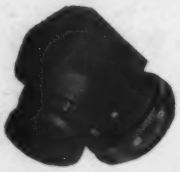


SK-C5 insert carries 3 No. 10 and 2 No. 6 contacts (1/16" clearance). One of the 188 inserts available.



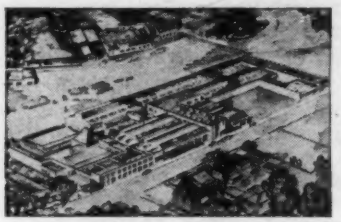
Assembled "K" Angle 90° end bell and cap after finishing.

Finished cap and end bell, conduit entry type, sand blasted and lacquered (or tin plate and lacquered). These are standard finishes. Others are available upon special request. quest.



NK-L8 insert carries 4 No. 14 and 4 No. 8 contacts (3/32" clearance).

K-23C Complete angle 90° assembly with integral clamp (cable entry).



While the Los Angeles Cannon Electric plant comprises two buildings, the manufacturing and assembly are now located in Plant No. 2 under one roof, so to speak, where the various processes from diecasting aluminum and zinc shells through the molding and fabricating of the insulators (insert arrangements) and the milling of contacts through to final assembly are completed.

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SINCE 1915

ELECTRON ART
(continued from p 138)

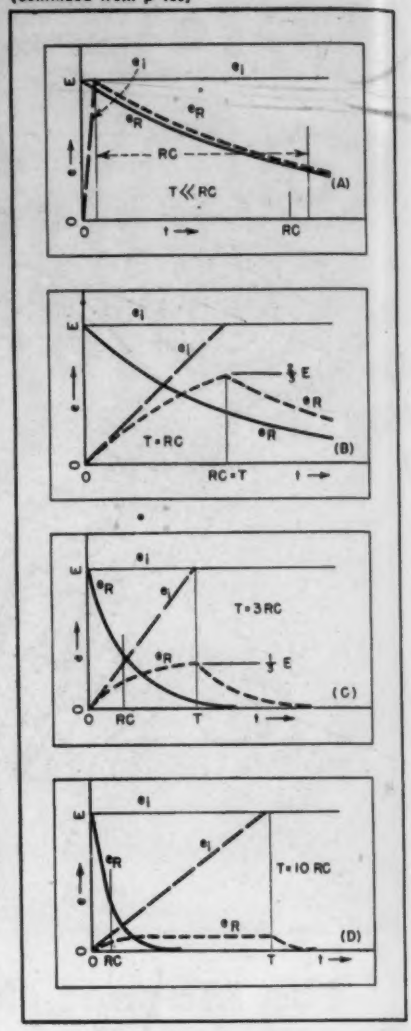

FIG. 3—Comparison of response to square-wave to that of linearly rising wavefront for various pulse periods relative to the circuit time constant shows considerable difference for relatively long pulses

$= 7$ or greater, the exponential term in Eq. 3 becomes less than 0.001, and the voltage rises in time T to $e_R = E/a$. For example, consider the case of $a = 10$. When the input voltage wavefront reaches its full value of E at $t = T = 10RC$, the voltage across the resistor reaches within better than 0.01 of one percent of its maximum value of $0.1E$. Actually the voltage has reached 95 percent of this value at $t = 3RC$ thus

$$e_R = (E/a)\,(1 - \epsilon^{-3}) = 0.095E \quad (7)$$

These four cases are illustrated in Fig. 3. The solid e_i curves are the square input voltage; the solid e_R curves are the output voltage (Fig. 2) resulting from such an input. The dashed e_i curves are a linearly rising input voltage; the dashed e_R curves are the output

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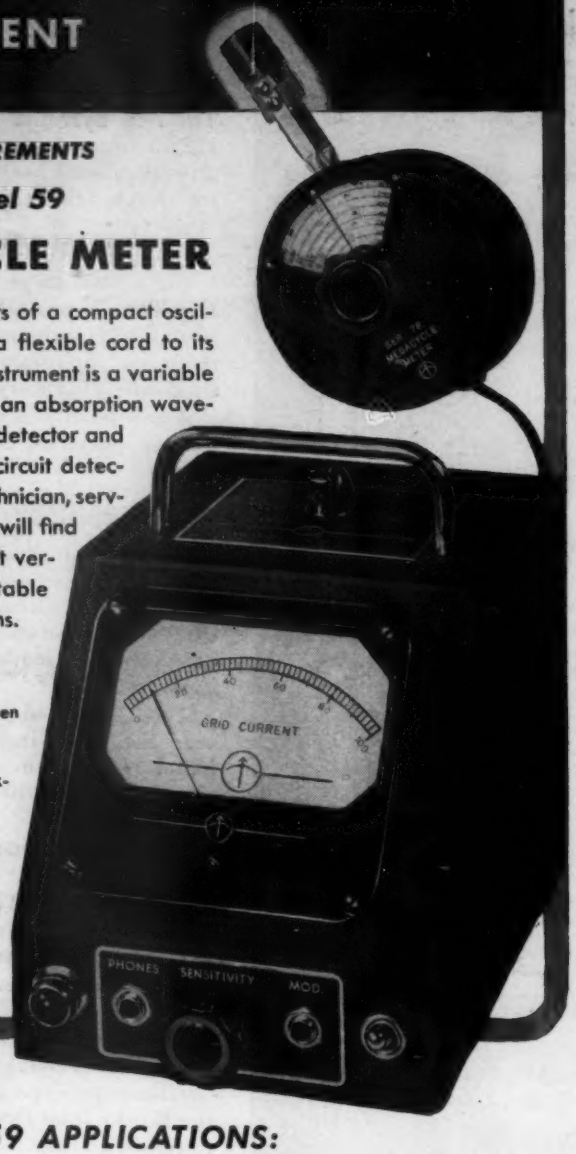

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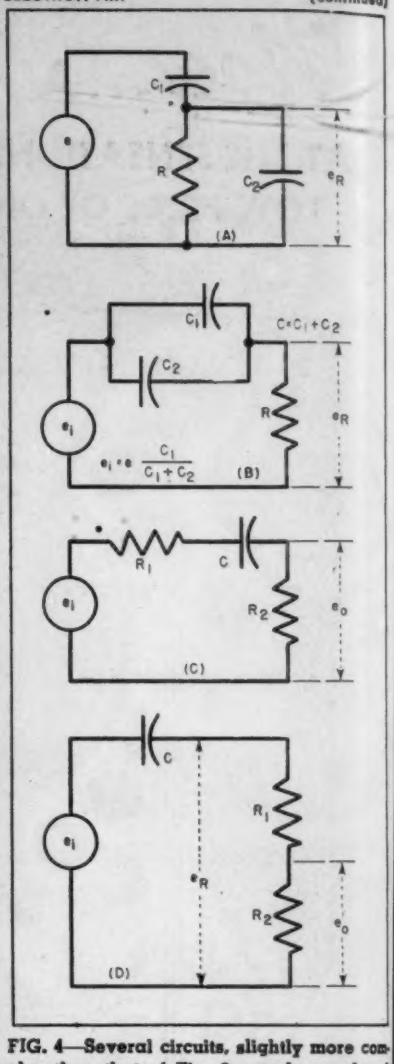

FIG. 4—Several circuits, slightly more complex than that of Fig. 2, can be analysed by the exact same technique. Other more complex circuits can be analysed by analogous techniques

from such an input. In Fig. 3A there is little difference between the output obtained from a square or a linearly rising waveshape. However in Fig. 3B there is a noticeable difference. In Fig. 3C and especially Fig. 3D there are no similarities. The errors resulting from assuming a squarewave are obvious.

Until now we have been discussing the rise period. After the input voltage has reached its maximum at time T, it remains constant for the duration of the pulse, and the voltage across the resistor falls exponentially, in accordance with the equation

$$e_R = E_R e^{-(t-T)/RC} \qquad (9)$$

where E_R is the maximum voltage reached across the resistor, and time is still measured from the origin of the pulse front. At the

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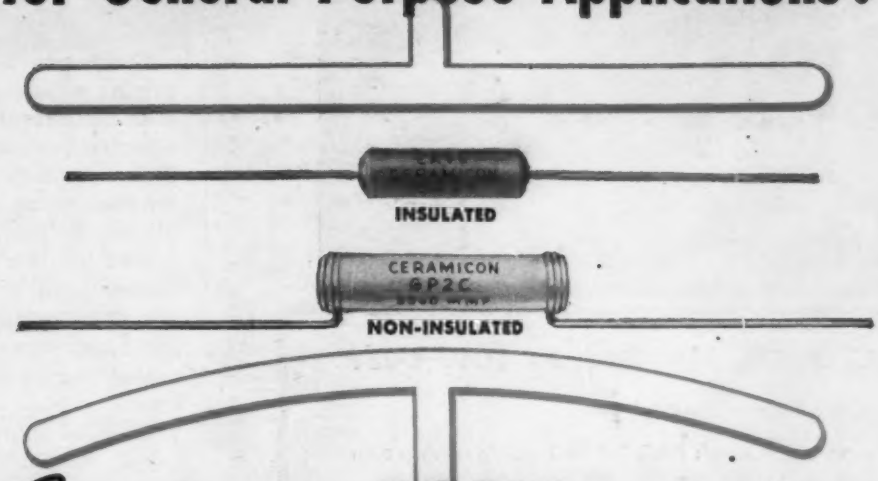

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tail of the pulse, a similar differentiated pulse appears with opposite polarity. We have used the term differentiation somewhat loosely, as the center portion of the differentiated pulse, between the two spikes, must be included in the concept of differentiation as applied to pulses.

Rules of Thumb

Comparing the results from assuming a linearly rising pulsefront with those resulting from assuming a perfectly square pulsefront (Fig. 3), we see how misleading the latter assumption can be. For example, in case four, the output voltage approaches very closely the true mathematical differential of the input wave, except for amplitude, as the input wave is Et/T and the output is $e = E/a$ during the rise time.

Rules of thumb for anticipating pulse response of circuits are therefore as follows:

(1) If the time constant of the circuit is much greater than the rise time of the pulse, the pulsefront is transferred across the circuit with slight loss in voltage (Fig. 3A).

(2) If the time constant of the circuit is about equal to the rise time of the pulse, the voltage output maximum is about $\frac{2}{3}$ of the input (Fig. 3B).

(3) If the time constant of the circuit is $\frac{1}{3}$ or less of the rise time of the pulse, the output voltage is approximately this fractional proportion of the input (Fig. 3C and 3D).

Practical Applications

There are several practical points which should be considered regarding the effect of the associated physical circuits to which the above differentiating circuit is connected. First, R should represent the complete effective load on the differentiating circuit. Second, usually some capacitance exists across the resistor, shown as C_2 in Fig. 4A. In this case, by application of Thevenin's theorem, there results Fig. 4B, to which we can apply the results arrived at above. The voltage produced across R by placing C_2 in parallel with C_1, and replacing the input voltage with the reduced input voltage $e_1 = eC_1/(C_1 + C_2)$,

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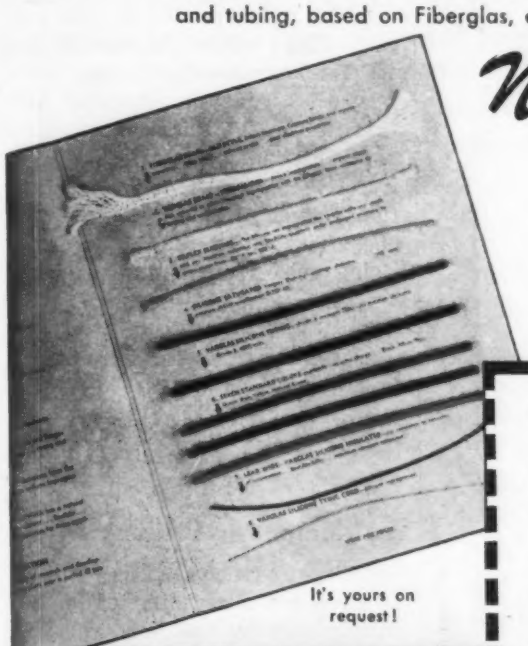

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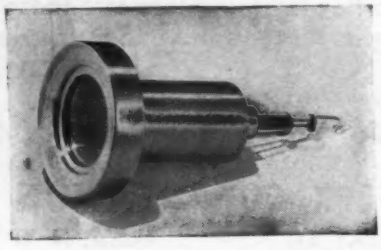

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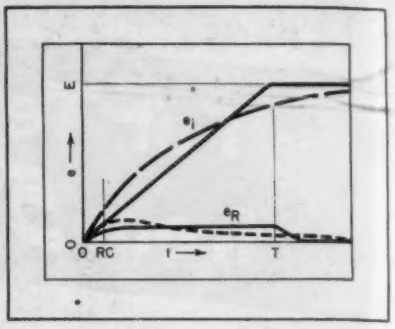

FIG. 5—Comparison of results assuming a linearly or an exponentially rising wavefront shows that they give results of the same order of magnitude

as shown in Fig. 4B, is the same as in the original circuit (Fig. 4A).

Third, the above analysis assumes that the output resistance of the generator supplying e_i is negligible in comparison with R. If this is not so, we can reduce it to this case as follows. In Fig. 4C, R_1 is the output resistance of the source, and R_2 is the load resistance. This circuit is exactly the same as Fig. 4D. Thus the results of the previous analysis apply by using the value of $R_1 + R_2$ as R to get e_R, and then multiplying by the ratio $R_2/(R_1 + R_2)$ to get the actual output voltage e_o.

Figure 5 illustrates, for the sake of comparison, an exponential pulse front (dashed e_i) with a time constant of one half the rise time of the comparable linear pulse front (solid e_i) shown with it. The output voltage across the resistor due to the exponential pulse front (dashed e_R) may be compared with that due to the linear front (solid e_R) as an indication of the reliability of the second order approximation compared to the third order one. The assumption of linearly rising pulse front is quite reliable.

(1) G. P. Ohman, Square-Wave Differentiating Circuit Analysis, ELECTRONICS, p 132 Aug 1945.
(2) Gershon J. Wheeler, Laplace Transforms for Electronic Engineers, ELECTRONICS, p 304 Feb 1945.
(3) N. W. McLachlin, "Complex Variable and Operational Calculus with Technical Applications", McMillan Co., 1944, Chapter 10.

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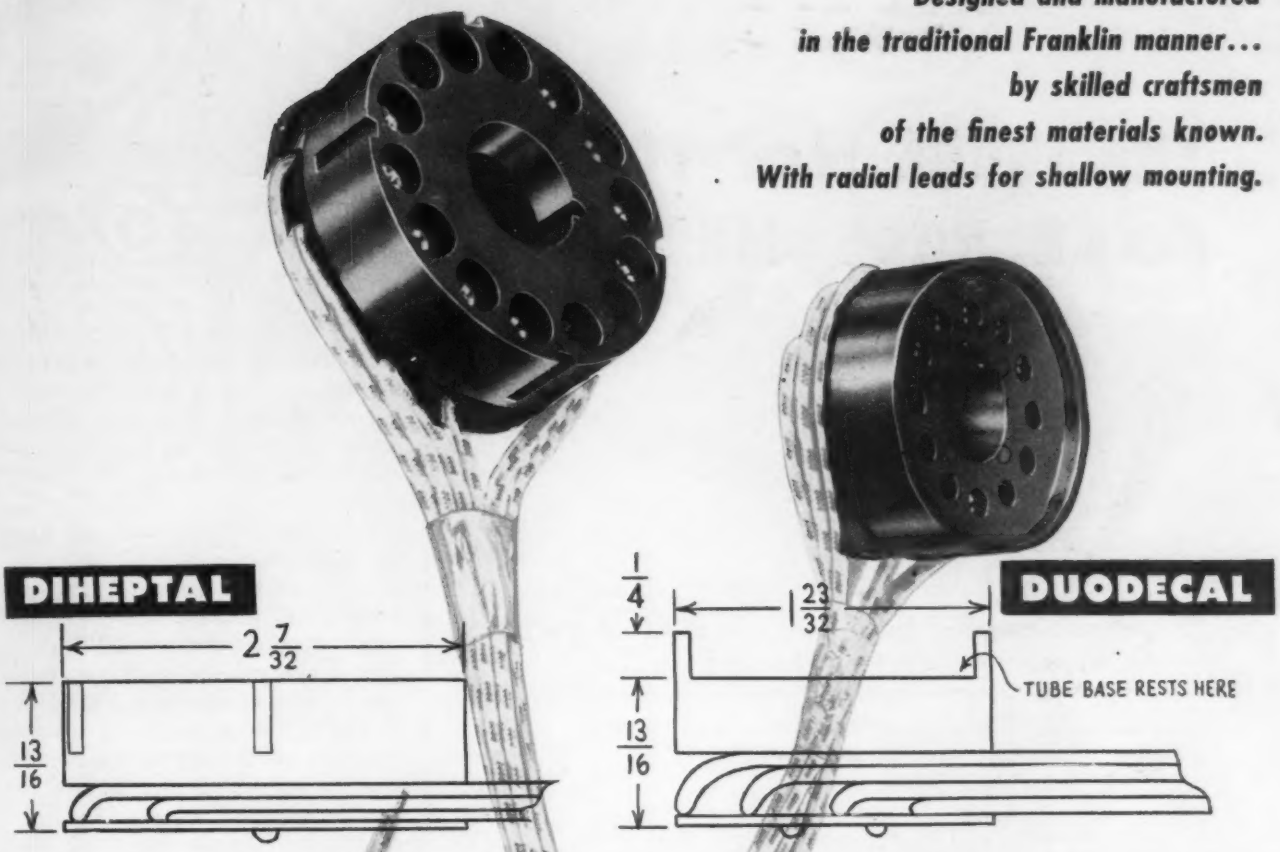

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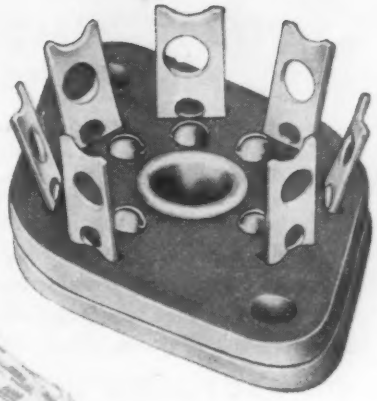

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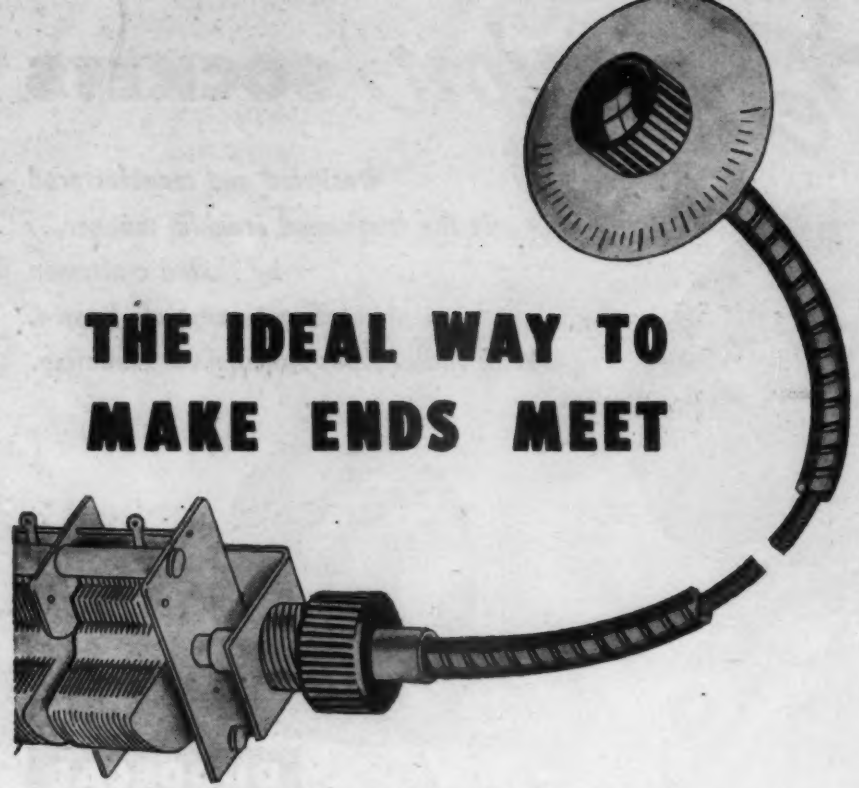

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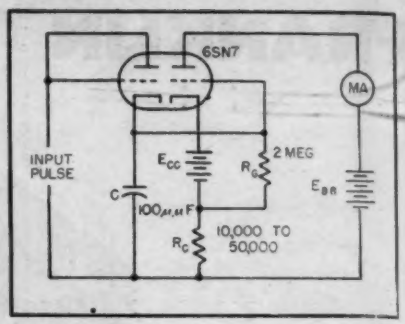

Voltmeter responds equally well to pulses from one to 20 microseconds duration

section of a 6SN7 with grid tied to plate, followed by a cathode coupled amplifier (cathode follower). Other tube combinations might be better (perhaps a 6SQ7). Amplifier grid bias is adjusted, depending on the plate supply voltage, to approximately cut off the amplifier plate current.

Capacitor C is charged through the diode during a positive pulse. The charge then leaks off through resistors R_g and R_c. However, the triode section of the 6SN7 then passes current developing a voltage across R_c to oppose the discharge of the capacitor. The time constant of the output circuit is thus increased. This increase of time constant of the actual circuit over that of the passive circuit elements alone approaches the amplification factor of the amplifier tube as R_c approaches the internal tube resistance. Because of the increased time constant of the output circuit, the capacitor can be small, thus increasing the input impedance of the voltmeter. Indication can be either by a plate milliammeter as shown, or by a conventional voltmeter across R_c. (A High Impedance Pulse Voltmeter, D. E. Howes, *Rev Sci Inst*, p 322 Nov 1945)

Pulse Rise and Decay Time Measurement

BY ALLEN EASTON
Long Island City, N. Y.

PRECISE MEASUREMENT of rise and decay characteristics of pulses is a necessary adjunct to their expanding application. Various characteristics of pulses are defined in Fig. 1. The rise or decay time can be measured by one of the common methods illustrated in Fig. 2. If

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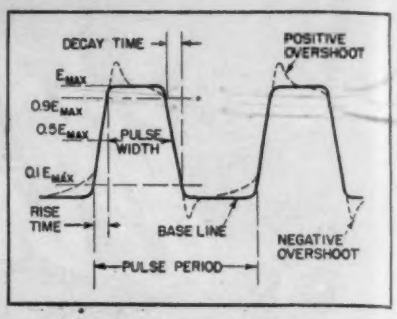

FIG. 1—Principal features of pulses show times to be measured

the pulse is symmetrical, that is the rise and decay times are substantially equal, an elliptical sweep can be used[1]. However when the pulse has a very steep rise or a very rapid decay, the pulse itself can be used as the sweep to obtain accurate time measurements.

Basic System of Measurement

Figure 3 shows two alternate arrangements of the measuring equipment. The trigger used to initiate the test pulse may be derived from a free running trigger generator or from a frequency divider driven by a high-frequency source. In the former case a pulsed oscillator is required to provide the measuring frequency. In the latter instance the high-frequency oscillator from which the trigger is derived may be the source of the measuring frequency.

The pulse whose rise time is to be investigated is connected to the

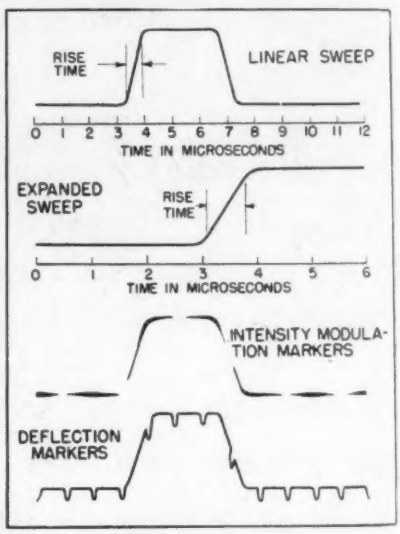

FIG. 2—Several techniques are frequently used to measure pulse rise or decay time. The top two methods depend on a linear sweep for a time base; the bottom two use time interval markers

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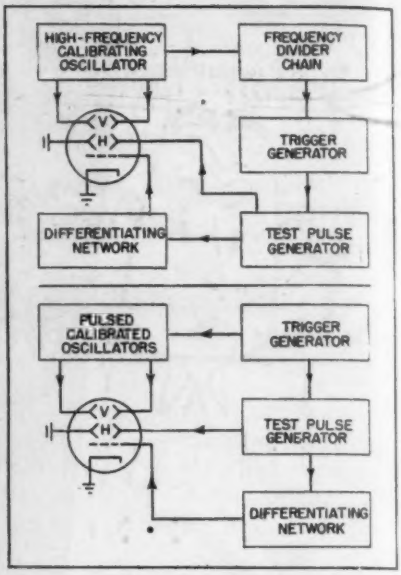

FIG. 3—The measuring technique can be accomplished by several arrangements of equipment

horizontal plates of the cathode-ray tube. Thus the pulse provides the sweep. The pulse is simultaneously passed through a differentiating network whose output is connected to the cathode-ray tube grid or cathode.

Before the pulse begins to rise, the beam of the cathode-ray tube is stationary and appears as a bright spot marked A in Fig. 4A. (assume for the moment the calibrating oscillator is not operating). As the

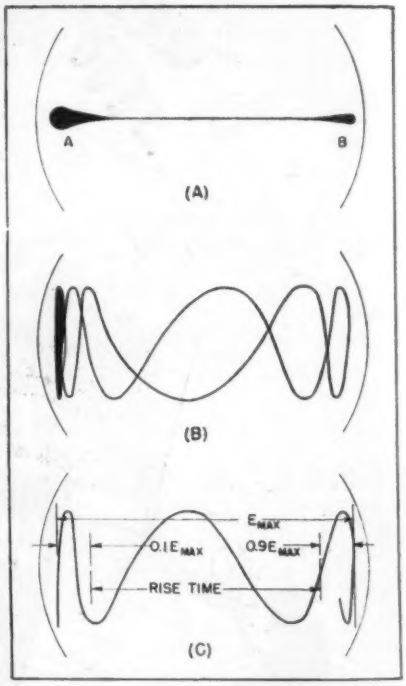

FIG. 4—Appearance of traces produced on oscilloscope during measurements

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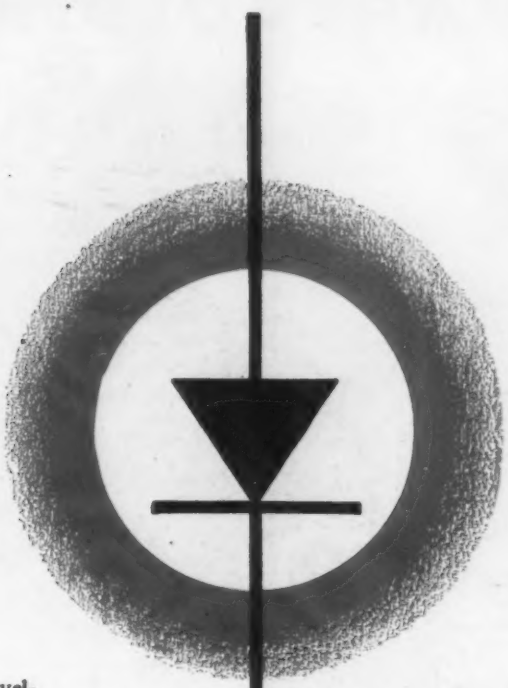

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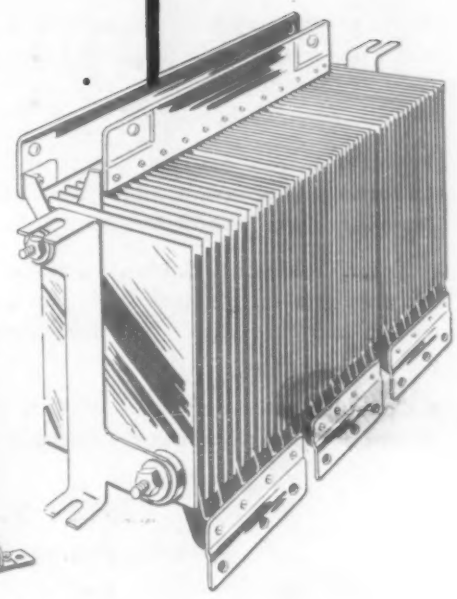

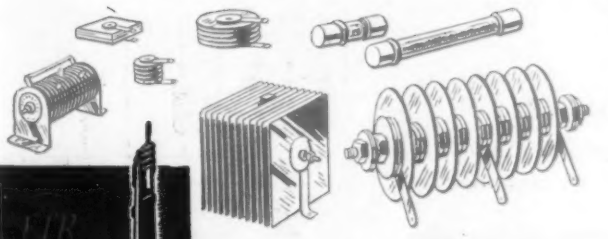

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pulse voltage rises, the spot sweeps across the screen at a velocity proportional to the rise time and for a distance proportional to the peak voltage and deflection sensitivity. The pulse eventually reaches the peak voltage and remains at this value for a while causing the bright spot at B, after which it sweeps back to the starting point A until the next cycle. Both rise and decay times contribute to the total deflection while the periods of relatively constant amplitude cause only bright spots. This fact should be contrasted with the other methods mentioned above wherein rise and decay times occupy only a very small percentage of the screen trace.

Now suppose the marker oscillator, which is synchronized with the initiating trigger, is connected to the vertical plates of the cathode-ray tube. This connection is analagous to the familiar use of sine-waves for sweep calibration. Figure 4B shows a possible result. The display in this figure still has limited utility because it contains both the forward and return traces superimposed, which may prove confusing. Blanking of the return trace is the purpose of the differentiating circuit.

It is well known that a pulse passed through an R-C circuit of proper time constant (sufficiently short compared with the pulse width) results in a pair of pulses, one positive and one negative. The times of occurrence of the pair of pulses coincide with the rise and decay times of the input pulse. If the output of the differentiating cir-

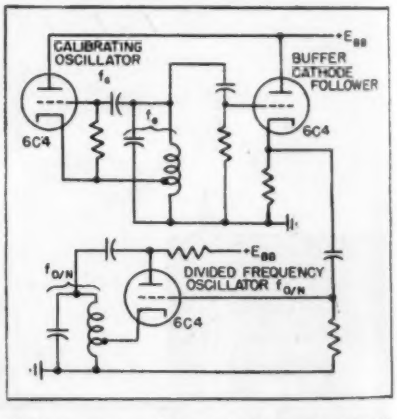

FIG. 5—A synchronized frequency dividing oscillator can be used to obtain a pulse triggering signal co-ordinated with the deflection signal

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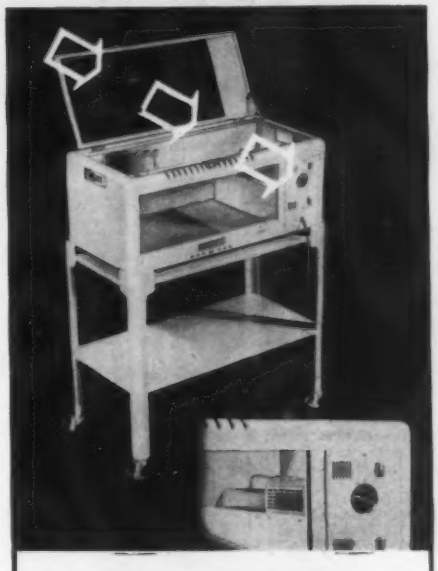

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cuit is connected to the intensity grid of the cathode-ray tube, the positive pulse will cause brightening of the trace and the negative pulse dimming. When the grid is initially biased near cutoff, then brightening will result for a time corresponding to the rise time of the test pulse. (If a negative test pulse is used the output of the differentiating circuit should be connected to the cathode of the cathode-ray tube to observe the rise time). The trace will then appear as in Fig. 4C. In some applications it has proved desirable to square the tops of the intensifying pulses in order to reduce the tendency toward defocusing of the trace.

If the frequency of the calibrating oscillator is known, the rate of rise can be studied between any two points along the rise characteristic by counting the number of cycles of the calibrating oscillator between them.

To observe the decay time of the test pulse, the output of the differentiating circuit may either be inverted or connected to the cathode of the cathode-ray tube. (A negative test pulse would require connection to the intensity grid to observe the decay time.)

Design Considerations

There are several design considerations which affect the choice of equipment. Because the rise time of a pulse may be exceedingly fast (for example 0.05 microsec), a cathode-ray tube employing high accelerating voltages should be used in order to obtain usable brilliance, especially if a pulse of low repetition frequency is to be studied. Obviously the highest feasible value of pulse repetition rate should be used wherever possible. High accelerating voltages required for good brilliance are usually accompanied by lower deflection sensitivity although new tubes with 20,000 volts of accelerating potential and high deflection sensitivity are available commercially[3]. Thus a fairly large peak pulse voltage is needed to secure sufficient deflection. If small pulses are to be measured, a pulse amplifier is essential but usually will distort the rise time of extremely fast pulses. This defect is common to most

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10	10	10 V.	100		0-200,000 (1200 ohms center)
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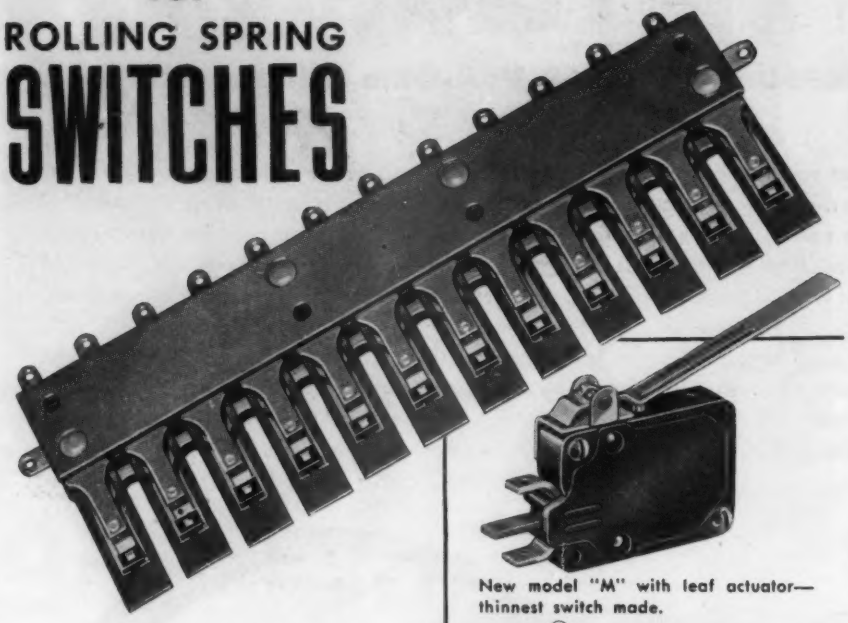

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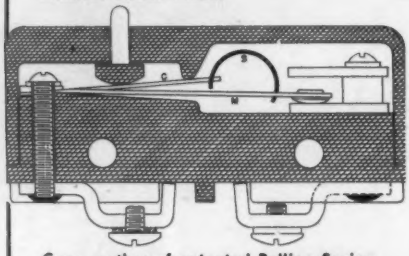

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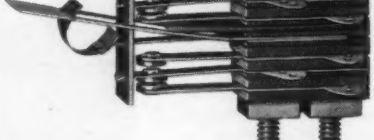
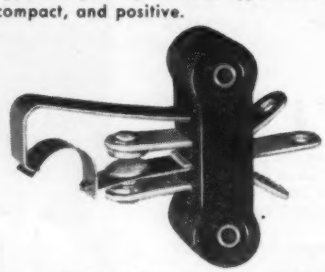


Typical 2 pole open blade type. Small, compact, and positive.



Open Blade Model "M" for limited space and low cost.

ACRO ELECTRIC COMPANY

1316 SUPERIOR AVENUE, CLEVELAND 14, OHIO

methods of rise time measurement.

It is, of course, not essential that the measured pulse be connected to the horizontal plates of the cathode ray tube. The vertical plates may be preferred.

The calibrating oscillator may take several forms; the commonest is a free running oscillator operating at a fairly high frequency. (For example, a 100 mc oscillator will give 10 marker cycles when observing a pulse with a 0.1 microsec rise time, whereas a 10 mc oscillator will give 10 cycles with a pulse of 1 microsec rise time.) The use of a chain of synchronized sinewave oscillators[3] makes it possible to obtain division from 100 mc down to trigger frequency, with each oscillator providing a frequency division of ten to one, or better yet, five to one. A fundamental circuit diagram of a suitable synchronized sinewave frequency dividing oscillator is shown in Fig. 5. It is also possible to utilize a pulsed oscillator as a source of calibrating frequency[4].

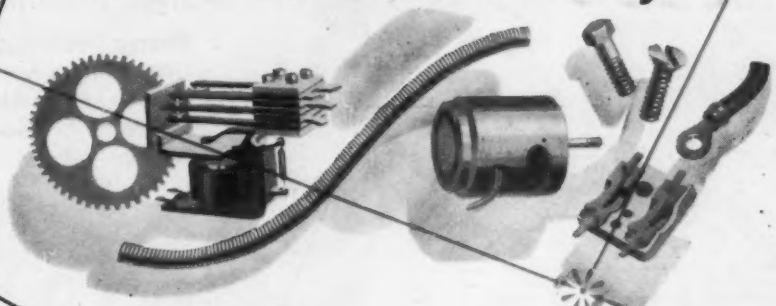
Numerous variations of the principles outlined in this paper are possible. No attempt has been made to show particular circuit arrangements as these details will depend on the types of pulses to be measured and the available laboratory equipment. Also the equipment arrangement for a laboratory measurement will differ considerably from that for measuring and standardizing production pulse generators.

(1) Allan Easton, Measuring Pulse Characteristics, ELECTRONICS, p 150 Feb 1946.
(2) I. E. Lempert and R. Feldt, The 5RP Tube, Intensifier Type CR Tube for High Voltage Operation, *Proc Inst Radio Engrs*, p 432 July 1946.
(3) C. W. Carnahan and H. P. Kalmus, Synchronized Oscillators as FM Limiters, ELECTRONICS, p 109 Aug 1944.
(4) Allan Easton, Pulse Modulated Oscillator, ELECTRONICS, p 125 March 1947.

SURVEY OF NEW TECHNIQUES

INFRARED RADIATION to wavelengths as long as 30,000 Angstroms is recorded on a spectrograph developed at Northwestern University under Dr. R. C. Nelson's direction. Mirrors mounted on precision bearings disperse an infrared ray into its component wavelengths and passes it across a Cashman cell. The electrical current so generated is amplified and recorded on a 10-inch

In the electronic industry



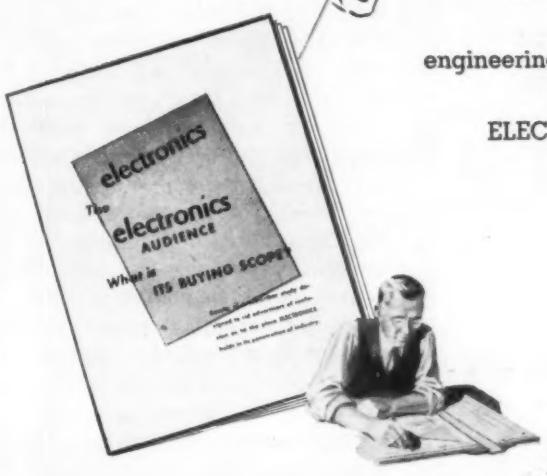
WHO buys these products?

Products that are not basic electronic components but which are vitally important in the manufacture of electronic equipment and components.

These are not basic electronic components such as tubes, resistors, coils and capacitors for which the market is well known and established. These are products such as precision gears, flexible shafting, sensitive relays, snap action switches, fractional hp motors, fasteners, etc. Products that, although not basic electronic components, are vitally important to, and used *in volume* in the manufacture of electronic equipment, components and controls. Although it is known that the market for these products is large in volume, it is not so well known *who buys these products.*

electronics ***determined to find out . . . and did . . .*** by asking every ninth subscriber whether he directed or influenced the purchase of these products.

the results of this subscriber study defining the buying scope of ELECTRONICS' circulation in relation to these products is now available. A request to your local ELECTRONICS' representative will bring you a factual analysis of "What They Buy" — from drafting equipment to telemetering systems; "Who Does the Buying" — an amazing galaxy of research, engineering and miscellaneous titles throughout the industry; and, based on a percentage of purchasing power, a determination of your ELECTRONICS advertising budget. Study it carefully and use it as a valuable guide to the electronic market for these products.



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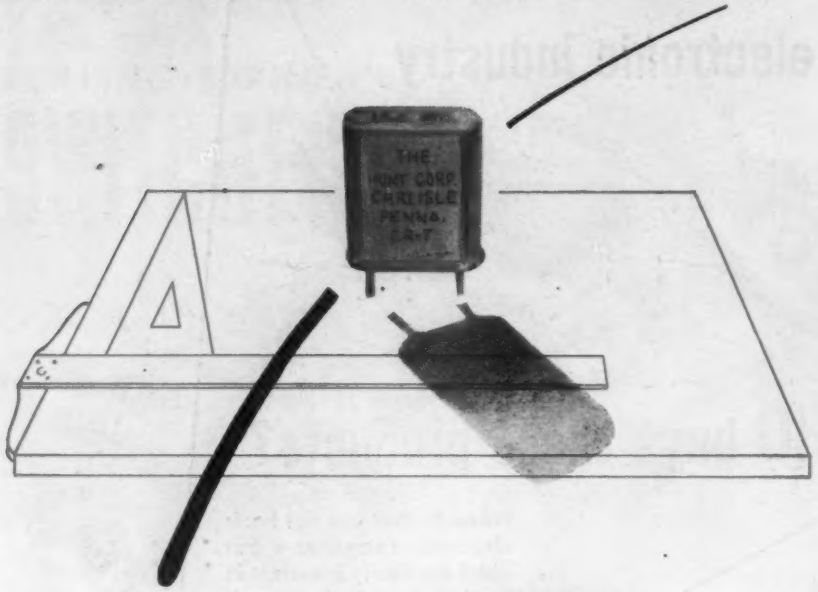

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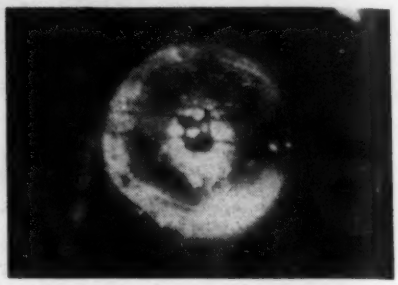
CARLISLE, PA.

Pacemaker in Stability

graph a quarter of a mile long. This infrared region is of interest to astronomers in locating relatively cold bodies, studying atmospheres of planets, and gases of the Sun, and to physicists in studying radiations from atoms.

TRACER TECHNIQUES using radioactive isotopes are made more effective by magnetically focusing an image of the radiation given off by the test material onto a photo-



Tracer micrograph shows distribution of radioactive isotopes within a thin layer of test material

graphic plate. The technique is being used at the National Bureau of Standards by L. Matron in cooperation with P. H. Abelson of Carnegie Institute. The tracer micrographs provide greater resolution in localizing positions of radioactive particles and thus enable their course in biological and chemical actions to be more accurately followed.

TRUE TEMPERATURES of Alnico and Alcomax permanent-magnet alloys are measured in the research department of Messrs William Jessop and Sons (Sheffield, England) by a simple photoelectric pyrometer developed by T. Land and H. Lund. A single lens of 30-cm focal length and 4-cm aperture projects an image of the molten surface on a 2.5-cm diameter selenium barrier-layer photoelectric cell behind a 1-cm aperture. Light baffles in the mounting tube exclude ambient light. The photocell operates a 500-ohm 15-μa microammeter. In calibrating the instrument against an immersion type pyrometer, the spectral emissivities of the molten surfaces were calculated and introduced into the photopyrometer by means of a variable resistance shunt. For melts having clean surfaces, the instrument provides absolute temperatures within $\pm$ 10 C.

HEAVY ARTILLERY IN WAR ON T. B.

ROCKBESTOS
Firewall Hookup Wire

Insulated with high-dielectric synthetic tapes and a "firewall" of impregnated felted asbestos, and covered with color-coded lacquered glass braid, this heat, flame and moisture resisting wire may be operated *continuously* at rated voltage under 125° C. without baking out. Widely used in airborne, marine, ground and mobile communications systems, electronic devices and compact apparatus in which dependable performance is essential. Also ideal for small motor, coil, transformer and dynamotor leads. Sizes 22 to 4 AWG in 1000 volt rating and 12, 14 and 16 AWG in 3000 volt rating.

One of 125 different constructions designed by Rockbestos for severe operating conditions in ratings ranging from 300 to 5000 volts.

In the battle against tuberculosis the Fairchild Fluoro-Record Camera pictured above has recorded the condition of hundreds of thousands of lungs. Mounted on X-Ray equipment, it is used by the U. S. Public Health Service and other public and private organizations for mass screenings. By this new method the automatic 70 mm camera photographs the chest as reflected on a fluoroscopic screen, and with 400 single negatives per roll of film hundreds of permanent records can be made daily.

Fairchild wires these cameras with Rockbestos Firewall Hook-up Wire for several good reasons. They liked its performance record in their military cameras. They needed a small-diameter wire with high dielectric strength that could withstand the heat generated by mass radiography — Underwriters' approved for operation at 90° C. — and they had to be sure that it wouldn't dry out or bake brittle and would have ample resistance to aging, moisture and flame.

This is but one of hundreds of cases in which manufacturers of electrical products ranging from appliances to precision instruments protect performance with *permanently insulated* Rockbestos wires, cables and cords. 125 different standard types to select from. For recommendations or engineering assistance, write to nearest district office or:

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NEW PRODUCTS

(continued from p 142)

Pickering model 120M cartridges. Arm resonance is less than plus or minus 1 db over the audio range from 40 to 15,000 cycles and provides a pressure of 22 grams. The arm is 15⅛ inches long and weighs about 10 ounces.

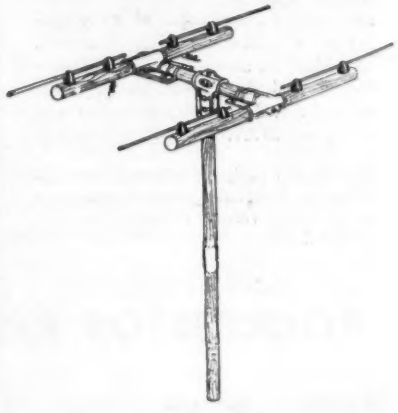
Miniature Tube VTVM (14)

ALLIED LABORATORY INSTRUMENT, INC., 355 West 26th St., New York 1, N. Y. The model 730 vacuum tube voltmeter incorporates miniature

tubes, an r-f probe with flat response to 120 mc, metal film resistors, and an accuracy of plus or minus 3 percent. The unit is light, small and low in price.

F-M and Television Antennas (15)

JFD MANUFACTURING CO., Inc., 4117 Ft. Hamilton Parkway, Brooklyn 19, N. Y. A new line of f-m and television antennas, together with wall insulators for holding the



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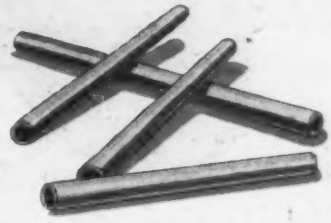
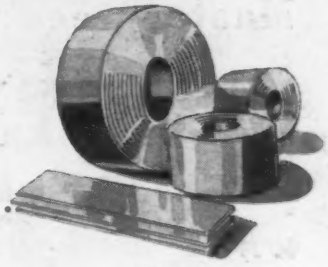
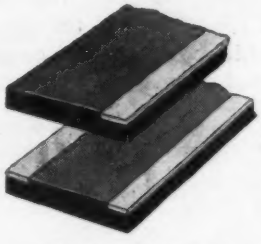


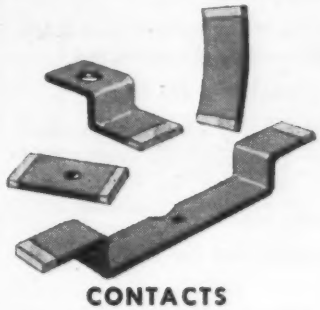
TUBE



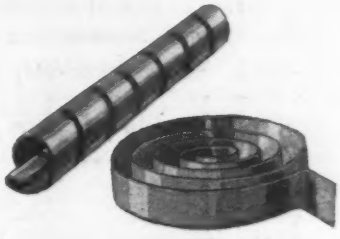
SHEET.



INLAY METALS



CONTACTS



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Try this easy 4 point test to prove that Soundcraft offers you the *greatest* potential for a recording masterpiece.

1. Cut several unmodulated grooves in any competitive blank disc. Make certain cutter is completely disconnected.
2. Take a Soundcraft blank and repeat step No. 1.
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Soundcraft will be quieter, have less crackle and the meter will be steadier.

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• *The 'Playback'* 8½" 8" 10" 12" 16"

• *The 'Maestro'* 12" 13¼" 17¾"

REEVES SOUNDCRAFT CORPORATION

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transmission line in place, have been announced. A folder describing the equipment in detail is available.

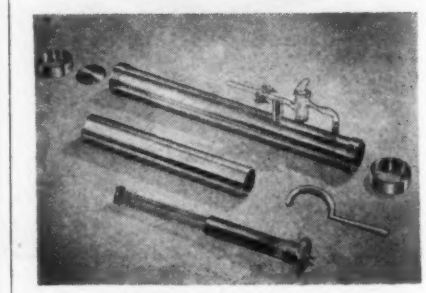
Subminiature Electrometer Tube (16)

VICTOREEN INSTRUMENT CO., 5806 Hough Ave., Cleveland 3, Ohio. Type XV-41 subminiature electrometer tubes have incorporated new design features reducing mi-

crophonics to a minimum. Filament voltage is 1.25 v; filament current, 10 ma; grid current, less than 10^{-15} amp; grid resistance, greater than 10^{15} ohms.

Screen Wall Counter (17)

RADIATION COUNTER LABORATORIES, 1451 East 57th St., Chicago 37, Ill. The Libby screen wall counter can be used to measure soft radiation



from a sample that can not be obtained as a suitable gas, or any radiation from a sample of extremely low specific activity that is available in large quantities. All types of particle radiation, alpha particles, beta particles and conversion (Auger) electrons can be counted.

F-M and A-M Tuner (18)

BROWNING LABORATORIES, Inc., Winchester, Mass., has recently developed an f-m and a-m tuner, Model RJ-14, in rack panel style for com-

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...BASIC TUNING UNITS

These transmitting condensers should be the basic tuning units of your transmitter designs.

Compact assembly, steatite insulation, sturdy construction, and conservative ratings will enable National condensers to prove their superiority in your transmitter.

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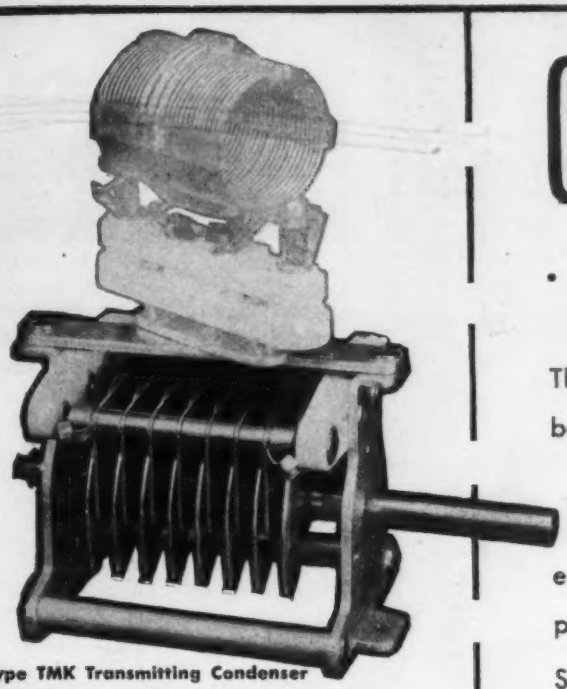


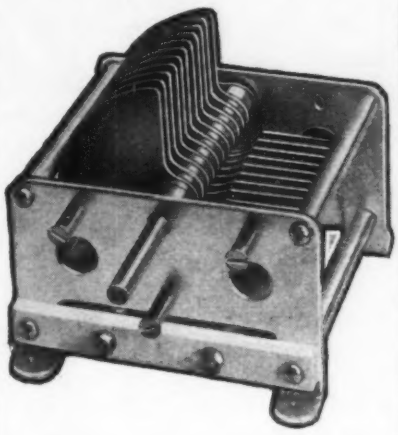
Type TMK Transmitting Condenser

An ideal condenser for exciters and low power transmitters. Available in single and double stator models. Steatite insulation. Special provision has been made for mounting AR-16 exciter coils in a swivel plug-in mount on either the top or rear of the condenser if desired. Over-all width 2-11/16", height 2-23/32".

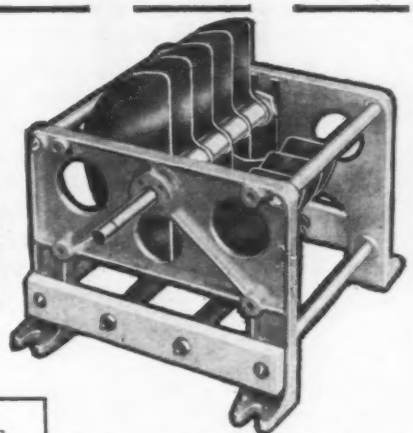
Capacity (max.-min.)	Air Gap	Peak Voltage	Length	Catalog
35 mmf.-7.5	.047"	1500v.	2-7/32"	TMK-35
50 -8	.047"	1500v.	2⅜"	TMK-50
100 -10	.047"	1500v.	3"	TMK-100
DOUBLE STATOR MODELS				
35/35 mmf.-7.5/7.5	.047	1500v.	3"	TMK-35D
100/100-10/10	.047	1500v.	4¼"	TMK-100D
Complete List in Catalog				

Type TMC Transmitting Condenser

Designed for use in the power stages of transmitters where peak voltages do not exceed 3,000. The frame is extremely rigid. Insulation is steatite. The stator in the split stator models is supported at both ends. Over-all width 3-9/16", height 3-5/8".



Capacity (max.-min.)	Air Gap	Peak Volt.	Len.	Cata.
50 mmf.-10	.077"	3000v.	3"	TMC-50
100 -13	.077"	3000v.	3½"	TMC-100
250 -23	.077"	3000v.	6"	TMC-250
DOUBLE STATOR MODELS				
50/50 mmf.-9/9	.077"	3000v.	4⅜"	TMC-50D
100/100-11/11	.077"	3000v.	6¾"	TMC-100D
Complete List in Catalog				



Capacity (max.-min.)	Air Gap	Peak Volt.	Len.	Cata.
75 mmf.-25	.719"	20,000v.	18 1/16"	TML-75E
50 -22	.469"	15,000v.	8 1/16"	TML-50D
500 -55	.219"	7,500v.	18 1/16"	TML-500A
DOUBLE STATOR MODELS				
30/30 mmf.- 12/12	.719"	20,000v.	18 1/16"	TML-30DE
60/60- 26/26	.469"	15,000v.	18 1/16"	TML-60DD
100/100- 27/27	.344"	10,000v.	18 1/16"	TML-100DB
Complete List in Catalog				

Type TML Transmitting Condenser

This is a 1 KW job throughout. Special steatite insulators prevent arc-covers. Sturdy cast aluminum end frames and dural tie bars permit an unusually rigid structure. Precision end bearings insure smooth turning and permanent alignment of the rotor. Over-all width 7", height 7-1/8".

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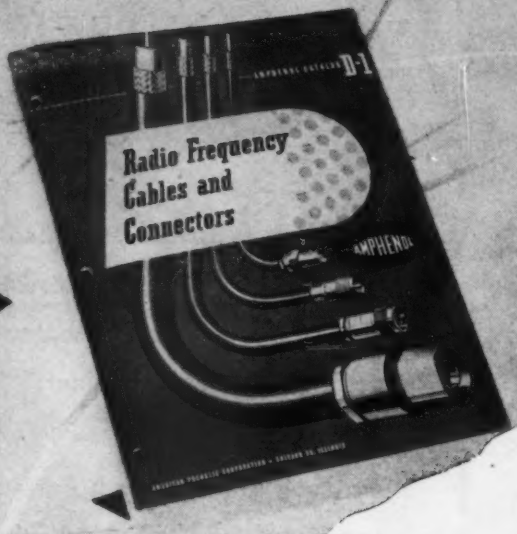

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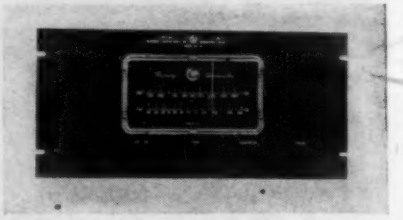


mercial or other application. Antenna leads and audio output are brought out at the rear of the chassis. The f-m antenna and feeder system also functions as the a-m antenna. The power supply is self-contained and operated from 115-volt 60-cycle source.

Image Camera (19)

INTERNATIONAL MUTOSCOPE CORP., Long Island City, N. Y. A television receiver, coin-operated, that is equipped with a camera for photographing images and delivering a finished print in one minute has been developed for use in amusement arcades.

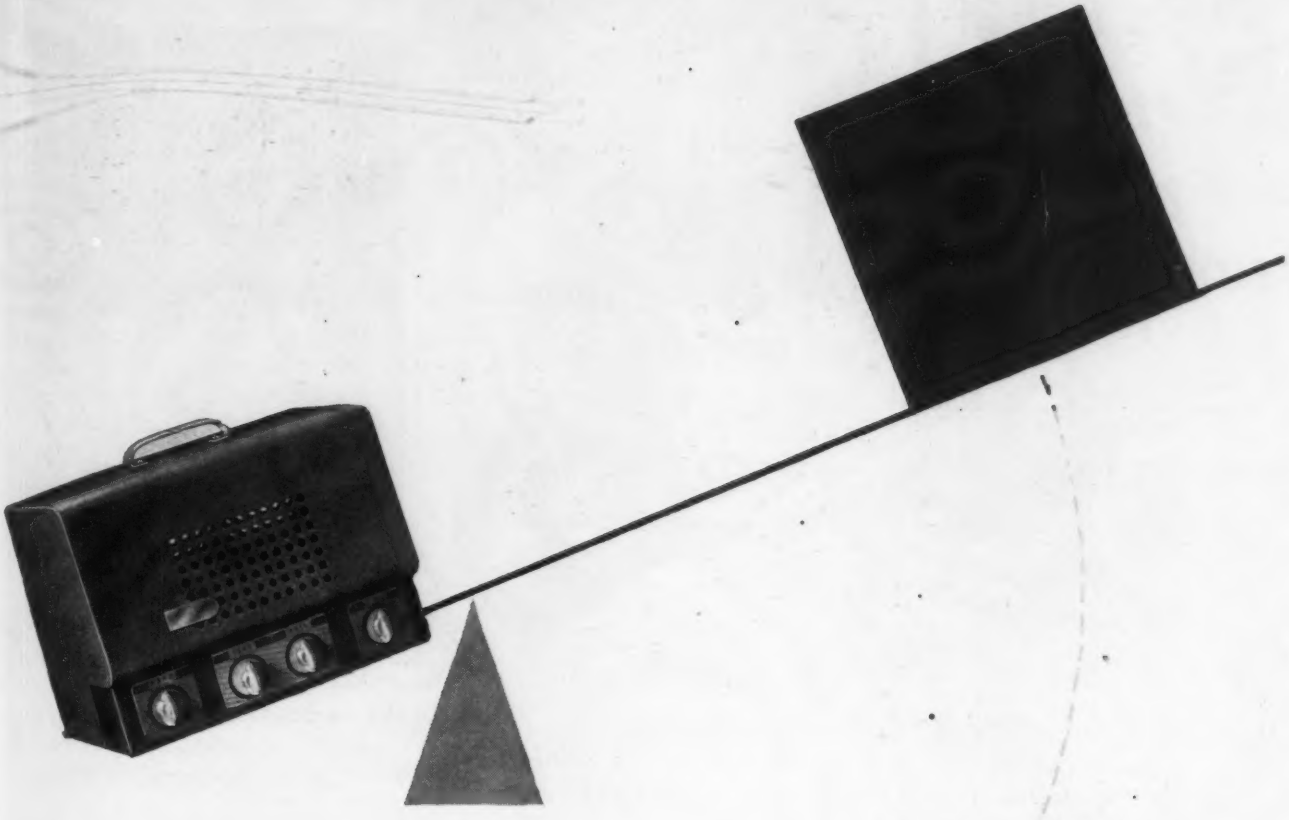
VFO Exciter (20)

BARKER AND WILLIAMSON, INC., 237 Fairfield Ave., Upper Darby, Pa. The model 500 variable frequency oscillator provides high-stability

fundamental driving power on all amateur bands into a type 807 output tube that can be used as a transmitter or driver for higher power. The model 502 contains only the essential variable frequency oscillator elements.

Thyratron Modulator (21)

DORMITZER ELECTRIC & MFG. CO., 782 Commonwealth Ave., Boston 15, Mass. The model 12 laboratory type pulse modulator has been developed for high-frequency studies. It has a peak pulse power of 250 kw, maximum pulse duration of 6 mi-



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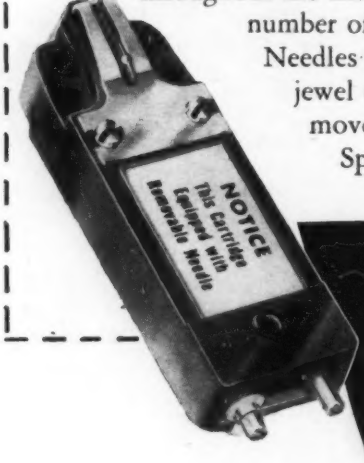
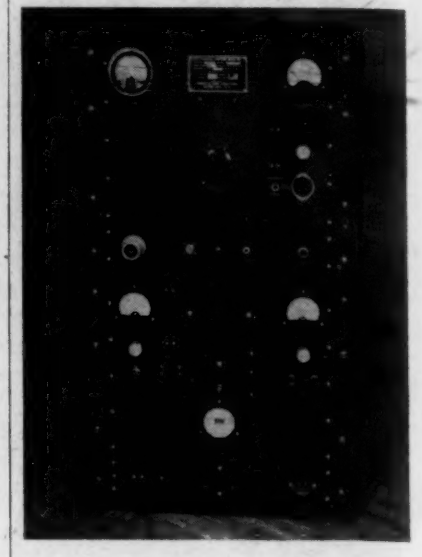
BECAUSE OF his build, character and performance, Astatic's Mr. "Q.T." Pickup Cartridge has earned the confidence of many leading radio-phonograph engineers and manufacturers, and is now "going places" as a vital unit in the newest, high quality-type record players.

If asked why the new Model "QT" Cartridge has been so generally approved, these designers and producers of phonograph equipment would undoubtedly state that the "QT" Cartridge supplies a clear, clean type of reproduction essentially free from annoying needle scratch, and that such reproduction remains constant during the life of the instrument.

This is true because the "QT" Cartridge is equipped with a MATCHED Needle, possessing all the qualities of a permanent needle yet having the advantage of being REPLACEABLE. This provides assurance that the original quality of reproduction shall be maintained throughout the life of the cartridge regardless of the number of times the needle is replaced. "QT" Needles are available with precious metal or jewel tip, and may be easily inserted or removed when replacement is necessary. Special literature is available.



THE Astatic CORPORATION
ASTATIC
CONNEAUT, OHIO
IN CANADA: CANADIAN ASTATIC LTD., TORONTO, ONTARIO
Astatic Crystal Devices Manufactured under Brush Development Co. patents.



croseconds, minimum duration of 1 microsecond, maximum repetition rate of 4,000 pulses per second and operates into a 50-ohm load. A 200-volt positive trigger is required.

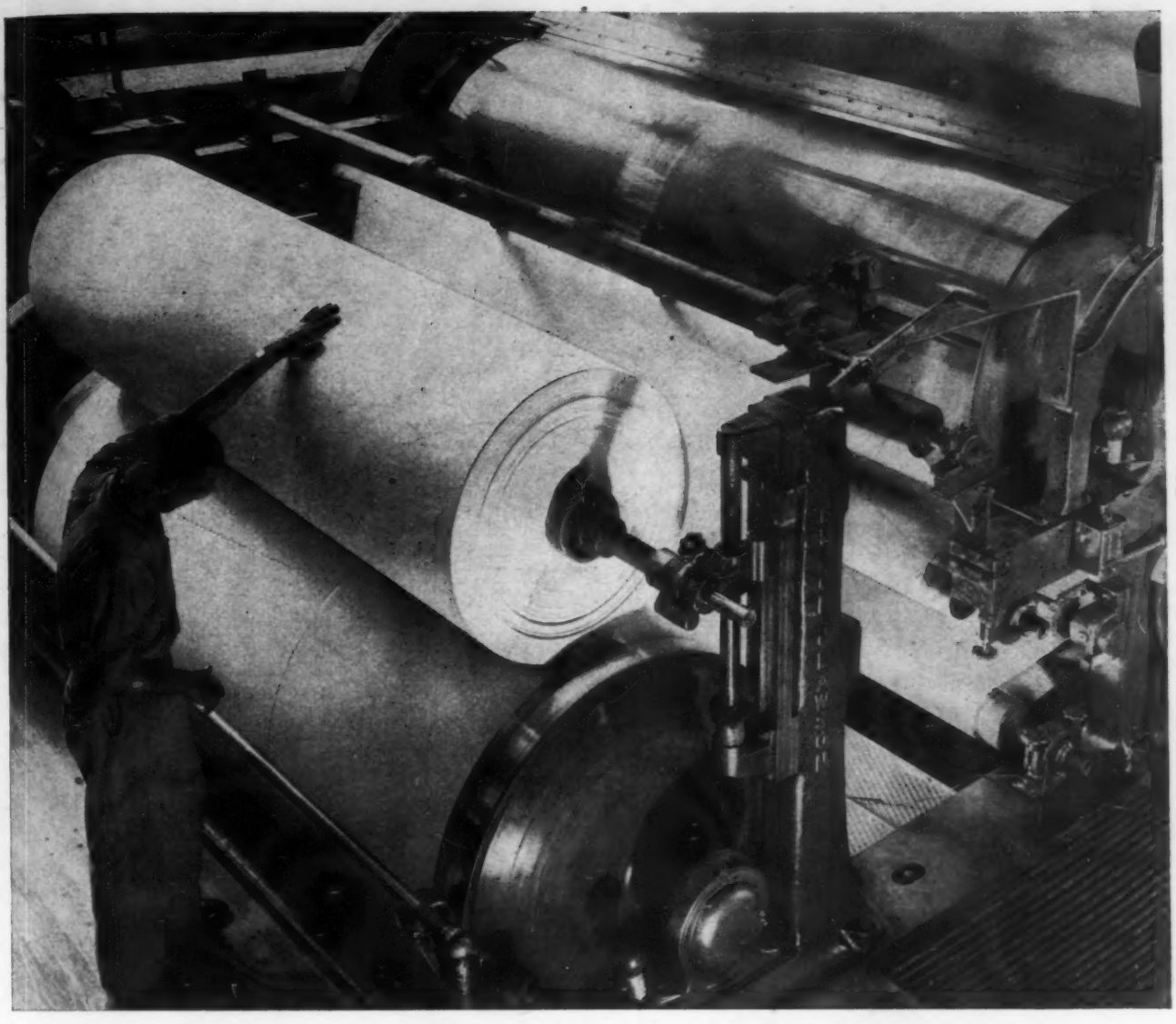
Freight Train Radio (22)

BENDIX RADIO, Towson, Md. The type MRT-2B vhf portable two-way railroad radio unit is designed primarily for use in end-to-end radio operation on freight trains. Overall

size of the equipment illustrated is 11 1/16 by 9 1/2 by 4 7/16 inches and the weight is only a little over 15 pounds. The equipment operates on railroad radio frequencies in the 160-mc region.

ECO Exciter for F-M (23)

COLUMBUS ELECTRONICS, Inc., 229 So. Waverly St., Yonkers, N. Y. The type FMO-428 eco exciter contains a reactance modulator for narrow-band f-m. Output for 80, 40, 20, and 10 meters can be modulated or



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Bendix-Scintilla* Electrical Connectors are precision-built to render reliable peak efficiency—day-in and day-out even under difficult operating conditions. The use of Scinflex—a new Bendix-Scintilla developed dielectric material—makes them vibration-proof, moisture-proof, pressure-tight, and materially increases flashover and creepage distances. Even under extremes of temperature—from −67° F. to +300° F.—their performance is remarkable. Dielectric strength is never less than 300 volts per mil.

The contacts, made of the finest materials, carry maximum currents with the lowest voltage drop known to the industry. Check the list of outstanding features below—then write for detailed information on these truly superior connectors. They belong on every job where there is no compromise with quality.

*TRADEMARK

- Vibration-proof
- Moisture-proof
- Radio Quiet
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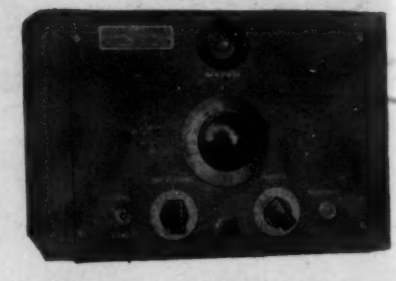
- No Temporary Overloads
- Pressure-tight
- Minimum Weight
- High Arc Resistance
- Easy Assembly and Disassembly
- Low Electrical Resistance

SCINTILLA MAGNETO DIVISION of SIDNEY, NEW YORK — Bendix AVIATION CORPORATION



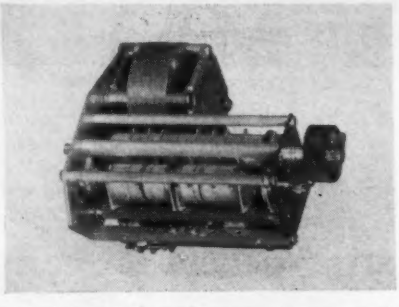
keyed as desired. Write for Bulletin E1.

Electrometer Triode (24)

RAYTHEON MFG. CO., Newton, Mass. The type CK570AX nonmicrophonic electrometer triode is now commercially available. Capable of measuring currents as low as 0.01 micromicroampere, the tube requires only 12.5 milliwatts of filament power. A data sheet including the recently declassified Zeus circuit is available from the Special Tube Section.

Automatic Tuner (25)

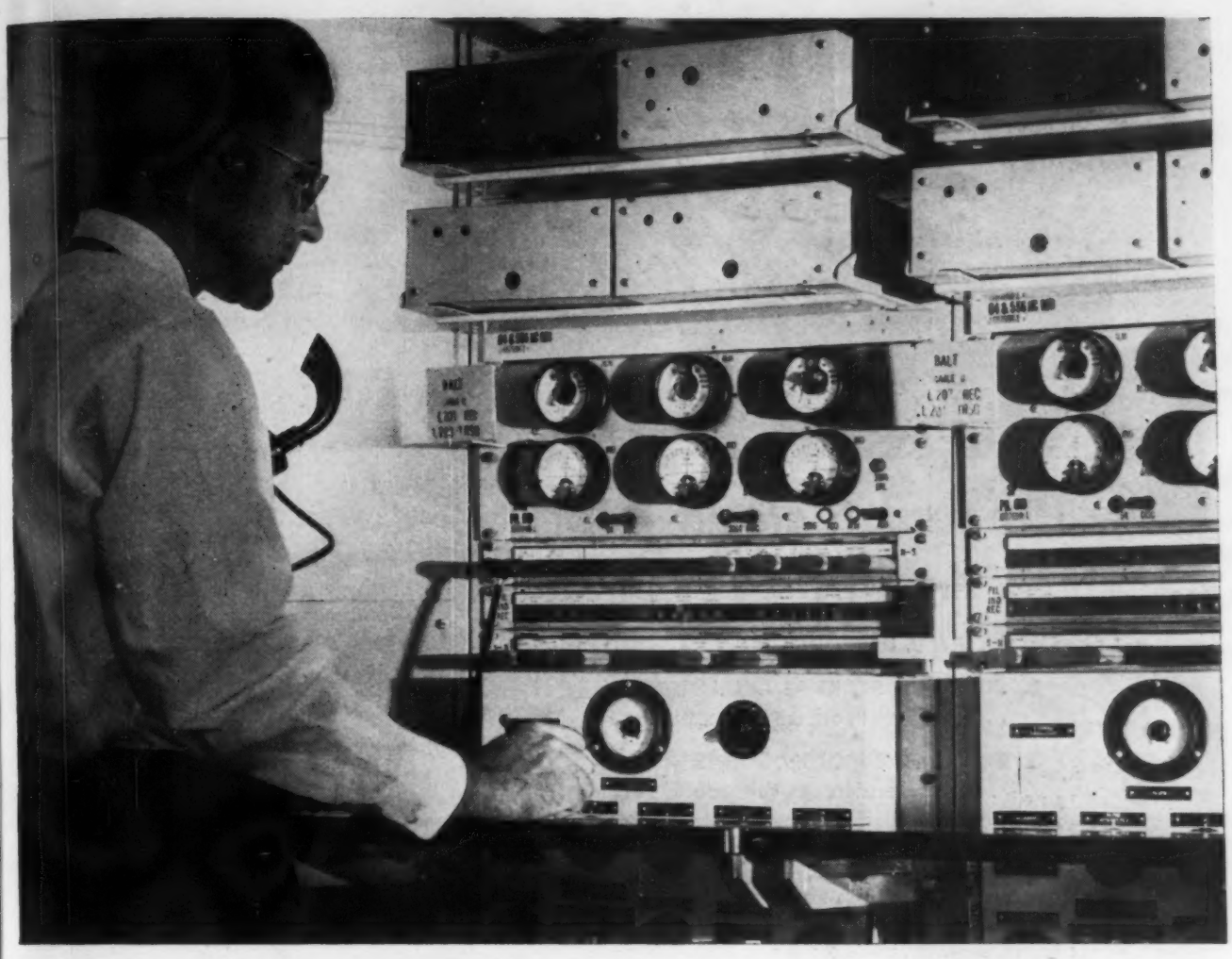
COLLINS RADIO CO., Cedar Rapids, Iowa. The type 496E Autotune unit is a precision tuning device for all radio and industrial applications



requiring an accuracy of one part in 36,000. Output torque is one inch-pound maximum and operating time is less than six seconds.

Marine Communicator (26)

WESTERN ELECTRIC CO., INC., 195 Broadway, New York 7, N. Y. Type 248A marine radiotelephone with dial code selector is distributed by the Graybar Electric Co. The remote control unit illustrated permits rapid operation of telephone service by providing instant selection or changeover from one to any other of the 30 optional operating frequencies between 2 and 20 mega-



At Philadelphia, a testboard man answers as an electronic watchman calls attention to conditions on one of the coaxial systems to Baltimore and Washington.

"Send Help to Manhole 83"

Strung out along every Bell System coaxial cable, electronic watchmen constantly mount guard over your voice. Some are in manholes under city streets; some are in little huts on the desert. Most situations they can deal with; if things threaten to get out of hand, they signal the nearest testboard.

Principal care of the electronic watchman is the transmission level. Sun-warmed cables use up more energy than cold ones, so a transcontinental call may take a millionfold more energy to carry it by day than by night. Each watchman — an electronic regulator — checks the transmission level and adjusts the amplification which sends your voice along to the next point. Many hundreds of regulators may be at work on a single long distance call.

Without automatic regulation, the precise control of energy in the Bell System's long distance circuits would be a superhuman task. So Bell Laboratories, which in 1913 developed the first high vacuum electronic amplifier, went on to devise the means to make them self-regulating in telephone systems. This is one reason why your long distance call goes through clearly, summer or winter.

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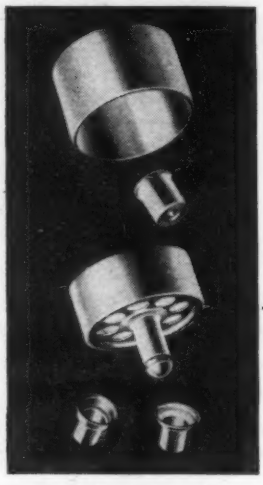
   . . . ends will roll without splitting.
3. **BETTER FOR MOLDED PARTS**
   . . closed end keeps compound out.

If you use pins for vacuum tubes, adapters, fluorescent lamps, plugs, or electrical equipment of any kind, the chances are you'll save time, money and rejections by using these super-smooth, *seamless,* patented Radio Pins. They are available in a wide variety of styles and sizes, with staking end either closed or open. For a quotation, simply send a sketch, sample or description and state the quantity you need.

Radio or Radar Equipment?

In addition to Radio Pins, we produce large quantities of top caps, base shells and adapter shells for vacuum tubes; also a wide variety of other metal products including deep drawn shells and cups, blanks and stampings, ferrules, grommets, washers, vents, fasteners—and, for almost every manufacturing requirement, the world's largest assortment of eyelets.



THE AMERICAN BRASS COMPANY
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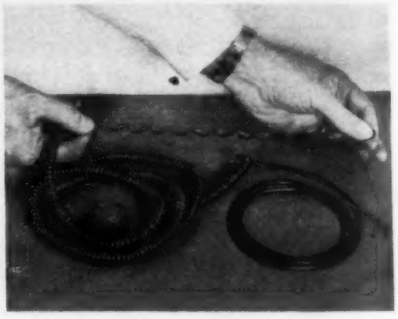
General Offices: Waterbury 88, Connecticut
Subsidiary of Anaconda Copper Mining Company
In Canada: ANACONDA AMERICAN BRASS LTD., *New Toronto, Ont.*

cycles. Three radio receivers facilitate traffic handling or simultaneous monitoring of three stations.

Retractile Cable (27)

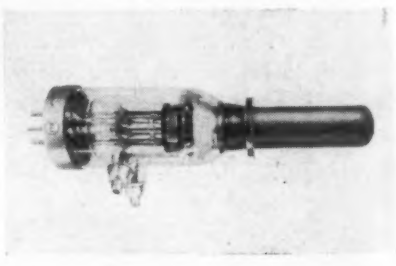
SPECIAL PRODUCTS CO., Silver Spring, Md. Specords are rubber-covered conductors formed in spirals so that they will not kink or snarl in use. A single-conductor shielded microphone cable, for in-



stance, has a retracted length of about 59 inches but stretches to 25 feet.

Thoriated Tungsten Tubes (28)

FEDERAL TELEPHONE AND RADIO CORP., Clifton, N. J. Thoriated tungsten filaments are used in tubes developed for use in 50-kw broad-



NEW! Wilco Ni-Span C*
CONSTANT MODULUS ALLOY

NI-SPAN C is a Titanium-Chromium-Nickel-Iron alloy with the outstanding property of having a constant modulus, i. e. a zero thermoelastic coefficient, over a range of temperature from—50 to + 150°F. Most metals have a negative thermoelastic coeffcient, that is they become elastically weaker or decrease in stiffness with rise in temperature. The important difference between Ni-Span C and other materials, whether of the constant modulus type or otherwise, lies in the fact that Ni-Span C is a precipitation hardening alloy in which the thermoelastic coefficient is adjustable through age hardening heat treatment. This unique combination of modulus control plus the high elastic and strength properties obtained through age hardening makes Ni-Span C a new material having very important commercial applications, the most important being those where constant elastic characteristics are required without the necessity of temperature compensating devices.

Advantages of NI-SPAN C:

1. Higher hardness and higher mechanical properties obtainable through a precipitation hardening heat treatment, than can be obtained with other constant modulus alloys through cold working.

2. An extremely important advantage of Ni-Span C is that the precipitation hardening heat treatment which brings about the marked improvement in physical properties can also be utilized to adjust the thermoelastic coefficient of a given Ni-Span C composition within a certain range by means of appropriate variation in this heat treatment.

Applications:

Watch and instrument hair springs, accurate weighing scale springs, and accurate springs of all kinds, where inherent temperature compensation is required. Diaphragms for pressure sensitive instruments. All types of parts in which extremely low creep or hysteresis values are required. While the controllable modulus property does not hold good beyond 150°F., Ni-Span C has shown very small relaxation in long time tests up to 550°F.

NOW READY! Comprehensive Wilco Data Bulletin on NI-SPAN C, including complete physical and mechanical properties, recommended times for heat treatment and other pertinent information. Send for your copy.

Whatever your requirements for Wilco materials, our engineers will gladly help you meet them successfully.

WILCO PRODUCTS INCLUDE: THERMOSTATIC BIMETAL—Wide temperature ranges, deflection rates and electrical resistivities. CONTACTS—Silver-Platinum-Tungsten-Alloys-Sintered Powder Metal. SILVER CLAD STEEL. JACKETED WIRE—Silver on Steel, Copper, Invar or other combinations requested. ROLLED GOLD PLATE AND WIRE NI-SPAN C New Constant Modulus Alloy. SPECIAL MATERIALS.

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● The wide selection of Webster Electric cartridges offers a cartridge with correct characteristics for your use. Correct weight, response, voltage output and other requirements so necessary for top performance.

Webster Electric cartridges are GOOD cartridges of balanced construction that produce maximum output at designated tracking pressures with minimum distortion and minimum mechanical reproduction. All models offer exceptionally uniform response over the desired range of frequencies, with low distortion and minimum needle noise.

Complete tone arm assemblies of improved design, in a wide selection of models, are also available to incorporate in your models.

All Webster Electric products are carefully designed and manufactured under highest quality standards. They have been on the market for years, and during this time have proved their value for long life and top performance.

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RACINE WISCONSIN
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cast transmitters. The type 9C30 illustrated is designed for r-f amplifier service at frequencies up to 20 megacycles. Filament voltage is 15 volts and current is 135 amperes. Maximum plate voltage is 15,000 volts, and current 8 amperes. Plate input is 120 kw and 40-kw plate dissipation is allowable.

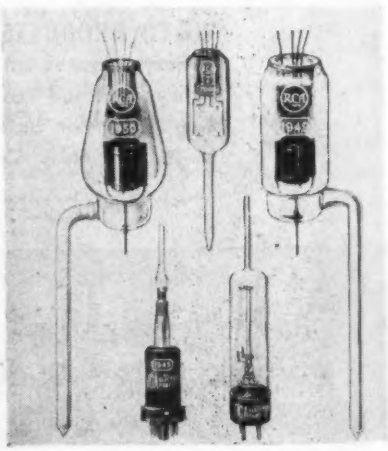
Microwave Diathermy (29)

RAYTHEON MFG. CO., Waltham, Mass. The Microtherm is a microwave diathermy equipment operating at 2.45 kilomegacycles and is intended to produce a rapidly penetrating heat in the human body. Because of its greater effect, less

power is required for this than for conventional equipment and the unit is therefore portable.

Vacuum Gage Tubes (30)

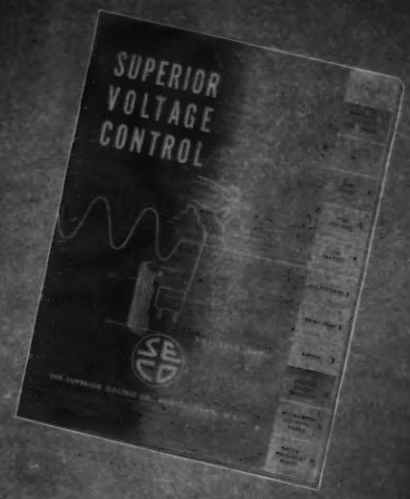
RADIO CORP. OF AMERICA, Camden, N. J. Type 1945 ionization gage tube is used in a new vacuum leak

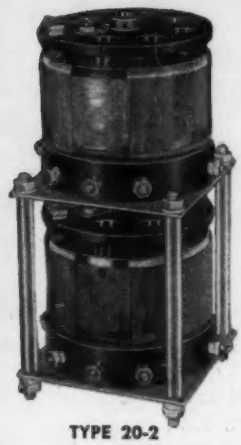


New Bulletin 547

Features VOLTAGE CONTROL by SUPERIOR ELECTRIC

All the information required for selection of the correct voltage control apparatus is included—
FOR EXAMPLE . . .

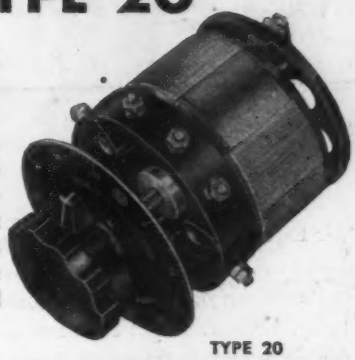




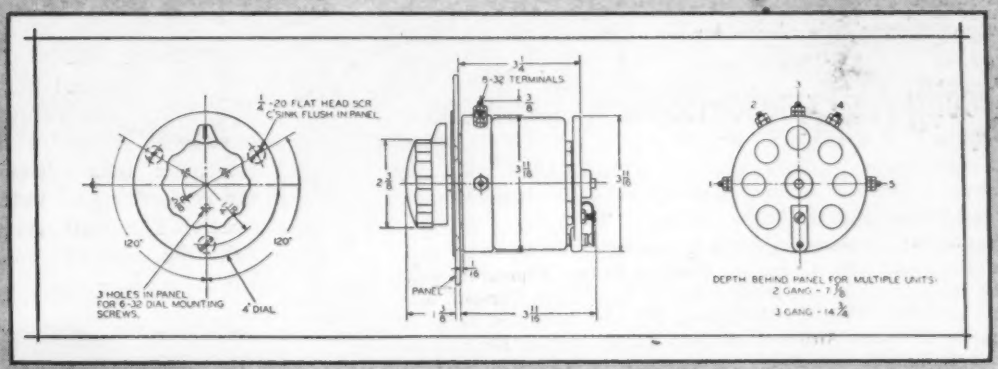
TYPE 20-2

POWERSTAT TYPE 20

Although the smallest in physical and electrical size of all POWERSTAT variable transformers, type 20 possesses all the essentials of superior voltage control equipment. It has excellent regulation . . . high efficiency . . . smooth control . . . and unusually rugged mechanical construction. Its current rating of 3.0 amperes exceeds all other units of comparable mounting dimensions. For three phase operation, type 20 can be ganged for wye or open-delta operation.



TYPE 20


1/4-20 FLAT HEAD SCR C'SINK FLUSH IN PANEL
120°
120°
3 HOLES IN PANEL FOR 6-32 DIAL MOUNTING SCREWS.
4" DIAL
6-32 TERMINALS
PANEL
DEPTH BEHIND PANEL FOR MULTIPLE UNITS:
2 GANG - 7 7/8
3 GANG - 14 3/4

To get your copy of Bulletin 547 . . . for more information on POWERSTAT type 20 and other voltage control equipment write

SUPERIOR ELECTRIC, 1091 Laurel Street, BRISTOL, CONNECTICUT.

THE SUPERIOR ELECTRIC CO.

BRISTOL, CONNECTICUT

POWERSTAT VARIABLE TRANSFORMERS • VOLTBOX A-C POWER SUPPLY • STABILINE VOLTAGE REGULATORS

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It's like having all your suppliers close at hand, when you specify Air Express delivery. Air Express gets you what's wanted in merc hours . . . no matter *where* your suppliers are located.

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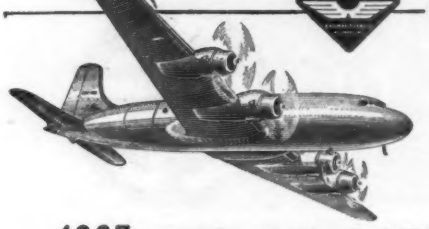

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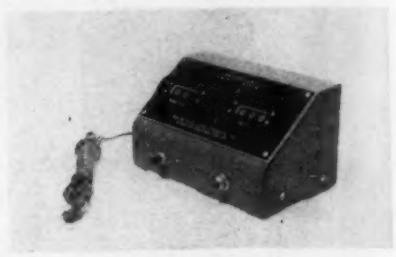
1927— 20TH YEAR OF GETTING THERE FIRST! —1947

detector. Sensitive only to hydrogen, the tube is connected into a vacuum system and a jet of hydrogen is then played over the areas to be checked for leaks. A small amount of the gas increases ion current in the tube to give an indication. Other ionization-type tubes are the type 1949 and type 1950. The 1946 is a thermocouple and the type 1947 is a Pirani gage.

Film Counter (31)

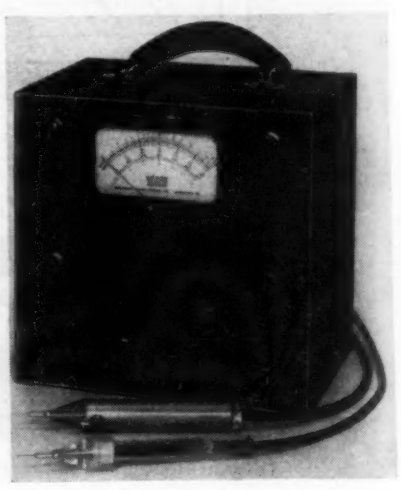
ARLINGTON ELECTRICAL PRODUCTS, INC., 18 West 25th St., New York 10, N. Y. A counter for use in motion picture viewing, dubbing and other operations can be remotely



located from projector, recorder or dubbing head and will read elapsed time to 0.1 minute. The unit illustrated can be furnished for 35 or 16-mm film.

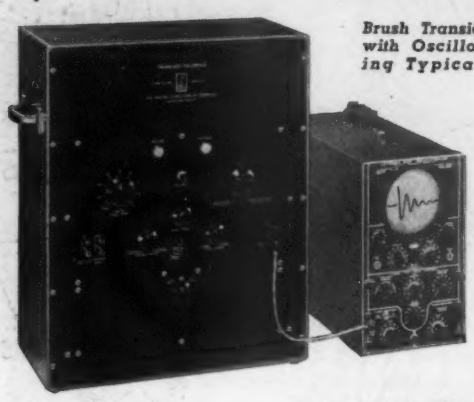
VTVM (32)

ELECTRONIC MANUFACTURING CO., 140 South Second St., Harrisburg, Pa., Model 110 vacuum tube volt-ohmmeter is useful for servicing television, transmitters, f-m, regular receivers and industrial appliances. The unit reads to 15,000 volts d-c, and 300 volts, a-c up to



Now study Transient Phenomena with Magnetic Recording!

Brush Transient Recorder



Brush Transient Recorder with Oscilloscope Showing Typical Transient

automatically records transient phenomena of less than 2/10 second duration—either electrical or capable of being picked up by electrical gauges.

Some typical phenomena which may be recorded by the Brush Transient Recorder include spot welding currents, light flashes, arc discharges, impact loads, etc. These results are accomplished by magnetically recording on a rapidly moving steel tape a frequency modulated carrier. Reset button clears tape and prepares it for re-use.

Brush representatives will be glad to discuss applications of this instrument in solving your particular laboratory or production problems. For complete details write today for further information.

The Brush Development Co.

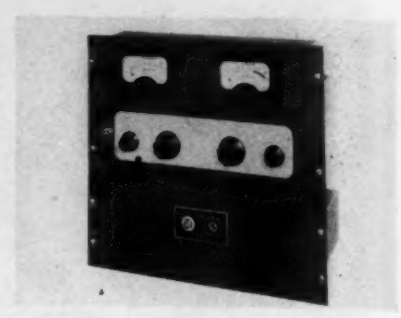
3415 Perkins Avenue • Cleveland 14, Ohio

Canadian Representatives: A. C. Wickman, (Canada) Ltd.
P. O. Box 9, Station N, Toronto 14, Canada

300 megacycles. Critical resistors have a heat loop and are fungus-proofed.

Audio Peak Limiter (33)

THE LANGEVIN CO., INC., 37 West 65th St., New York 23, N. Y. Type 119-A Progar unit is a fast acting peak limiter preceded by an automatic gain control amplifier with variable time constants. In broad-



cast station applications the circuits can be employed to insure a higher average percentage of modulation than can be obtained with a limiter alone. A catalog sheet is available describing the equipment and its use.

Indicator Lights (34)

AMERICAN ELECTRONICS CORP., 226 No. Fourth St., Columbus, Ohio, announces a line of standard neon-

glow lamp indicators. The units are composed of a high heat-resisting outer housing provided with suitable mounting bracket that encases a neon-glow lamp and proper resistance for operating on voltages of 70 volts and upward, a-c or d-c.

Oil Vapor Vacuum Pump (35)

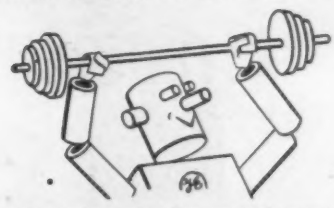
LITTON ENGINEERING LABS., San Carlos, Calif. Model 250 series oil

USE G-E MYCALEX INSULATION

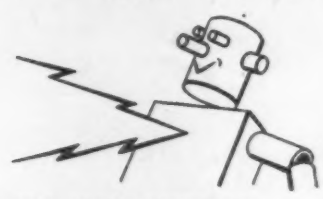
For HIGH HEAT RESISTANCE

Plus these 5 Insulation Advantages...

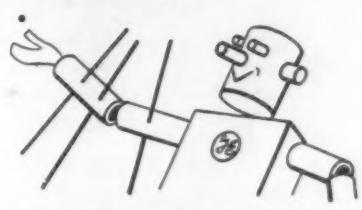
● HIGH DIELECTRIC STRENGTH



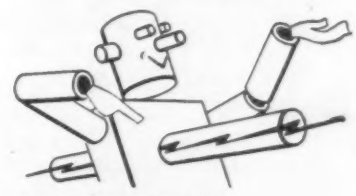
● HIGH MECHANICAL STRENGTH



● HIGH ARC RESISTANCE



● FIRM BOND TO METAL INSERTS



● LOW LOSS FACTOR

● General Electric mycalex is an exceptional insulation material because of its unique combination of properties. Use *G-E* mycalex when insulator designs call for a firm bond to metal inserts *plus* excellent resistance to heat and arcing, high dielectric and mechanical strength, and a low loss factor.

G-E mycalex is a gray, stone-hard compound of glass and mica that can be ordered in standard rods and sheets. Or it can be molded or fabricated to your specifications.

General Electric has complete molding and fabricating facilities for producing *G-E* mycalex parts in any quantity. Let General Electric's mycalex specialists fabricate sample parts for you to test. After testing, your designs can be converted to the speediest, most economical molding processes. To get the complete story, send for the booklet, "G-E MYCALEX". Write to Section S-21, Plastics Division, Chemical Department, General Electric Company, 1 Plastics Avenue, Pittsfield, Mass.

GENERAL ELECTRIC

CD47-M21

MYCALEX

For SIMPLE CONVERSION to A-C OPERATION

SEVERAL G-R instruments, equipped with a battery block to secure the maximum portability for field use, are often operated continuously for many hours in the laboratory, in testing and in production. For such uses an a-c operated power supply to replace the battery is a great convenience.

The Type 1261-A Power Supply was designed for this use. Mechanically and electrically it is interchangeable with a BA48 battery, supplying the same voltages and current. It is interchangeable with the BA48 battery in the Type 759-A or -B Sound Level Meter, the Type 720-A Heterodyne Frequency Meter and the Type 1231-A Amplifier & Null Detector.

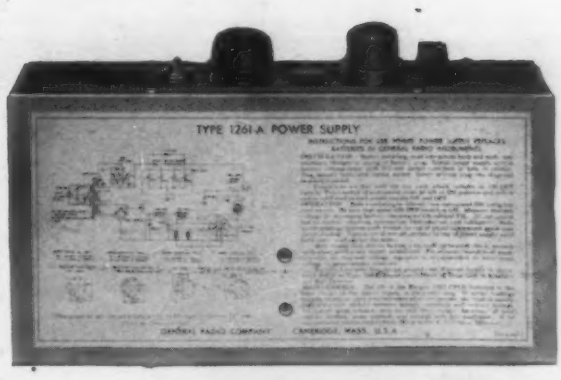
It also can be used as a general-purpose substitute for a BA48 battery. Its output is: FILAMENT SUPPLY: 1.5 or 3.0 volts up to 350 ma; PLATE SUPPLY, when used on a 115-volt 60-cycle line with normal filament current of 300 ma: 133 volts, open circuit; 107 volts at 3 ma; 89 volts at 5 ma; 72 volts at 7 ma; maximum output current of 8 ma.

A selenium rectifier and L-C filter with two flashlight cells floating across the output provide a low-impedance, well-filtered and regulated d-c filament supply. A relay opens the circuit when the instrument is not in use so that the life of the batteries is equivalent to their shelf life.

A conventional vacuum-tube rectifier and R-C filter provide the plate supply. A 4-terminal output socket fits the plug on the battery cable of the instruments which use the BA48 battery; octal selector plugs inserted into a socket on the power supply make it possible to select filament and plate voltages for various needs.

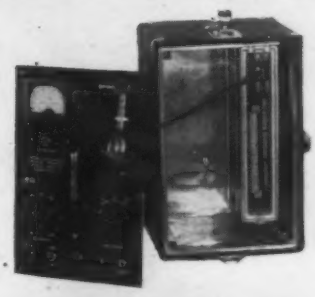
IMPORTANT — When ordering be sure to specify type and serial number of any of the three G-R instruments with which the supply is to be used; otherwise we will ship a selector plug which the user can wire to meet his individual requirements.

TYPE 1261-A POWER SUPPLY $95.00



TYPE 759-B SOUND-LEVEL METER

For all types of Noise Measurements, Accurate, very sensitive, anyone can operate.

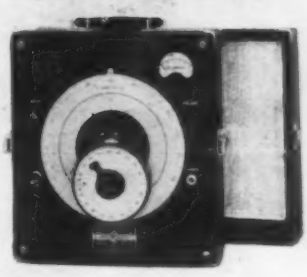


TYPE 1231-A AMPLIFIER & NULL DETECTOR

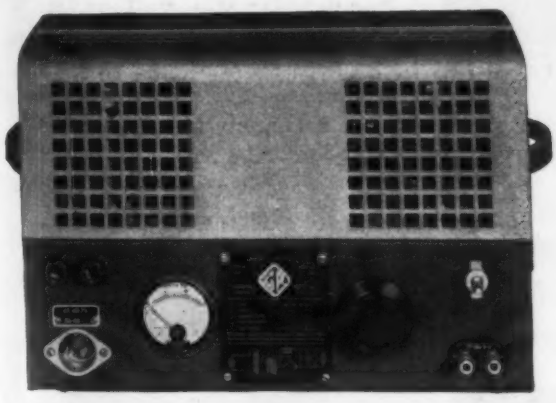
A high gain general-purpose laboratory amplifier and a sensitive visual null detector for bridge measurements.

TYPE 720-A HETERODYNE FREQUENCY METER

For measurement of frequencies by harmonic methods between 10 and 3,000 megacycles.



A-C SUPPLY TO REPLACE STORAGE BATTERIES



● The Type 1260-A VARIAC*-RECTIFIER effectively replaces the usual storage battery for general laboratory use. It is a convenient variable d-c voltage source for use in many types of experimental circuits for operation from 115-volt, 60-cycle lines.

Essentially this supply consists of a transformer with VARIAC*-controlled input, a selenium rectifier, an output filter and a d-c output meter.

The output voltage range is 0-10 volts at 4 amperes; maximum power of 40 watts; maximum current 4 amperes; no-load voltage 15 The power input at full 40-watt load is about 75 watts; the hum voltage at 10 volts, 4 amperes, is less than 100 millivolts or 1% of the output voltage, when the supply is operated on a 60-cycle line.

This supply is a very convenient unit for operating battery-driven equipment at voltages up to 10 volts. The ability to set the d-c voltage at any desired value between 0 and 10 is an added advantage in many cases.

TYPE 1260-A VARIAC*-RECTIFIER $125.00

*Reg. U.S. Pat. Off.

ORDER NOW — Delivery of BOTH Probably from STOCK

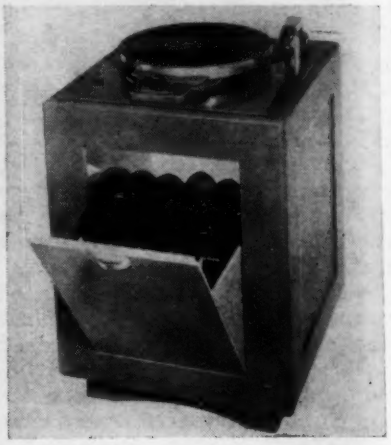
GENERAL RADIO COMPANY Cambridge 39, Massachusetts

90 West St., New York 6 920 S. Michigan Ave., Chicago 5 950 N. Highland Ave., Los Angeles 38

vapor vacuum pumps are constructed entirely of steel and are water cooled to insure proper action in hot places. Boiler and charcoal baffles are easily demountable for cleaning purposes. Complete characteristics, prices, and dimensions are available on request.

Recording Console (36)

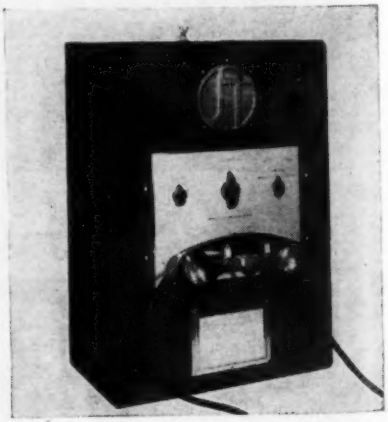
REK-O-KUT Co., 146 Grand St., New York, N. Y. A new console for recording or transcription equipment



has a drop-front door and pockets for holding approximately 100 sixteen-inch recordings. Four screw jacks are provided for leveling the console.

Marine Radiophone (37)

JEFFERSON - TRAVIS, INC., 111 Eighth Ave., New York 11, N. Y. Model 351 marine radiotelephone for 12 or 32-volt operation comprises 5 crystal-controlled channels for transmitting and receiving covering the complete 2 to 3 mc marine band. The entire unit is housed in



What's your Fine Wire problem?

Problem 1

Mr. Plate had miles of wire but he needed it gold plated. Just a call to Fine Wire Headquarters and we agreed to add the protection plus the fine appearance of gold. If you have a plating problem, we have the plating facilities.

Problem 2

Mr. Fine had to have some .001 copper wire to fill an important order. So Fine Wire Headquarters transformed his coarse copper wire to .001. We can do the same for you if you have wire that's too big.

Problem 3

Mr. Insulate needed enamelled fine wire. So he ordered from Fine Wire Headquarters, and received fine enamelled copper wire made to his specifications.

the answer

WHY not call Fine Wire Headquarters when you have a question about fine wire? We can't do the impossible, but we can do lots of things that can bring you the right fine wire for the job. So—when you have a Fine Wire problem—wire, phone or write to North American Philips, makers of NORELCO Fine Wires, and ELMET Tungsten and Molybdenum products.

NORTH AMERICAN PHILIPS COMPANY, INC.

Dept. **E-9,** 100 East 42nd St., New York 17, N. Y.

CORNISH

WIRES

at work

Courtesy HURLEY Machine Division, makers of GLADIRONS

WHY ARE CORNISH WIRE PRODUCTS SPECIFIED BY THIS LARGE MANUFACTURER OF IRONING MACHINES?

Because their ENGINEERING Department knows by test that they will give faithful and enduring performance . . .

Because their PRODUCTION Department finds that they have those qualities essential for quick installation on their assembly line . . .

Because their PURCHASING Department realizes that these Quality Products, backed by dependable service, are sold at prices that spell true economy . . .

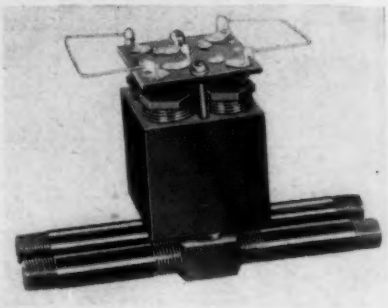
CORNISH WIRE CO., INC.

15 Park Row • New York City, 7

a steel cabinet 21 × 15¼ × 10 inches.

Gas Observing Unit (38)

GOW-MAC INSTRUMENT Co., 22 Lawrence St., Newark 5, N. J. The new thermal conductivity unit provides an accurate and inexpensive means for observation, analysis and testing of gases. It incorporates four filaments assembled in appropriate chambers and wired in a balanced bridge arrangement. Compact in size, the unit is specially constructed for high sensitivity, long life and stabliity.



Improved Headset (39)

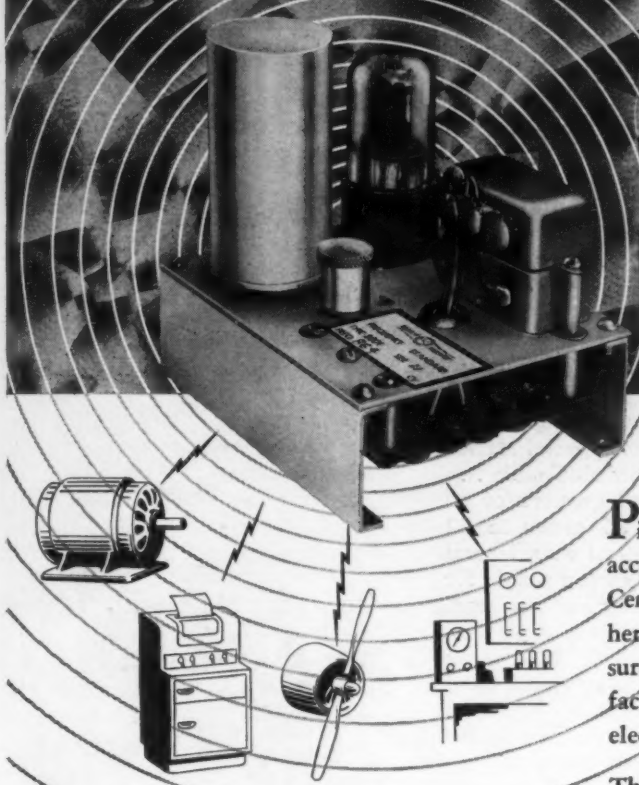
TELEX, INC., 1370 Northwestern Bank Bldg., Minneapolis 2, Minn. The Monoset has recently been improved by the addition of a volume control. The lightweight underchin construction of the headset makes it easy to wear for long periods.

Volt-Ohm-Milliammeter (40)

PRECISION APPARATUS CO., INC., 92-27 Horace Harding Blvd., Elm-

PICK A NUMBER

ANY FREQUENCY FROM 40 TO 1,000



Pictured here is a tuning-fork frequency standard with accuracy guaranteed to one part per million per degree Centigrade. The fork is temperature-compensated and hermetically sealed against variations of barometric pressure. This standard, when combined with basic equipment, facilitates accurate speed and time control by mechanical, electrical, acoustical or optical means.

The unit is available separately or in conjunction with complete timing instruments. Our engineers are ready to cooperate on any problem.

MOTORS · FACSIMILE · AIRCRAFT · LABORATORIES

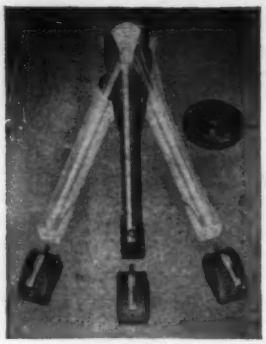
American Time Products, Inc.

580 Fifth Avenue New York 19, N. Y.

OPERATING SUBSIDIARY OF THE WESTERN ELECTRIC COMPANY

PARA-FLUX REPRODUCER

gives less wear on record . . . lighter impact of stylus . . . and improves a well-known tone quality



Para-Flux Reproducer with interchangeable Heads:
Universal . . .
Lateral only . . .
Vertical only

It's the low mechanical impedance designed into the improved PARA-FLUX . . . the special refined metals and other components now obtainable . . . that enable reducing the record pressure of all R-MC Reproducer Heads from 35 grams to 22 grams. And all three types: Vertical only, Lateral only, and Universal maintain the correct weight for permitting the pressure of 22 grams on the record. From our knowledge, we believe that PARA-FLUX Vertical only and Universal are the only heads obtainable today, which operate on commercial service at a pressure of 22 grams. This improved feature means less wear on records, and lighter impact of stylus when inadvertently dropped.

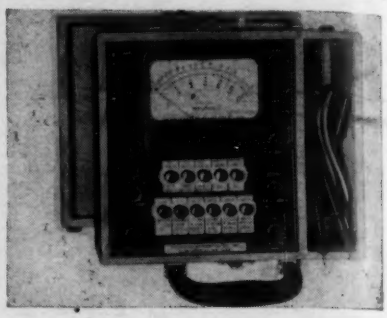
R-MC engineering skill applied to reproducer design gives all the advantages that discriminating users demand: More realistic reproduction of transcriptions . . . a reproducer of precision-build, sturdy construction, with finest materials obtainable . . . embodying up-to-the-minute features, including convenient finger lift for preventing slipping of Reproducer when lifted off record. A highly polished aluminum alloy center-piece of tone arm and head enhances the attractive design of Reproducer.

This new lightweight Head, either Vertical only, Lateral only, or Universal, functions correctly with all R-MC Tone Arms now in service. Therefore it is not necessary to change arm in service when ordering the new Head.

Whenever you may need a new PARA-FLUX Head, your R-MC Jobber will supply you with the new lightweight head . . . immediately . . . in exchange for your old one, in accordance with our standard replacement policy and exchange price.

Available through authorized jobbers. Illustrated Bulletin R-6 on request.

RADIO-MUSIC CORPORATION
EAST PORT CHESTER • CONN.



hurst, L. I., N. Y. The series 858 pushbutton meter has a resistance of 20,000 ohms per volt and ranges including 6,000 volts, 600 megohms, 12 amperes, 70 db and 60 microamperes. The model 858-P illustrated has a hardwood case with hinged cover for portable use. Laboratory users may prefer the model 858-L.

Carrier Systems (41)

F. W. LYNCH Co., San Francisco, Calif. Rack-mounted, multichannel carrier telephone equipment can be

used to expand present wire telephone facilities of industrial and railroad users. With repeaters, operation of over 1,000 miles is feasible.

Dual Capacitor (42)

CORNELL-DUBILIER ELECTRIC CORP., South Plainfield, N. J. The type BRL-2215SS dual 20-microfarad capacitor is equipped with two leads for each unit making the component more versatile than the con-

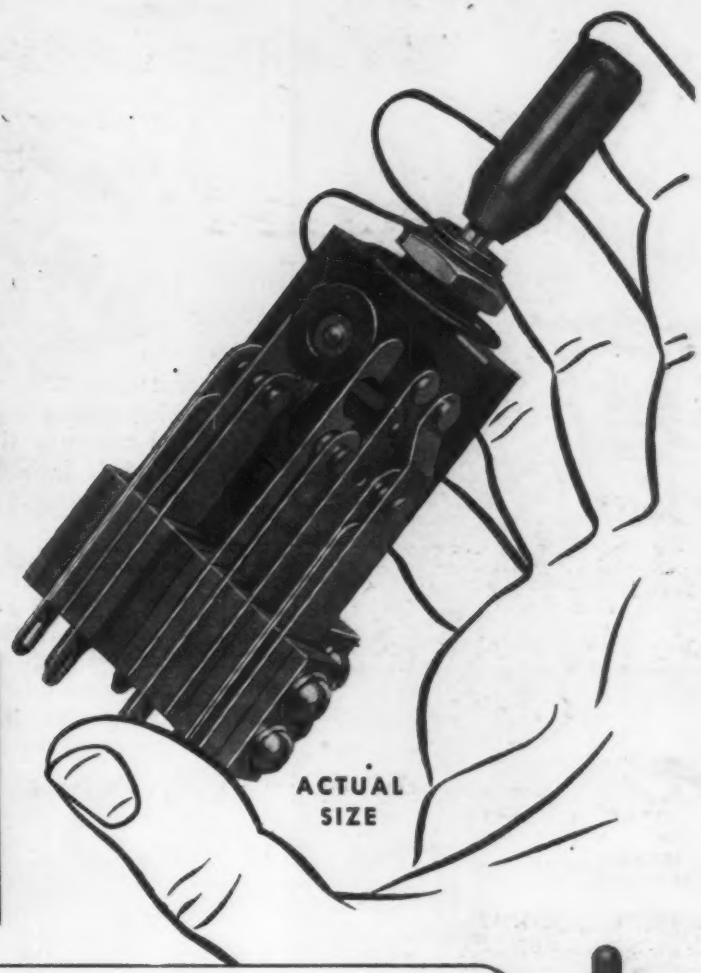
GENERAL CONTROL BUILDS

The world's most versatile SWITCH

MODEL

MCM

WITH PATENTED* DETENT INSERTS



General Control Company announces a new feature for its Model MCM "Midget" lever switches—a feature which makes possible true *custom-built* switches from *standard* parts. By means of small, stainless steel inserts, shown above, the MCM "Midget" can be provided with a wide variety of detent actions to meet the most exacting requirements. Light or heavy action, two or three positions, locking or non-locking and special intermediate stops—all are possible with this versatile detent mechanism. The inserts are riveted in permanently at the factory. They preserve the original "feel" of the switch throughout its life—tests show negligible wear after 1,000,000 operations!

Write for details on the complete line of General Control lever switches, or, better yet, send us your switching requirements and take advantage of our engineering service.

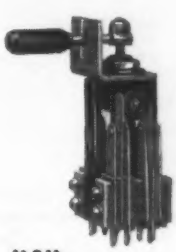
*PATENT NO. 2,411,086

OTHER MCM FEATURES

- ★ Single hole mounting for easy panel fabrication.
- ★ Removable contact assembly for easy wiring.
- ★ Light weight—3½ oz. with 12 contact blades.
- ★ Small size—depth behind panel only 2¾ inches.
- ★ High capacity—5 amperes non-inductive at 115 volts AC.
- ★ High dielectric strength—tested at 2500 volts AC to ground.
- ★ Versatility: Can be supplied with angle mounting bracket when space behind panel is limited. Rotary actuator available. Waterproof handle assembly for marine or aircraft applications.

MCM
With Waterproof Handle Assembly

MCM
With Rotary Actuator



MCM
With Angle Mounting Bracket

GENERAL CONTROL COMPANY

1202 SOLDIERS FIELD ROAD, BOSTON 34, MASSACHUSETTS

Designed for

Application

The No. 92101—Antenna Matching Preamplifier

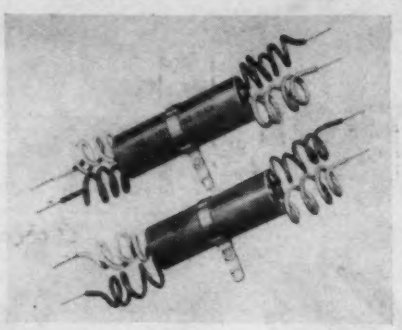
The Millen 92101 is an electronic impedance matching device and a broad-band preamplifier combined into a single unit, designed primarily for operation on 6 and 10 meters. Coils for 20 meter band also available. This unit is the result of combined engineering efforts on the part of General Electric Company and the James Millen Manufacturing Company. The original model was described in G. E. Ham News, November-December, 1946. The No. 92101 is extremely compact, the case measuring only 6¼" x 5¾" x 3". The band changing inductor unit plugs into the opening in the front of the panel. Plug is provided for securing power requirements for the 6AK5 tube from the receiver. Coaxial connectors are furnished for the antenna and receiver connections.

JAMES MILLEN MFG. CO., INC.

MAIN OFFICE AND FACTORY

MALDEN

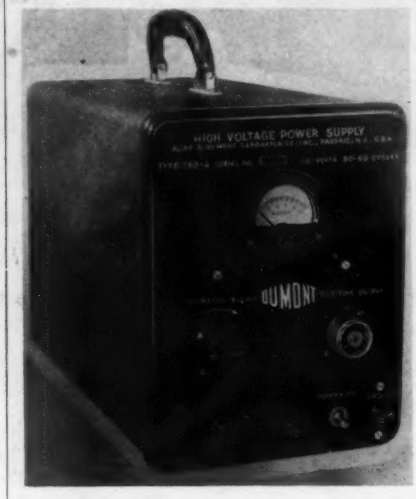
MASSACHUSETTS



ventional type. The capacitor has a working-voltage rating of 150 volts and measures 1⅛ inch in diameter and 2⅞ inches long.

Oscillograph Power Pack (43)

ALLEN B. DU MONT LABORATORIES, Inc., 2 Main Ave., Passaic, N. J. The type 263-A high-voltage power supply comprises an r-f oscillator, step-up transformer, half-wave rectifier, filter and metering system.



Continuously variable from 5,000 to 10,000 volts with loads up to 200 microamperes the equipment can be used to modernize cathode-ray oscillographs for high-writing rates.

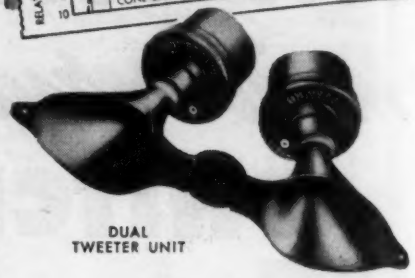
R-F Exciter (44)

COLLINS RADIO CO., Cedar Rapids, Iowa. Three new r-f exciter units

FOR HIGH FIDELITY AT LOW COST!

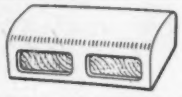
-the UNIVERSITY TWEETER

RESPONSE WITH CONE SPEAKER AND UNIVERSITY TWEETER
FREQUENCY IN CYCLES PER SECOND
10 50 100 1000 5000 10,000 15,000 20,000
RELATIVE LEVEL IN DB
10 5 0 5 10
RESPONSE OF TWEETER (MAX RESPONSE LEVEL SHOWN)
RESPONSE OF AVERAGE 12" CONE SPEAKER



DUAL TWEETER UNIT

The reproduction of music and voice with breath-taking realism, is now possible with the new UNIVERSITY Dual Tweeter. Used in conjunction with any standard 12" cone speaker in FM and AM radio equipment and wide range phonograph amplifiers, it adds the brilliant "highs" so frequently carried through all stages of amplification, only to be lost in the bottleneck of a single unit reproducer. Frequency response is 2,000 to 15,000 cycles. The die-cast dual horn design offers wider dispersion angle than the conventional single cellular horn—horizontal distribution is 100°, vertical distribution 50°. A high pass filter with auxiliary high frequency volume control, permits easy connection by merely attaching two wires to the existing speaker. Compact dimensions require a mounting space only 2¾" high x 9½" wide. Power handling capacity of the dual unit is 16 watts. For complete information write today to UNIVERSITY LOUDSPEAKERS, INC. 225 Varick St., New York 14, N . Y.



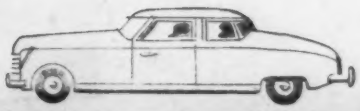
CABINET MODEL CONTAINS DUAL TWEETER UNIT

SINGLE UNIT TWEETER ONLY 2¾" DIAMETER

CLUSTERED TWEETERS FOR WIDE VERTICAL AND HORIZONTAL DISPERSION

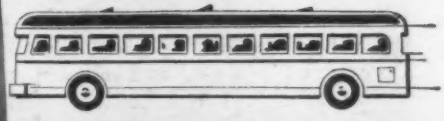
University Loudspeakers

If you manufacture...

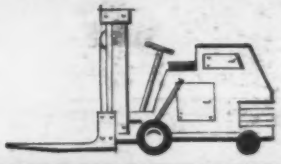


Automobiles

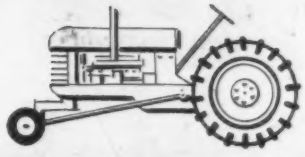
Trucks and Trailers



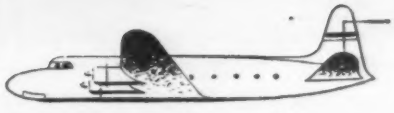
Busses and Trackless Trolleys



Industrial Electric Trucks



Tractors



Aircraft

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Essex "Packaged" Wiring Harness

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Scores of manufacturers have found that they save *time, trouble and money* by turning their electrical wiring harness problems over to Essex specialists.

Essex *One-Source* service handles the intricate job of producing lighting, ignition and control harness assemblies *custom-built* to your exact specifications and *complete* with all manual and electrical control devices for quick, efficient installation.

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Warehouses* and Sales Offices: *Atlanta, Georgia; *Boston, Mass.; *Chicago, Ill.; Cleveland, Ohio; Dallas, Texas; Dayton, Ohio; *Detroit, Mich.; Kansas City, Mo.; *Los Angeles, Calif.; Milwaukee, Wisc.; *Newark, N. J.; Philadelphia, Pa.; St Louis, Mo.; *San Francisco, Calif.

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Type 5530

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Type 5541

THEY'RE TOPS FOR FM!

Western Electric
FORCED AIR COOLED FM TRANSMITTING TRIODES

Their filaments are of thoriated tungsten — most efficient emitter for power tubes of these ratings. Their brazed and welded metallic joints, Kovar-to-glass seals, protected metallic vacuum "seal-off," and self-supporting filament structure all contribute to long, dependable service. Their terminal arrangements permit greatest flexibility of application. For full details, write to Graybar Electric Company, 420 Lexington Avenue, New York 17, N. Y., or . . .

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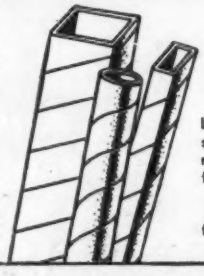
BROADCAST REPRESENTATIVE

*Division of Bendix Aviation Corporation

● This is typical of the wide use of PARAMOUNT paper tubes by leading manufacturers of electrical, radio and electronic products. With over 15 years of specialized experience, PARAMOUNT can produce exactly the shape and size tubes you need for coil forms or other uses. Square, rectangular, or round. *Hi-Dielectric, Hi-Strength,* Kraft, Fish Paper, Red Rope, or any combination, wound on automatic machines. Tolerances plus or minus .002". Made to your specifications or engineered for *you.*



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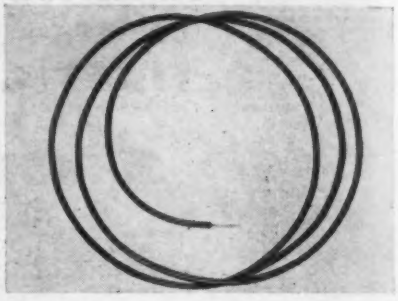
PARAMOUNT PAPER TUBE CORP.
616 LAFAYETTE ST., FORT WAYNE 2, INDIANA
Manufacturers of Paper Tubing for the Electrical Industry

for radio amateurs are now in production. The type 310C-1 and 310C-2 units serve as variable-frequency oscillators, while the 10B exciter illustrated has an output of 15 watts. Calibration is directly in frequency and is accurate within 0.015 percent. Frequencies between 3.2 and 4.0 mc and those in harmonic relation up to the range 25.6 to 32 mc are available.

Aircraft Antenna Wire (45)

FEDERAL TELEPHONE AND RADIO CORP., Clifton, N. J. An insulated aircraft antenna wire that largely overcomes some of the undesirable



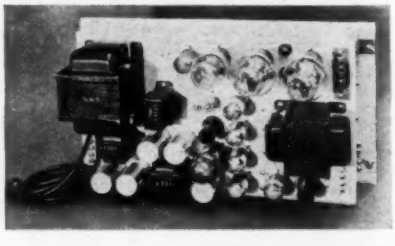
effects of precipitation static in aircraft communication is designated Intelin K-1064. A copperweld wire is covered with an insulating jacket of polyethylene, that helps to limit corona discharge.

Ribbon Microphone (46)

AMPERITE CO., 561 Broadway, New York 12, N. Y. A new ribbon microphone has a frequency range from 40 to 14,000 cycles within plus or minus 3 db. Pickup angle front and rear is 120 degrees. Standard equipment includes switch, 25-foot cable and plug.

Amplifier Units (47)

FAIRCHILD CAMERA AND INSTRUMENT CORP., 88-06 Van Wyck Blvd., Jamaica 1, N. Y. The Unitized Am-



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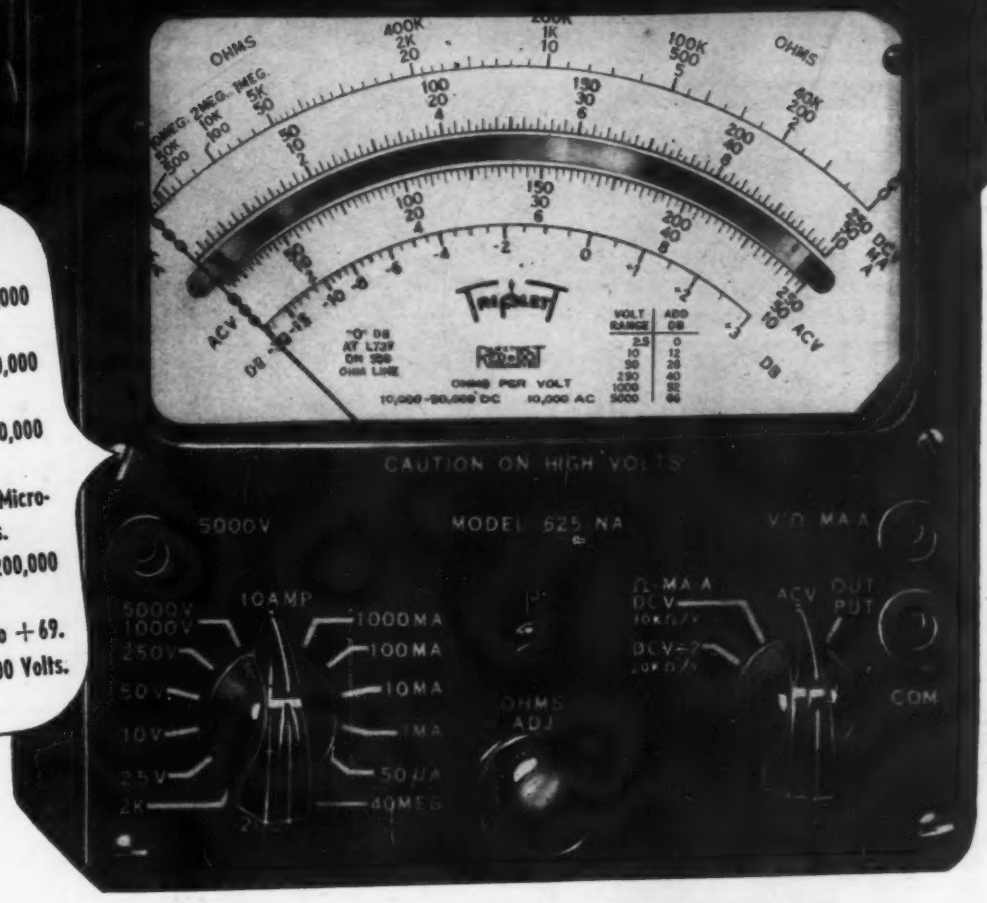
625 NA* VOLT-OHM-MIL-AMMETER



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Six D.C. Volts to 2,500 at 20,000 Ohms per Volt.

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Six Current Ranges: 0-50 Microamperes to 0-10 Amperes.

Three Resistance 0-2000-200,000 Ohms; 0-4 Megohms.

Six Decibel Ranges: —30 to +69.

Six Output Ranges to 5,000 Volts.

*High Ohms—Mirror Scale—Thirty-Nine Ranges

For the Man Who Takes Pride in His Work

The new Model 625NA, with 39 ranges and many added features, is the widest range tester of its type. Note the long mirror scale on the large 6" meter for easier more accurate reading. Resistance ranges to 40 megohms give you all the ranges needed for general servicing, plus Television and FM. And with 10,000 ohms per volt A. C. you can check many audio and high impedance circuits where a Vacuum Tube Volt meter is ordinarily required. A proven super-service instrument.

Write for details today about Model 625NA and the many other new Triplett testers. Address Dept. E97

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Precision first ...to last

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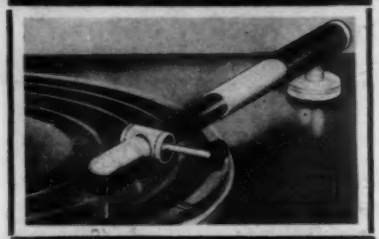
VIBROMASTER TYPE K
PROFESSIONAL ARM



for General Electric Variable Reluctance Pick-ups Or the New Pickering 120M Cartridge

Resonance characteristic flat plus or minus ½ Db 40 to 1000 cycles (measured in 10 cycles steps throughout resonant range).

22 grams pressure when used with General Electric variable reluctance cartridge.

Floating bearing coil spring maintains pressure even under abnormal ambient changes.

"Morning After" handle prevents record damage when cueing. Pick-up head will not move until entire weight is lifted.

Metal bearing race is constructed that bearing pin actually floats free.

Universal "One-hole" mount permits installation in any turntable, with hairline height adjustment.

Permanent stop prevents overswing to injure connections. Internal loop prevents damage.

For any lateral recording up to and including 16 inch, 78 or 33⅓ rpm.

Arm will not "walk" even on 30 cycle modulated grooves.

Precision made and individually serial numbered for constant reference.

Finish is brushed aluminum and black to match any existing turntable.

Tests indicate that in conjunction with a normal General Electric Cartridge and a good equalizer, overall response will be exactly as the record is cut. The pick-up arm will neither introduce nor amplify lateral distortion and has no self-vibration.

PRICE ..... net less cartridge $34.50
GENERAL ELECTRIC CARTRIDGE ..... $ 4.77
PICKERING CARTRIDGE $15.00
GENERAL ELECTRIC SPX-001 PRE AMPLIFIER ..... $ 5.95

GE Electronic Switch, Type YE-9. This instrument was designed for special electrical studies of wave-form, phase, frequency relationship, etc. It will show the simultaneous observation, for comparison, of two or more independent signals on the screen of a cathode-tube oscilloscope. $59.95

Note: All prices are Net, F.O.B. N.Y.C. and are subject to change without notice.

Telephone: LO. 3-1800

hrc HARVEY RADIO COMPANY INC.

103 West 43rd St., New York 18, N. Y.

plifier System consists of 12 units, several of which are currently in production. These units are designed to fit together in a variety of ways so as to serve the needs of the sound recording industry. The 620 power amplifier illustrated and the 621 microphone preamplifier are representative of the line, with essentially flat response from 30 to 15,000 cycles, low noise level, and distortion of less than 5 percent at full output.

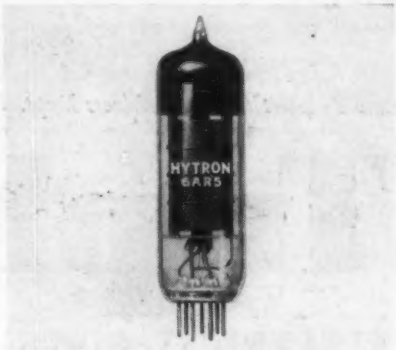
Wire Recorder Mechanism (48)

WEBSTER-CHICAGO CORP., 5610 Bloomingdale Ave., Chicago 39, Ill. The model 79 mechanism operates from socket power and records at 2 feet a second, rewinding at seven

times that speed. A standard spool permits continuous recording for an hour. Shorter lengths are also available. The basic unit can be used by amateurs or experimenters.

Miniature Pentode Power Amplifier (49)

HYTRON RADIO AND ELECTRONICS CORP., Salem, Mass. The type 6AR5 tube is a miniature version of the type 6K6GT. It will be found useful in applications where low heater power and low plate current drain are important considerations. The lower heat dissipation makes the



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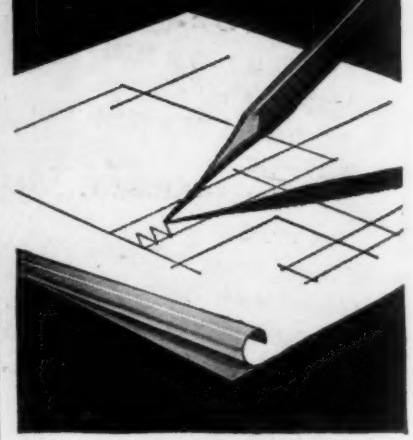

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American Condenser Co.
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3633 S. Racine Ave.

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FORT WAYNE, IND.
Essex Wire Corp.
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HOUSTON, TEXAS
Navigation Instrument Co., Inc.
P. O. Box 7001, Heights Station

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Cole Instrument Co.
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Standard Arcturus Corp.
99 Sussex Ave.
Tung-Sol Lamp Works, Inc.
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Southern Electronic Co.
512 St. Charles St.

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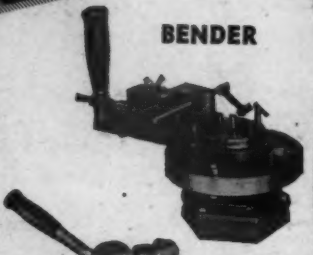

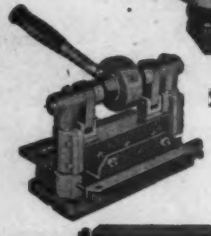

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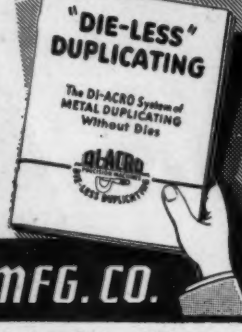

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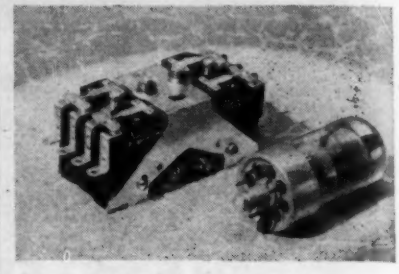
new tube desirable for compact equipment. Details are given in Bulletin E 129.

Latching Relay (50)

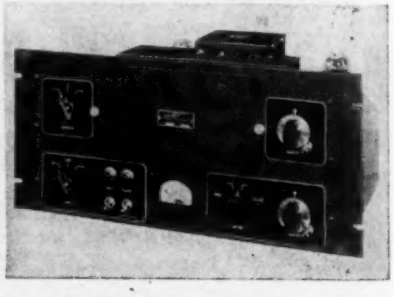
SIGMA INSTRUMENTS, 70 Ceylon St., Boston, Mass. The type 6FZ2A2B relay is a multicircuit device that remains latched after initial operation until it is desired to release the



mechanism. The coils are wound for d-c operation and can be supplied in various resistance values up to 10,000 ohms. In one type of service, a relay can be adjusted to trip from the 50-volt charge on a 0.25-microfarad capacitor.

Frequency Shift Telegraph (51)

ERCO RADIO LABORATORIES, Inc., Garden City, N. Y. Frequency shift telegraph equipment is built in three units. The type 177-T exciter



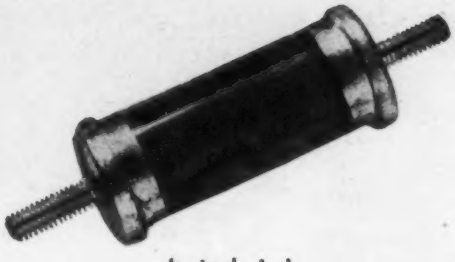
for keying transmitters illustrated is crystal controlled with provision for 3 frequencies. Mark and space frequencies are both adjustable. The type 216-S keyer-converter and type 87-R receiver complete the equipment necessary for a circuit.

Outdoor Microphone (52)

ST. LOUIS MICROPHONE CO., Inc., 2726 Brentwood Blvd., St. Louis 17, Missouri. The Outdoor Rugged Dynamic Microphone has been built specifically to weather outdoor serv-

Now Available---

A Line of PLASTICON* GLASSMIKE RF Transmitting Capacitors



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Superior to mica capacitors because:

- Greater safety factor (3500 VDC Operating; 7500 VDC Test)
- Lower RF losses (See current rating below)
- More conveniently mounted
- Less chassis space
- Smaller overall volume
- Impervious to moisture (The GLASSMIKE construction is 100% sealed)
- Silicone-fluid filled

The above advantages are possible by the use of the Type L film dielectric which has lower losses than mica.

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TYPE LSG — PLASTICON* GLASSMIKES

3500 VDC Operating — 7500 VDC Test

Cat. No.	Cap. Mfd.	Dimensions OD Length	Current Rating in RF Amperes 100 Kc	300 Kc	1 Mc	3 Mc	List Price
LSG500.....	.00005	19/32"x1-3/16"	.02	.05	.16	.47	$1.50
LSG101.....	.0001	19/32"x1-3/16"	.03	.09	.31	.94	1.50
LSG251.....	.00025	19/32"x1-3/16"	.05	.25	.5	2.2	1.50
LSG501.....	.0005	19/32"x1-3/16"	.15	.5	1.6	3.0	1.50
LSG102.....	.001	19/32"x1-9/16"	.31	.94	2.5	4.5	1.70
LSG202.....	.002	3/4" x1-9/16"	.62	1.9	4.5	7.0	2.45
LSG502.....	.005	3/4" x1-3/4"	1.6	3.1	6.0	7.0	3.50
LSG602.....	.006	29/32"x1-9/16	1.9	3.5	6.2	7.0	3.75
LSG103.....	.01	29/32"x1-3/4"	3.1	5.0	7.0	7.0	4.25

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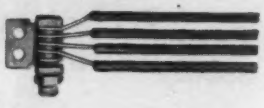
COPPER OXIDE RECTIFIERS

INSTRUMENT RECTIFIERS FOR BETTER A.C. SCALE



Vacuum-processed, gold-coated, full-wave bridge, Bradley instrument rectifiers allow more accurate determination of very low A.C. current. Especially designed for use where stability and permanence of calibration are important, "Coprox" rectifiers meet the most exacting requirements. Yet they cost no more than ordinary rectifiers — in most cases, less.

Temperature error is exceptionally low with Bradley rectifiers. Aging is practically nil. Shown above, CX2E4F rated up to 4.5 volts A.C., 3 volts and 5 milliamperes D.C.

Illustrated literature, available on request, shows more models of copper oxide rectifiers, plus a line of selenium rectifiers and photocells. Write for "The Bradley Line."

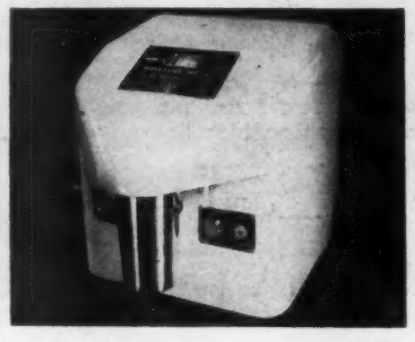
BRADLEY LABORATORIES, INC.

82 Meadow St. New Haven 10, Conn.

ive and mechanical jarring. Range of the microphone is 40 to 9,000 cycles and a variable-impedance output provides low to high adjustments, including 200 and 500 ohms.

Moisture Meter (53)

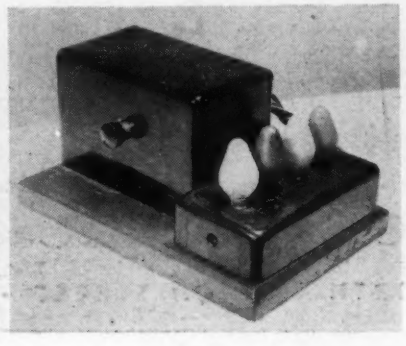
C. J. PATTERSON Co., 3947 Broadway, Kansas City, Missouri. The direct-reading electronic moisture



meter is applicable to any solid, liquid, granular, or powdered material. Automatic in action, it reports the percentage moisture in fifteen seconds.

Interval Timer (54)

MARITIME QUALITY HARDWARE CO., INC., Belfast, Maine. The motor-driven timer illustrated can be used for any program or interval with a range from a five-minute cycle to twenty-four hours. It has four separate circuits that can be con-



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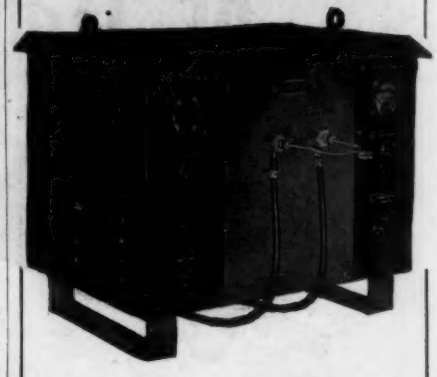


For Only $650.

Never before a value like this new 2-KW bench model "Bombarder" or high frequency induction heater . . . for saving time and money in surface hardening, brazing, soldering, annealing and many other heat treating operations.

Simple . . . Easy to Operate . . . Economical Standardization of Unit Makes This New Low Price Possible

This compact induction heater saves space, yet performs with high efficiency. Operates from 110-volt line. Complete with foot switch and one heating coil made to customer's requirements. Send samples of work wanted. We will advise time cycle required for your particular job. Cost, complete, only $650. Immediate delivery from stock.

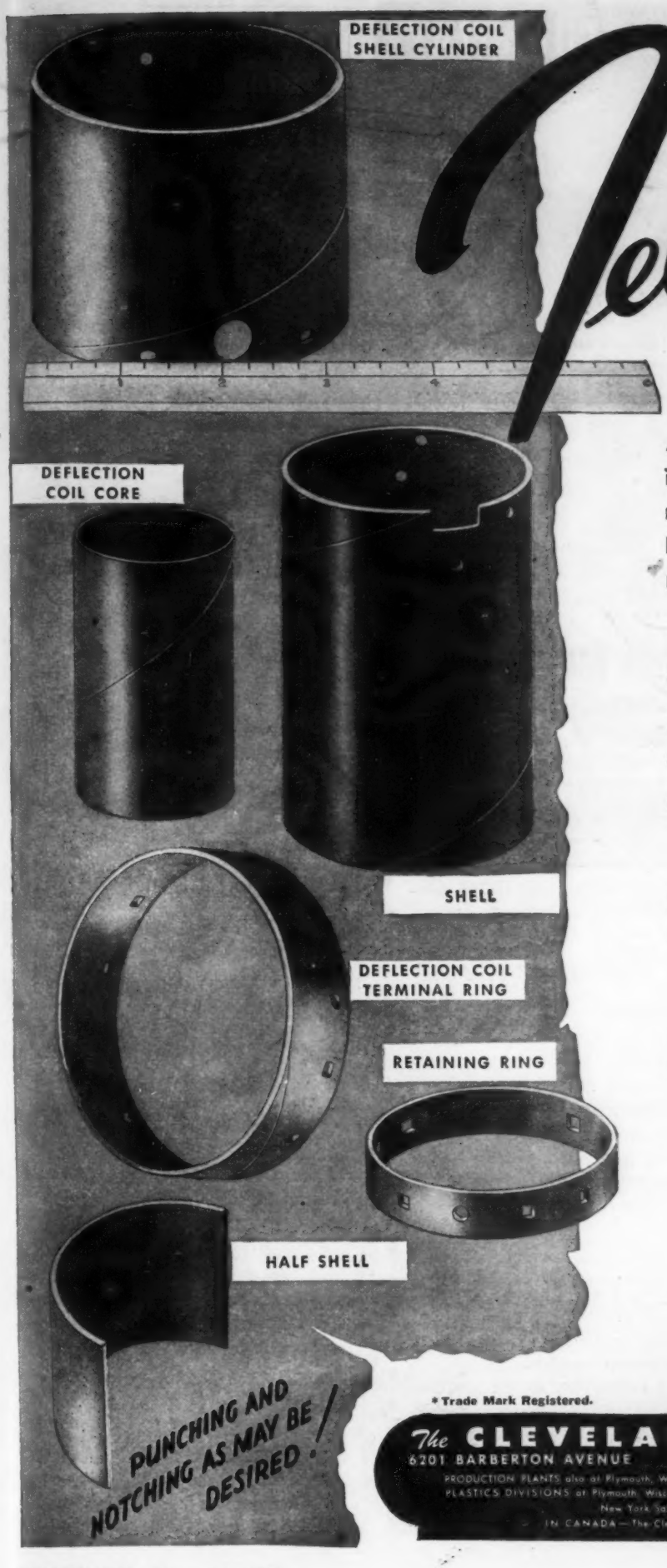
Scientific Electric Electronic Heaters are made in the following range of Power: 1-3-5-7½-10-12½-15-18-25-40-60-80-100-250-KW. — and range of frequency up to 300 Megs. depending on power required.

Scientific Electric

Division of

"S" CORRUGATED QUENCHED GAP CO.

107 Monroe St., Garfield, N. J.



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Spirally wound kraft and fish paper Coil Forms and Condenser Tubes.

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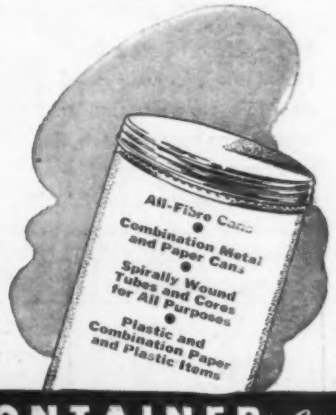
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Combination Metal and Paper Cans
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Plastic and Combination Paper and Plastic Items

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PRODUCTION PLANTS also at Plymouth, Wisc. Ogdensburg, N. Y. Chicago, Ill. Detroit, Mich. Jamesburg, N. J.
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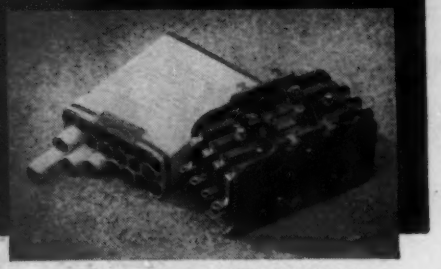
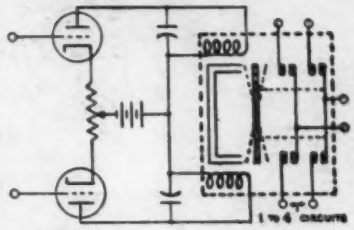
SIGMA 3 POSITION OUTPUT RELAY

FOR CONTROL SYSTEMS AND SERVOMECHANISMS



This new relay by Sigma, already proven in months of development by prominent manufacturers of control and guiding equipment, overcomes many shortcomings previously inherent in relay-output control systems.

SIGMA TYPE 6 FX8A RELAY
Weight 8 Ounces
(Available Hermetically Sealed.)



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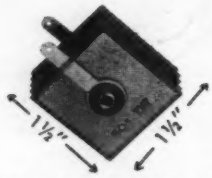
2. True magnetic cancellation at balance — not affected by variations in standby current.
3. Exceptional sensitivity — approximately .005 watt differential power required per contact circuit.
4. Polarized magnetic circuit increases speed of action and facilitates discrimination as to polarity of error.
5. Fully balanced moving system withstands vibration to exceptionally high degree.
6. Flexible contact structure (maximum shown above) with substantial load handling ability (nominal rating 5 amp 110 Volts A.C. per contact).

SIGMA

Sigma Instruments, INC.
Sensitive **RELAYS**
62 Ceylon St., Boston 21, Mass.

Seletron **5Q1** **NEW!** 5Q1 · MINIATURE Selenium Rectifier..

Here's the 5Q1, the newest SELETRON miniature selenium rectifier — built on aluminum.



5Q1 — 250 mil 5-plate stack. 1½"x1½"x⅞". Maximum AC input 130 volts r.m.s.; rectifier voltage drop approximately 5 volts r.m.s. Recommended input filter capacitor 80 m.f.d.

Check These Features

- STARTS INSTANTLY
- RUNS COOL
- WILL NOT BREAK
- TAKES MOMENTARY HEAVY OVERLOADS
- BOOSTS PERFORMANCE
- EASILY INSTALLED
- EACH UNIT FACTORY TESTED
- HIGH EFFICIENCY ASSURES COMPACTNESS
- NORMALLY LASTS LIFETIME OF SET

5Q1 now joins the family of SELETRON miniature selenium rectifiers which include the 5P1, 150 mil 5-plate stack, 1 3/16"x1 3/16"x⅞" and the 5M1, 100 mil 5-plate stack, 1"x1"x⅞".

For maximum efficiency, reliability and service, specify SELETRON miniature selenium rectifiers.

WRITE TODAY FOR FULL INFORMATION. ADDRESS DEPT. S-43.

Our new bulletin "SELETRON Selenium Rectifiers" will soon be off the press. Write for your copy now

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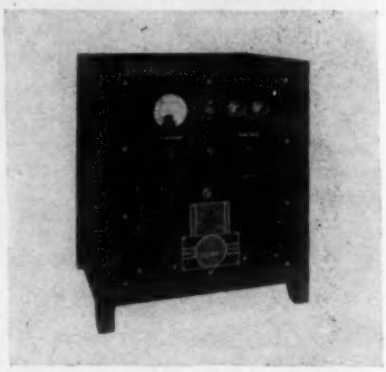
SELETRON DIVISION
RADIO RECEPTOR CO., Inc.
Since 1922 in Radio and Electronics
251 WEST 19th STREET NEW YORK 11, N. Y.

trolled on and off with any commercial accuracy desired. When used on 115-volt circuits each set of contacts will carry a maximum of 30 amperes.

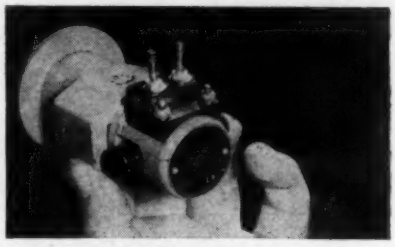
Automatic Voltage Regulator (55)

RICHARDSON-ALLEN CORP., 15 W. 20th St., New York 11, N. Y. The model 442A automatic voltage regulator maintains a constant d-c output voltage regardless of ampere load current. An easily handled unit, it is designed for 115-volt, single phase, 50 or 60-cycle a-c input, and 1 to 9 volt d-c output.



Permanent-Magnet Motor (56)

AMPEX ELECTRIC CORP., 1155 Howard Ave., San Carlos, Calif. A new line of permanent-magnet motors derived from wartime experience is now available. The 25-watt unit illustrated weighs 15 ounces. A complete range of speeds, voltages, shaft sizes and mounting-flange designs can be supplied.



Miniature Tube Shield (57)

STAVER MFG. CO., 33-21 85th St., Jackson Heights, N. Y. Mini-Shields come in three lengths and of a diameter to fit the smaller

ANOTHER KAY DEVELOPMENT

THE MEGA-MATCH....

MEASURES REFLECTED ENERGY...

GIVES VISUAL DISPLAY...

A BASIC LABORATORY INSTRUMENT

FEATURES

- 10 to 250 MC and up. Complete television and FM coverage.
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- Saves engineering time—Visual display presents instantly data which would take hours to tabulate.
- Can be adapted for balanced lines.

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This unique instrument presents a visual display of REFLECTED energy over any band up to 30 MC. By the use of the MEGA-MATCH it is possible to instantly observe and measure mismatches. Thus this instrument will check transmission lines, antennas, input and output impedances of amplifiers, converters, transformers, etc.

Price: $695 F.O.B. Newark, N. J.

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WIDE RANGE SWEEPING OSCILLATOR THE MEGA-SWEEP.....

Reference August 1947
"Electronics" page 112.

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Continuous frequency coverage up through the color television bands

Shows at a glance the response of any network or amplifier. Eliminates the tedious point to point analysis. Saves engineering time and stimulates research. Valuable for television production alignment.

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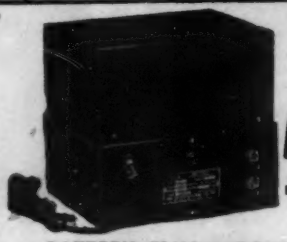

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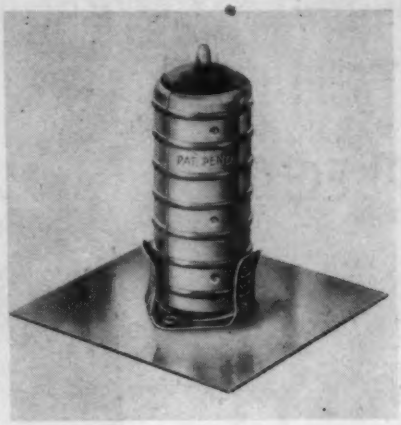
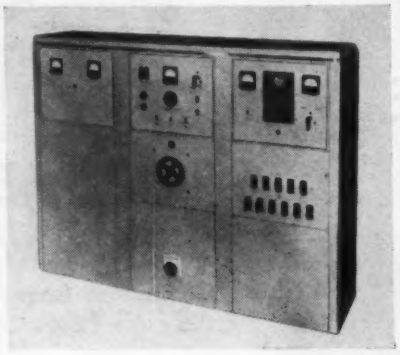


miniature tubes. An expansion joint allows the shield to be pushed over tubes of larger diameters. Free air circulation is provided by dimples that space shield from tube.

Spectograph Source (58)

APPLIED RESEARCH LABORATORIES, 7707 Michigan Ave., Detroit 10, Mich. Electrical discharge equipment for spectrochemical analysis is available for high quantitative



precision. A high voltage source uses 240 spark trains per second with short duration and the low voltage source uses 60 arc discharges a second of long duration for high sensitivity.

Portable Volt-Ohmmeter (59)

HICKOK ELECTRICAL INSTRUMENT CO., 10527 Dupont Ave., Cleveland 8, Ohio, introduces the Model 214 portable, battery-operated electronic volt-ohmmeter to eliminate the need for plugging into an a-c line. It is built for self-checking of battery condition at any time and can be used for 1,350 complete cycles of 4-minute operations. Ranges

"SPINTITES"

One of the surest ways of sustaining assembly line speed is to standardize on Spintite wrenches. Made to meet the particular problems of radio and electrical assembly and repair, they're designed for precision performance, volume production, durability and ease of operation with a minimum of skill.

Built like a screwdriver, the Spintite shaft readily reaches difficult assembly spots, and it is partly hollowed to permit tightening of nuts through which the bolt protrudes.

Available with either fixed or chuck-type handle, Spintites can be had to fit square, knurled or hex nuts in sizes from 3/16" to 5/8". For the radio and electrical industry's three requisites in tools, speed, accessibility and quantity — specify Spintites.

T-73 SET, has 7 sizes of hex heads. Shock-proof handles, and cold forged sockets assure safety and strength.

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New! PALNUT SHIELD CAN FASTENER*

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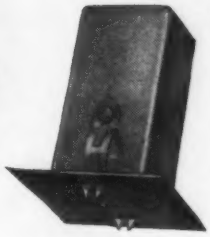

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● A quick snap of the Palnut Shield Can Fastener into the chassis provides a secure job—faster, cheaper than other fastening methods. Good ground contact is maintained. May be used on any chassis thickness.

SAMPLES *and data on Palnut Shield Can Fasteners sent upon request on your company letterhead.*

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The Series 1110 Indicator Light comes in styles to accommodate either the long bulb or the small round bulb. Unscrew the jewel on the front and there's the bulb—no need to mess around behind the panel. All brass construction except nut and lock washer. Parts heavily plated. Not affected by heat. Furnished with famous Gothard rigid, non-short terminals; fibre space washer; lock washer and nut. Requires 11/16" mounting hole. Jewel is ½" dia., either faceted or smooth glass; colored red, green, amber, blue, opal, clear.

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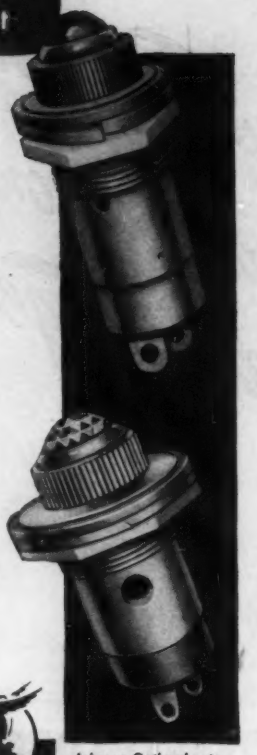
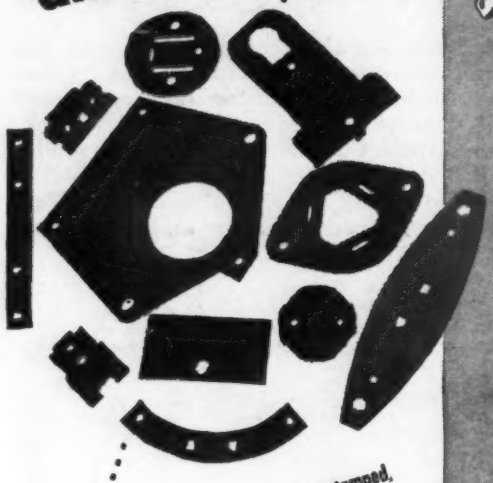


Johnson-Gothard catalog fully describes this Indicator Light and many others—write for it.

Baer can supply your vulcanized and phenol fibre fabrications



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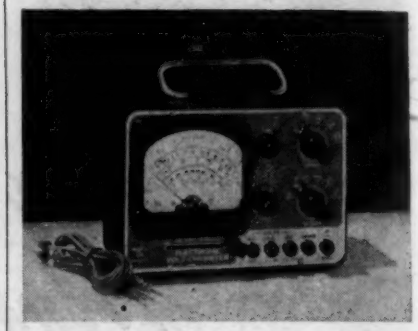

include a-c or d-c voltage up to 1,200 v.; and resistance up to 1,000 megohms.

Literature

(60)
Magnetic Relays. R-B-M Division,, Essex Wire Corp., Logansport, Indiana, has issued Bulletin 570 covering general purpose a-c and d-c relays. Complete descriptions, with photographs and diagrammatic dimensions, are given.

(61)
Vibrator Supply. Raytron Inc., 407 North Jackson St., Jackson, Mich. Portable radio or other electronic equipment can be furnished high voltage from a unit that is complete with input and output leads and screw studs for mounting. The new type vibrator is practically noiseless. Suggested circuit arrangements are given in the 3-page brochure.

(62)
Engineering Literature Service. T. A. Farrell Jr., 141 Welch's Point Road, Milford, Conn. Technical manuals, engineering surveys, service literature, and similar publication work are among the services offered to electronic engineers and industries.

(63)
House Organ. Consolidated Engineering Corp., 620 N. Lake Ave., Pasadena 4, Calif., issues a small publication known as CEC Recordings. In it are reported facts of interest about the line of mass spectrometers, recording oscillographs, and similar research

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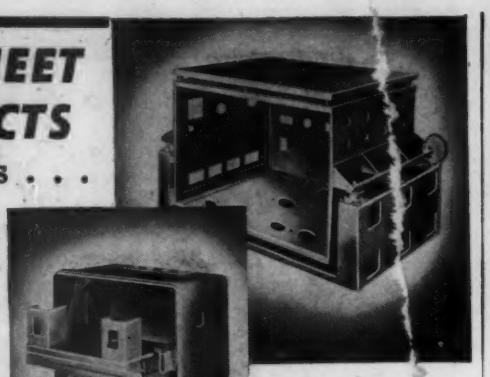

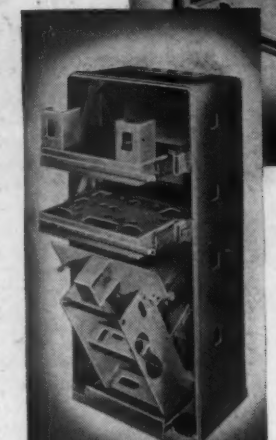

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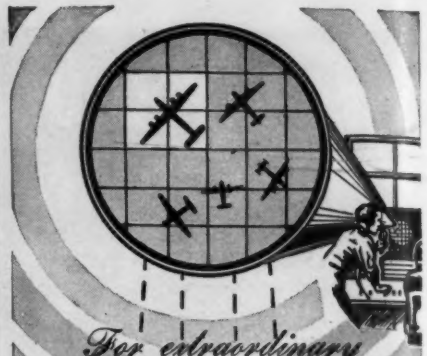

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equipment manufactured by the company. The quarterly publication will be of interest to many engineers.

(64)
Radar Components. Carl W. Schutter, Rockville Centre, N. Y. Factory representative H. L. Hoffman & Co., 673 Broadway, New York 12, N. Y. Six illustrated pages show the line of chokes, choke flanges, coaxial connectors, and component parts available for radar use. Special components can also be furnished.

(65)
Industrial Tubes. National Electronics, Inc., Batavia Ave., Geneva, Ill. A new technical data book in loose-leaf form gives rating and application information on eleven rectifier and thyratron types suitable for welding or motor control and industrial rectifier applications.

(66)
Insulation Tester. James G. Biddle Co., 1316 Arch St., Philadelphia 7, Pa. A new plug-in Megger insulation tester requires no hand crank to cover 2,000 megohms at 1,000 volts. Bulletin 21-46-14 tells about it.

(67)
Universal Parts. General Electric Co., Syracuse, N. Y. A 24-page brochure on universal electronic parts, ESD-93, lists price, specifications, and other data on 16 parts in the line that includes resistors, controls, antennas, and the variable-reluctance pickup as well as loudspeakers.

(68)
Selenium Rectifiers. Selectron Div., Radio Receptor Co., Inc., 251 West 19th St., New York 11, N. Y., announces a new 8-page bulletin on the subject of selenium rectifiers for direct-current requirements, including electrical characteristics, dimensions, and weights.

(69)
Big Tubes. Radio Corp. of America, Harrison, N. J. The Quick Se-

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Manufacturers of
VOLTAGE REGULATORS, NOBATRONS & ELECTRONIC APPARATUS

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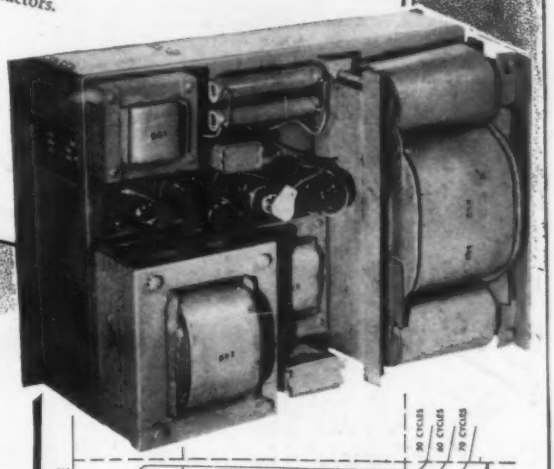
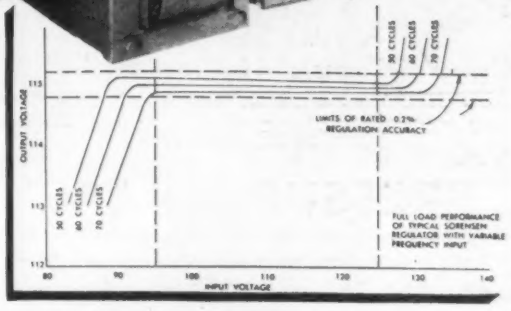
RUNAWAY VOLTAGES STOPPED AT 1/10 OF 1%

Rated performance of Model 1750-S *guarantees* delivery of output line voltages at a regulation accuracy of 0.2% under varying load. However, in *actual tests* of this unit voltage stabilization was held to within 0.1% under full operating conditions. This conservative safety rating of 0.2% is typical of all Sorensen performance factors.

Input voltage range	95-125
Adjustable output between	110-120
Load range	200-2000 VA
Regulation accuracy	0.2%
Harmonic distortion	2% max.
Recovery time	6 cycles
Input frequency range	55-65 cycles



IT IS "A NATURAL" FOR CONTROLLING VOLTAGES IN LABORATORIES, ASSEMBLY LINE TESTING AND AS A COMPONENT OF YOUR ELECTRICAL UNIT.



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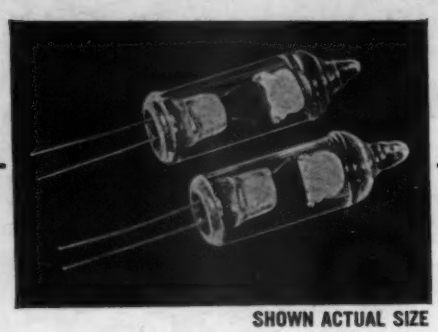

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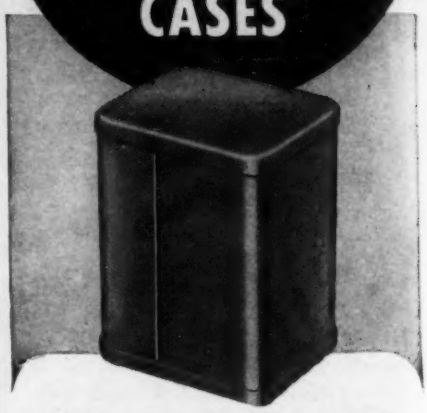
Compact — hermetically sealed — high performance — for multiple applications in the range of 2,500 to 10,000 kilocycles with a tolerance of ±250 cycles at room temperature . . . MINIGLASS* crystals are designed for your low cost quantity production items. Immediate delivery.

Manufacturers of Radio Quartz Crystals.

MELVIN L. SMITH LABORATORIES
KANE, PENNSYLVANIA

**Trade Mark †Patent applied for*

OLYMPIC STANDARD TRANSFORMER CASES



Now available in quantity, OLYMPIC standardized transformer cases are specifically designed to meet all normal requirements where standard cases are used. Construction is rigid, with rounded corners, and tight-fitting covers top and bottom. OLYMPIC standard transformer cases can be furnished with pierced covers, studs, brackets or channels. Inquiries are invited—write for illustrated bulletin—no obligation.

CORE	CASE	A	B	C
E1-21	1	1⅛"	1⅛"	2 1/16"
E1-625	2	1⅞"	1⅞"	2 1/16"
E1-75	3	2¼"	2⅛"	2 1/16"
E1-11	4	2⅝"	2½"	3 1/32"
E1-12	5	3"	2⅞"	3 3/32"
E1-3A	6	3¼"	3"	3 1/32"
E1-112	7	3⅜"	3¼"	4 1/32"
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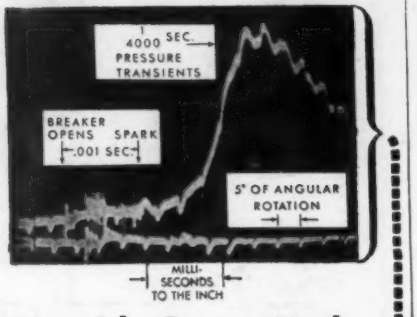

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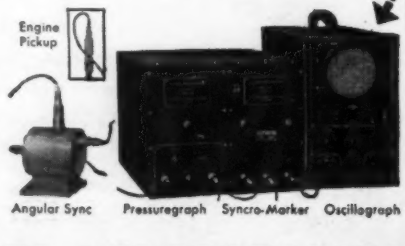

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lection Guide offers technical data on more than 200 transmitting and industrial tubes. Dimensions, ratings, and other essential information are presented in abridged form.

(70)
Mass Spectrometer. Consolidated Engineering Corp., 620 N. Lake Ave., Pasadena 4, Cal. Copies of a 20-page, illustrated folder, describing the model 21-102 mass spectrometer, are now available. The mechanism provides a direct approach to the determination of mixture composition by the sorting and counting of molecules.

(71)
Tube Data. Hytron Radio and Electronics Corp., 76 Lafayette St., Salem, Mass. Transmitting and Special-purpose tubes have been cataloged in a four-page brochure that gives complete operational data and socket connections.

(72)
Radar. Raytheon Mfg. Co., Waltham, Mass. A slick-paper 12-page booklet in two colors covers the highlights of a navigational radar for merchant shipping known as Mariners Pathfinder.

(73)
Amateur Antennas. Workshop Associates Inc., 66 Needham St., Newton Highlands, Mass. Broadband antennas and accessories are summed up in a recently printed 4-page pamphlet that gives descriptions, characteristics, and prices of the line most suitable for amateur use.

(74)
Control Relay. Niagara Electron Laboratories, Andover, N. Y. Bulletin T-1/9-46 describes the Thermocap relay that is extremely sensitive to minute changes in electrical capacitance. Many applications are possible in industrial control.

(75)
Vibration Mountings. Lord Manufacturing Co., Erie, Pa. The chief features and applications of Mul-

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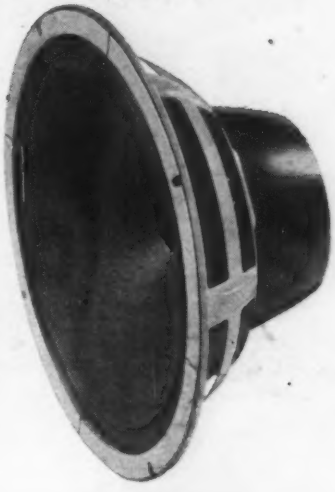
make the difference



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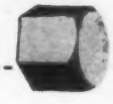

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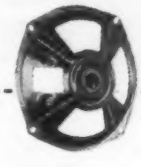

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tiplane mountings, shear-type bonded-rubber vibration controls, are given in a 12-page booklet. Also included are tables showing sizes and specifications, and instructions on selection and installation.

(76)

Permanent Magnets. General Electric Co., Pittsfield, Mass., has issued a 36-page illustrated booklet on permanent magnets. General information as well as technical data on characteristics, design, properties and applications are given. Also included is a list of definitions of magnet terms.

(77)

Aluminum Wire. United States Rubber Co., Wire and Cable Dept., Rockefeller Center, New York, N. Y. A 25-page handbook points out that aluminum building wire is on a par with copper in regard to quality, has lower production costs, and is in more plentiful supply. Tables show relative weights of various size aluminum and copper conductors.

Wire and Cable Manual. Rome Cable Corp., Rome, N. Y. A pocket size manual including wire and cable technical tables, audio and radio frequency wire and cable data, and mathematical and conversion tables is available directly from the company at a cost of $2.50.

Magnetic Tape Recorders. A bibliography of OTS reports on magnetic tape recorders may be obtained free of charge by writing direct to the Reference Service Section, Office of Technical Services, Department of Commerce, Washington 25, D. C.

Master Index. John F. Rider, Publisher, Inc., 404 Fourth Ave., New York 16, N. Y. has just issued a master index to cover information on all radio sets from 1919 through 1946. Printed in the same size as the well-known manuals, the paper-covered volume sells for $1.50.

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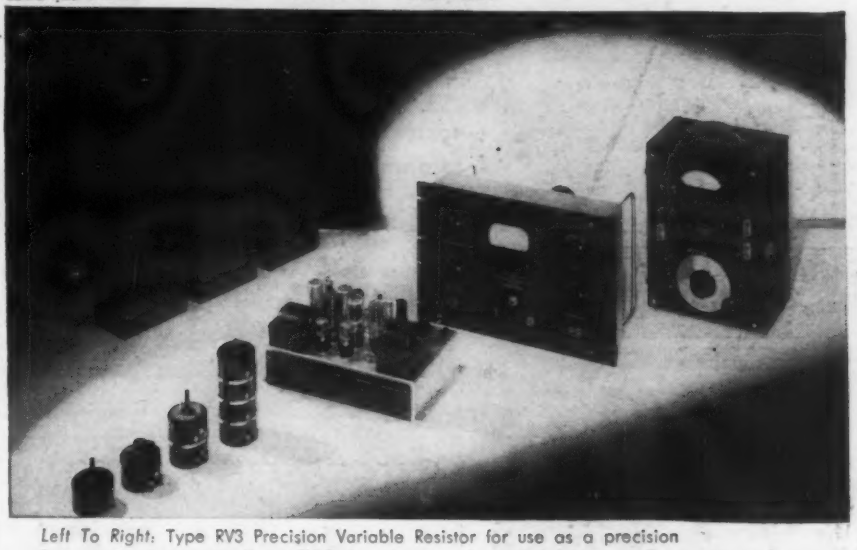

Left To Right: Type RV3 Precision Variable Resistor for use as a precision instrument component; Type RVL3 for the experimental laboratory; Two examples of ganging simplicity (Type RV3 Resistors); Type 210-A Laboratory Amplifier with Dynamic Noise Suppressor for radio-phonograph use; Type 910-A Dynamic Noise Suppressor, Broadcast station model; Z-Angle Meter for checking "match up" in any sound circuit.

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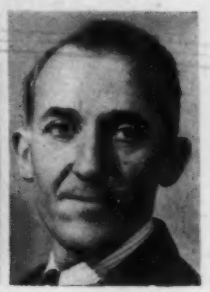

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G. P. Adair

ducted for small stations by G. Porter Houston, chief engineer of WCBM, Baltimore, Md., and for medium-size stations by Alfred E. Towne, director of engineering for KSFO in San Francisco, Calif.

Directional Antennas, Their Care and Maintenance, by Dixie B. McKey, consulting radio engineer, Washington, D. C.

Technical Regulation of Radio, by George P. Adair, radio engineering consultant, Washington, D. C.

FCC-Industry engineering roundtable; Commission representatives, headed by chief engineer George E. Sterling, will be: John A. Willoughby, assistant chief engineer; James E. Barr, chief of Standard Broadcast Division; Cyril M. Braum, chief of F-M Broadcast Division; Curtis B. Plummer, chief of Television Broadcast Division.

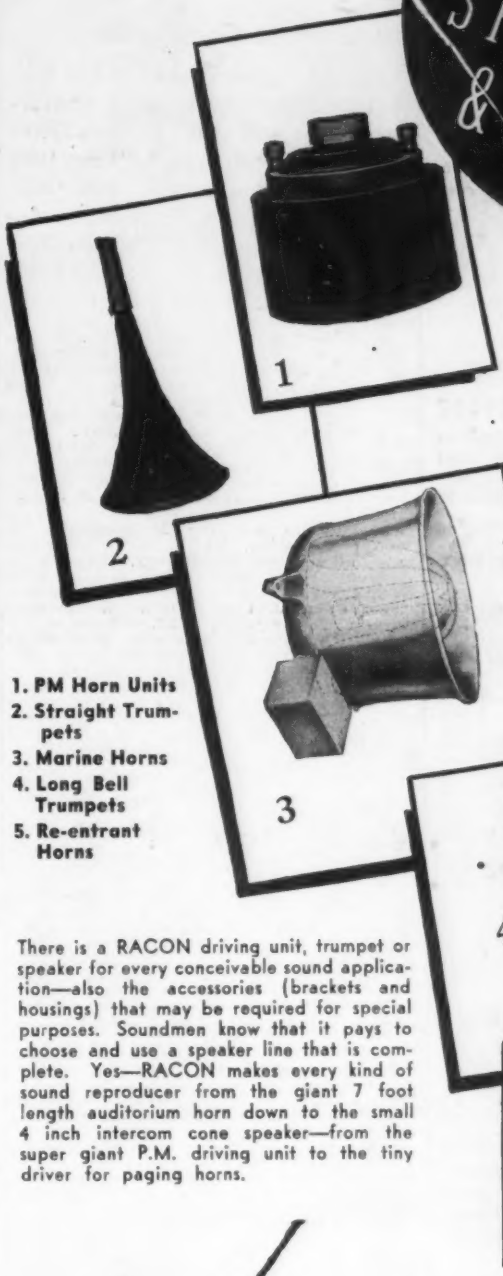
Also of engineeringwise interest at the convention will be the exhibit and demonstration of latest television, facsimile, f-m, and a-m broadcasting equipment.

FCC Licenses for Radar

PURCHASERS of surplus radar and other electronic devices capable of radio emissions must obtain appropriate station licenses before operating such equipment, according to an FCC warning. Section 301 of the Communications Act requires licensing of radar equipment, and Section 318 stipulates that radar stations may be operated only by persons licensed or authorized by the Commission.

Retailers of surplus equipment are urged to cooperate by attaching to apparatus a tag calling attention to the penalties involved in unauthorized operation. These tags may be obtained without charge from the Secretary, Federal Com-

use RACON SPEAKERS & HORNS



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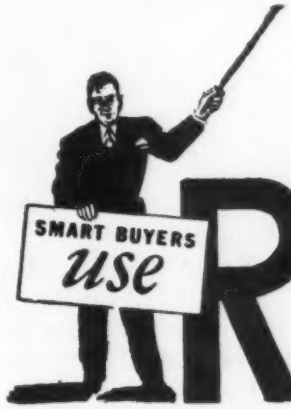

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6. The Gathering and Presentation of Radar Data
7. The Employment of Radar Data
8. Radar Beacons
9. Antennas, Scanners, and Stabilization
10. The Magnetron and The Pulser
11. R-F Components
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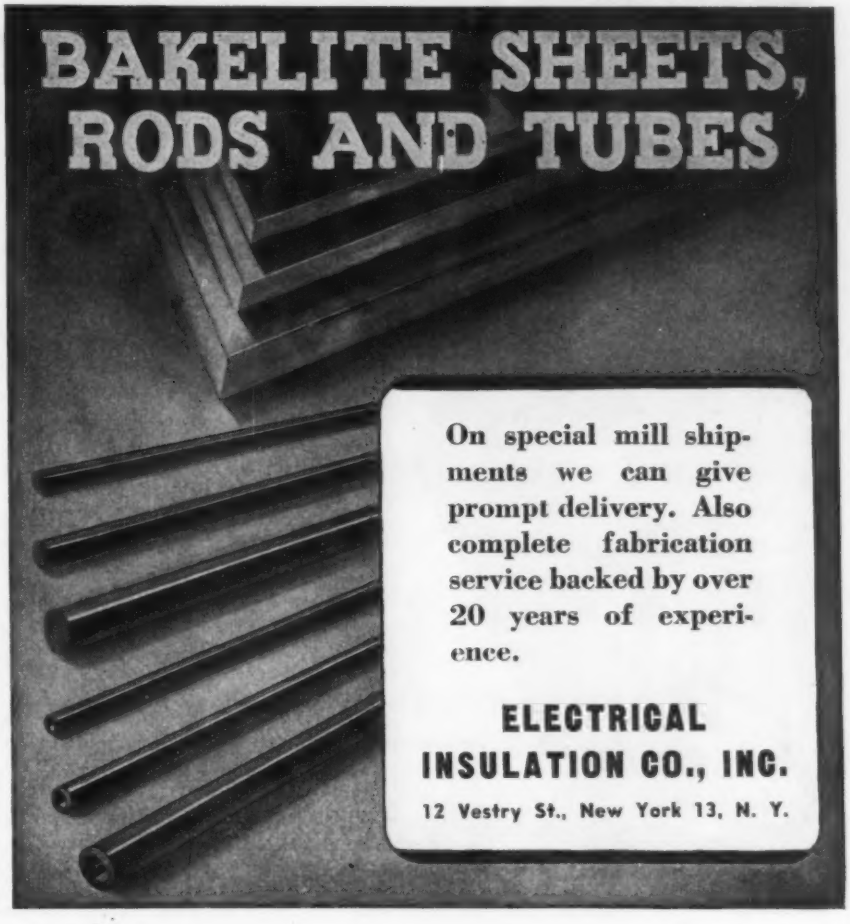

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Instrument Conference

ADVANCE registrations for the Second National Instrument Conference, to be held at the Stevens Hotel in Chicago Sept. 8-12, indicate that attendance will exceed 7,500. General chairman is A. F. Sperry, president of Panellit, Inc., Chicago. The technical program for the four days is as follows:

Sept. 9—9:00 a.m.: Quality Control—A Management Tool for Process Industries, by W. E. Jones of Management Controls, Desplaines, Ill.

Precision of Measurements, by F. Trowbridge of Sentinel Radio Co., Evanston, Ill.

9:00 a.m. (concurrent session): The Fundamentals of Radioactivity, by G. J. Overbeck of Northwestern University.

The Application of Radioactive Indicators to Industrial Problems, by R. D. Evans of MIT.

Instruments for the Measurement of Tracer Radiation, by F. R. Shonka of Argonne National Laboratory, Chicago.

2:00 p.m.: Applications of Viscosity Measurements in Petroleum Refining Operations, by D. J. Pompeo of Shell Development Co., Emeryville, Calif.

Polarographic Analysis — Principles, Techniques, and Applications, by L. D. Wilson of Corn Products Refining Co.

Adaptation of the General Electric Spectrophotometer for Special Measurements, by R. H. Kienle and E. I. Stearns of American Cyanamid Co.

Some Problems of Application of Conductivity Equipment, by A. L. Chaplin of Johns Hopkins University.

8:00 p.m.: Automatic control and servo mechanism symposium.

Sept. 10—9:00 a.m.: Demonstration of Automatic Control Principles, by G. F. Akins of Eastman Kodak Co., assisted by John H. Kowalski.

2:00 p.m.: Instrument Installation and Maintenance Methods and Practices in a Modern Oil Refinery, by G. A. Larsen of Texas Co.

Instrumentation in Turbine Testing, by A. B. Sisson of Public Service Co. of Northern Illinois.

Organization and Operation of a Calibration Laboratory, by H. N. Hayward of University of Illinois.

Sept. 11—9:00 a.m.: Instrumentation of an Acetic Acid Concentration Process, by J. E. Voytilla of du Pont.

Sea Water Bromine Process Chemical Control System, by P. Hart of Dow Chemical Co.

2:00 p.m.: Dew Point Recorder, by B. Suomi of University of Chicago.

Instrumentation for Aeronautical Research, by Edward Blom of Cornell Research Foundation, Inc.

Design and Application of A New Metals Comparator, by D. E. Bovey of G-E.

Sept. 12—9:00 a.m.: Basic Design and Field Performance of the Control and Measurement Instrumentation of an Alkylation Unit, by H. F. Moore and G. W. Gross of Standard Oil.

Application of Conventional Industrial Instruments to Power Plant Control, by W. H. Fortney of Humble Oil Co.

Western IRE Conference

OPENING at the Palace Hotel, San Francisco on Sept. 24, the IRE conference held in conjunction with the West Coast Electronic Manufacturers' Association will extend through Sept. 26 and include six technical

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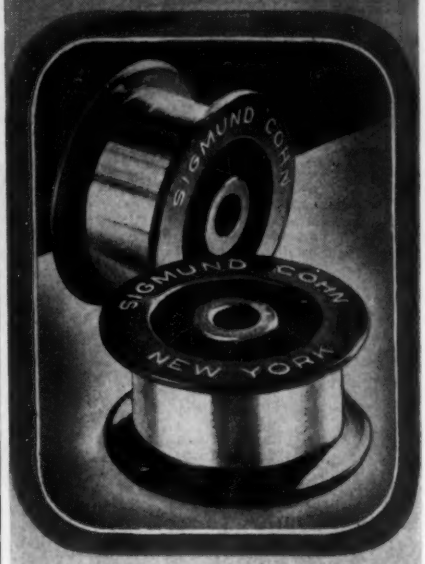

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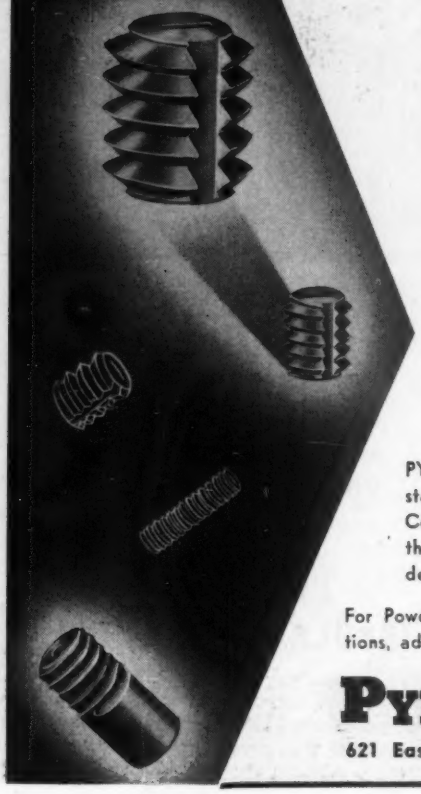

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sessions. Beginning Sept. 26, the WCEMA will offer closed sessions of exhibits at the Whitcomb Hotel for two days, then open the hall to the general public on Sunday Sept. 28. Papers scheduled for the IRE session include the following:

Technical Problems of Military Radio Communications of the Future, by John Hessel of Signal Corps Engineering Labs.

Some Applications of Electronics to Underwater Ordnance, by Ralph D. Bennett of Naval Ordnance Lab.

Some Experimental Determinations of Mutual Impedance of Antennas, by F. R. Abbott of U. S. Navy Electronics Lab.

A New Pulse-Time Telemetering System, by J. N. Davis of Sylvania Electric.

Microwaves in Ordnance Work, by F. G. Suffield of Allison Associates.

Telemetering Guided Missile Performance, by James C. Coe of U. S. Naval Air Missile Test Center.

A 50,000-Watt FM Transmitter for 100.5 Mc, by Leigh Norton of Eitel-McCullough.

Susceptibility of F-M Receivers to Interfering Signals, by D. E. Foster of Hazeltine Research Inc. of California.

Frequency Modulation Detectors, by S. W. Seeley of RCA Industry Service Lab.

Limiters and Discriminators in F-M Receivers, by W. G. Tuller of MIT.

A VHF Bridge for Impedance Measurements, by Robert Soderman of General Radio.

A Pulse-Counter Type F-M Station Monitor, by David Packard and Norman Schrock of Hewlett-Packard.

Supersonic Flaw Detection, by Donald Erdman of Triplett and Barton.

Electronic Gage Methods and Apparatus, by R. L. Sink of Consolidated Engineering.

Equivalent Networks for Wave-Guide Problems by J. R. Whinnery of U. of Calif.

The New York-Boston Radio Relay, by J. W. McRae of Bell Labs.

High-Quality Loudspeakers, by J. K. Hilliard of Altec Lansing.

A New Solution of the Antenna Problem, by C. Lanczos of Boeing Aircraft.

The Proton Linear Accelerator, by L. W. Alvarez of U. of Calif.

The Electron Linear Accelerator, by W. W. Hansen of Stanford University.

Oscillation and Gain Properties in New Types of Traveling-Wave Tubes, by L. M. Field of Stanford.

Resnatron Design, by W. W. Salisbury of Collins Radio.

Radio-Wave Propagation in the F-M Broadcast Band, by K. A. Norton of National Bureau of Standards.

A Five-Kw Television Transmitter, by J. E. Keister, J. W. Downie, H. B. Sancher, and L. M. Ewing of G-E.

A Modern Television Transmitter, by C. D. Kentner of RCA Labs.

Receiving Antennas for F-M and Television.

Airline Inaugurates Radiograms

PASSENGERS on Pan American World Airways new round-the-world route will be able to send radiograms to any part of the United States. Use of Radiomarine's powerful radiotelegraph station at Chatham, Mass. as the key station for this service has been approved by the FCC. Other RCA stations in Manila, Honolulu, and Bolinas, Calif. will assist in handling correspondence if needed.

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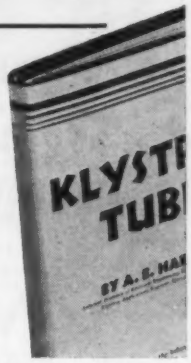

This authoritative book provides electronic engineers, radio technicians, and others, with a sound working knowledge of both the operating principles and practical applications of klystron tubes. It supplies the bulk of the answers to your questions concerning methods of construction — performance characteristics—procedures for testing, etc. —and shows how these tubes may be used as oscillators, amplifiers, frequency multipliers, and detectors or mixers.

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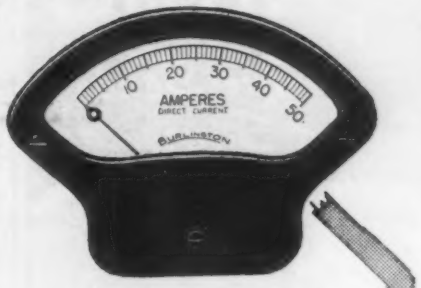

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RMA Section Chairmen

SECTION chairmen for the Parts Division of RMA during 1947–48 fiscal year have been selected by division chairman J. J. Kahn, president of Standard Transformer Corp., Chicago, as follows:

Coil Section: Edwin I. Guthman, Edwin I. Guthman & Company, Inc., Chicago, Ill.
Fixed Capacitor Section: W. Myron Owen, Aerovox Corporation, New Bedford, Mass.
Fixed Resistor Section: J. Hall Stackpole, Stackpole Carbon Company, St. Marys, Pa.
Instrument & Test Equipment Section: R. L. Triplett, The Triplett Electrical Instrument Co., Bluffton, Ohio.
Insulations Section: John W. Apgar, Irvington Varnish & Insulator Co., Irvington, N. J.
Metal Stampings and Metal Specialties Section: S. L. Gabel, Superior Tube Company, Norristown, Pa.
Phonograph Cartridges and Pickups Section: George B. Fraser, The Astatic Corporation, Conneaut, Ohio.
Plastics and Molded Parts Section: John J. Bachner, Chicago Molded Products Corporation, Chicago, Ill.
Record Changer and Phono-Motor Section: Allan W. Fritzsche, The General Industries Company, Elyria, Ohio
Socket Section: Frank Holmstrom, Hugh H. Eby, Inc., Philadelphia, Pa.
Speaker Section: Laurence A. King, The Rola Company, Inc., Cleveland, O.
Speaker Parts Section: A. D. Plamondon, Jr., The Indiana Steel Products Company, Chicago, Ill.
Special Products Section: William R. MacLeod, King Laboratories, Inc., Syracuse, N. Y.
Switch Section: William S. Parsons, Centralab, Milwaukee, Wisc.
Transformer Section: R. A. Hoagland, Jefferson Electric Company, Bellwood, Ill.
Variable Condenser Section: G. F. Behringer, The American Steel Package Co., Defiance, O.
Variable Resistor Section: D. S. W. Kelly, Allen-Bradley Company, Milwaukee, Wisc.
Vibrator Section: Ray F. Sparrow, P. R. Mallory & Co., Inc., Indianapolis, Ind.
Wire Section: R. G. Zender, Lenz Electric Mfg. Co., Chicago, Ill.
Wire Wound Resistor Section: D. T. Siegel, Ohmite Manufacturing Company, Chicago, Ill.

Chairmen of the Transmitter Division, announced by division chairman S. P. Taylor, distributor sales manager of Western Electric Co., New York, are:

Aviation Section: H. M. Hucke, RCA Victor Division of RCA, Camden, N. J.
Broadcast Transmitter Section: C. W. Miller, Industrial Electronic Division, Westinghouse Electric Corp., Baltimore, Md.
General Communications Section: Natale Gada, General Electric Company, Syracuse, N. Y.
Marine Section: C. E. Maass, Western Electric Company, New York, N. Y.
Transmitter Tube Section: A. Frankel, Lamp Division, Westinghouse Electric Corp., Bloomfield, N. J.

Status of Facsimile

AT THE PRESENT time, facsimile broadcasting is on an experimental basis pending the formulation of

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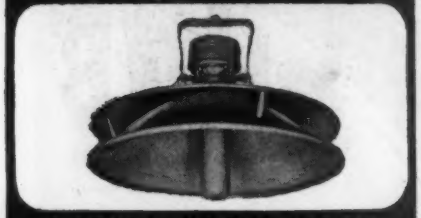

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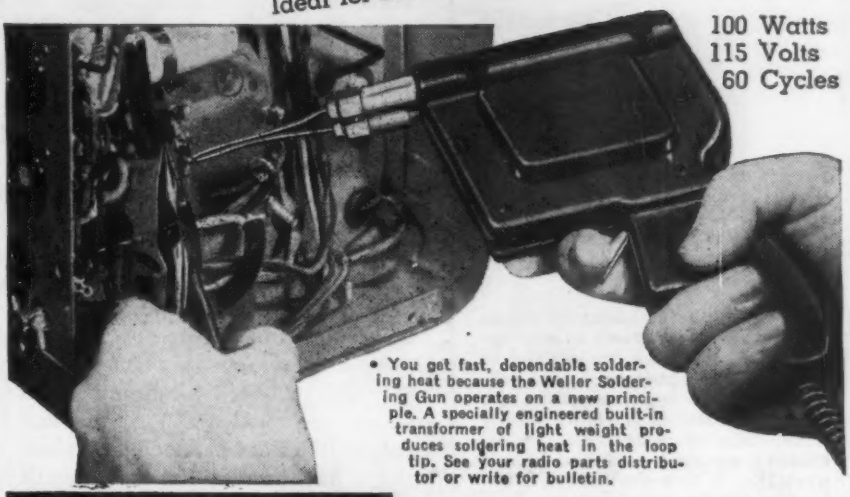

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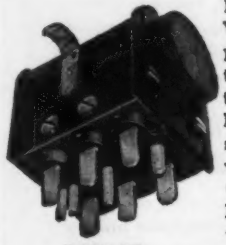
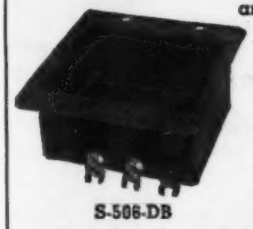
(Heavy Duty)



P-506-CE



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Designed for 5000 Volts and 25 amperes per contact. Socket Contacts of phosphor bronze, knife-switch type, silver plated. Plug Contacts are of hard brass, silver plated. Made in 2, 4, 6, 8, 10 and 12 Contacts. All Plugs and Sockets are Polarized. Long leakage path from Terminal to Terminal and Terminal to ground. Caps and Brackets are of steel, parkerized. Plug and Socket blocks interchangeable in Caps and Brackets. This series is designed for heavy duty electrical work and will withstand severest type of service.

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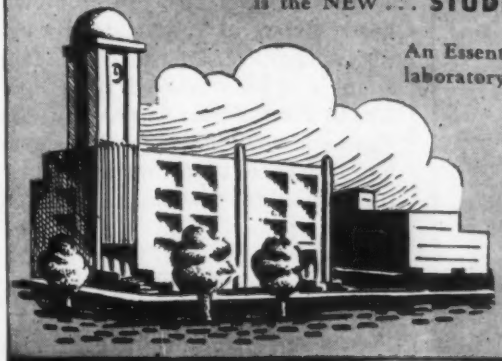

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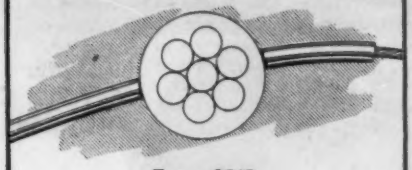

Type SRIR

Approved by the Army-Navy Electronics Standards Bureau (under Spec. JAN-C-76) for Radio Hook-up use where the voltage does not exceed 1,000 V. Sizes: 24 Solid to 6 Stranded.

Recommended for communications and industrial electronic control applications, this wire has proved to be superior to push-back and rubber insulated types because the Synkote insulation is almost totally impervious to water, oils, acids, alkalies, sunlight, cold, and fungus growths. It is vermin-proof, unusually resistant to abrasion, flexing, and tearing, and will not support combustion. In addition, it has extremely high dielectric breakdown strength and low dielectric leakage. In tests performed by the Bureau of Ships, TYPE SRIR did not break down under 12,000 V.

To cut out interfering fields, it can be shielded to any specifications with a braid of tinned or bare copper, woven (loose or tight) to any specified degree of coverage. Gauge of shielding strands varies according to individual requirements.

Other, standard, approved, Synkote Radio Hook-up Wires: Types SRHV, WL, and (Underwriters' Approved) TF, TFF, and Appliance Wiring Material are available in both small and large quantities — shielded, if desired, to any specifications.

Many other types of Hook-up Wire, to meet special requirements, can be manufactured on short notice — in practically any quantity. For complete information, consult our Engineering Department.

Synkote Antenna Wire, and Two-conductor Parallel Cord (with solid, unbreakable, molded plug) are also available.

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rules and standards. Several f-m stations have from time to time been authorized to experiment with facsimile during hours not devoted to regular broadcasting, and these demonstrations have attracted considerable attention.

Since facsimile transmitters and receivers have a lock-and-key relationship, as in television, transmission standards are required so that any facsimile receiver will operate from any facsimile station in its area.

The Radio Technical Planning Board recently submitted proposed transmission standards to the Federal Communications Commission for consideration. Since there has been a difference of opinion in the development of the proposed standards, however, as to whether both 8.2 and 4.1-inch scanning lines should be provided at the same line rate of 105 lines per inch, and since there has been a limited amount of experimental operation and demonstrations to indicate public preference, the Commission has requested that further operation and comparisons be conducted. Upon the completion of such tests, it is believed that standards may be adopted promptly.

Other facsimile activity includes a facsimile news service for airplane passengers, which has been tested in flight. Radioed press dispatches were printed on an airliner in four columns at the rate of 500 words a minute. Operations by a New York bank were aided through rapid transmission of reproductions of checks and other documents from the bank's downtown central signature file to its uptown headquarters in 57 seconds. By using microwave transmission, facsimile has been relayed from Boston to New York. The Army Air Forces is installing a facsimile system to transmit weather maps over the nation for the information of pilots.

BUSINESS NEWS

HOUSTON CORP., Los Angeles, Cal., received a $1,000,000 Navy contract for the manufacture of 100 new airborne radar sets to be used in all four-engine transport planes of the

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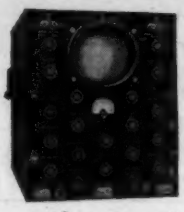
Precision Instruments for Electronics and Radio Research and Testing



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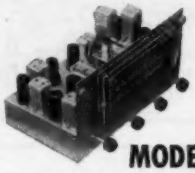

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OTHER EQUIPMENT manufactured by Browning Laboratories includes an accurate frequency meter and ECO Model MJ-9, for operating in the Ham bands, and a frequency meter (Model S-4) especially designed for checking mobile transmitters.

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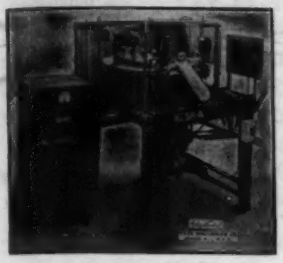

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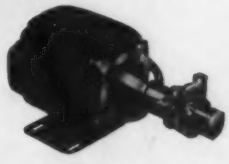
TWO NEW EASTERN PUMPS FOR NONLUBRICATING LIQUIDS

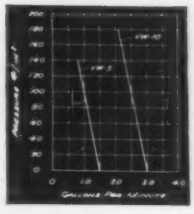
The recently announced VW line of positive pressure vane type pumps for nonlubricating liquids such as water is now supplemented by two new models — VW-5 and VW-10. The basic characteristics of construction that were built into the initial model of the series, VW-1 are incorporated in the new models. Composition bearings that require no lubrication beyond that furnished by the liquid being pumped eliminate all possibility of contamination. Vanes of the same material as the bearings are held in contact with the pump chamber by their special design and radii. In this way volumetric efficiency is maintained without centrifugal force. A built-in bypass is furnished for outlet pressure regulation. Shaft seal is accomplished by a mechanical rotary seal.



Specifications:

Model	Motor H.P.	Weight Lbs.	Length	Width	Height
VW-5A	1/8	26	12 7/16"	5 1/2"	5 3/4"
VW-5B	1/4	32	14 9/32"	6 5/16"	6 21/32"
VW-10A	1/4	34	15 1/8"	6 5/16"	6 21/32"
VW-10B	1/3	34	15 1/8"	6 5/16"	6 21/32"
VW-10C	1/2	52	17"	7 1/2"	7 7/8"

Performance: See chart.



Eastern Industries has engineered more than 300 different pumps for industrial applications. If you have a pumping problem where small size, light weight and high performance are factors, write us. Ask for Bulletin 113 and address all inquiries to Eastern Industries, Incorporated. 296 Elm St., New Haven, Conn.

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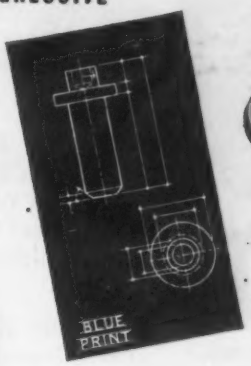
ANOTHER SPECIAL BY PROGRESSIVE

BUILT PROMCO BRAND



ACTUAL SIZE

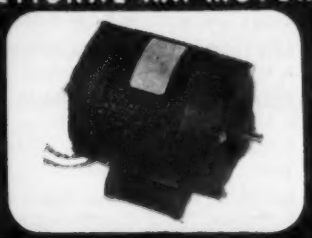
ENLARGED FOR DETAIL

CATALOG SHOWS DOZENS MORE

Special heads, threads and finishes on any metal or alloy adapted to cold upset. Weekly output: 25,000,000 pieces. Many specials, suggesting production savings for you, illustrated in latest catalog. Includes weights per 1M standard pieces, dec. equivs. of fractions, other purchasing and engineering helps. Write for Catalog 19.

The PROGRESSIVE MFG. CO.
50 NORWOOD ST. TORRINGTON, CONN.

NEW Elinco Model G
FRACTIONAL H.P. MOTOR



SYNCHRONOUS OR INDUCTION MOTORS, DESIGNED AS SPLIT-PHASE, SINGLE-VALUE CAPACITOR, OR THREE-PHASE TYPES

THE new Model "G" Elinco motors are rated as synchronous motors at 1/20 h.p. at 1800 r.p.m.; as induction motors at 1/10 at 1700 r.p.m., 115 volts, 60 cycle A.C. A wide variety of other specifications can be filled depending on requirements.

Model "G" motors are 4 3/8" dia. 5 7/8" long, and internal fan cooled. Cast aluminum housings and end brackets, wrinkle finish. Equipped with ball bearings, single shield instrument type. Either base or face mounted. Available with automatic reset thermal overload protection, and as totally enclosed units at reduced ratings.

SPECIAL DESIGN....

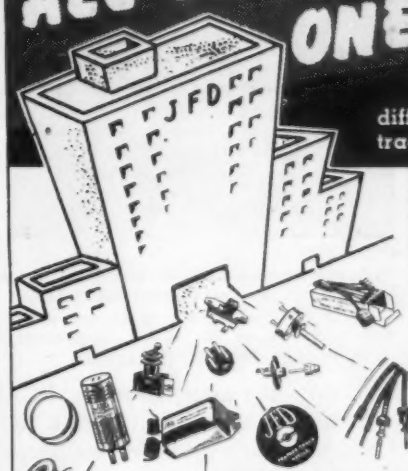
ELINCO DOES NOT MANUFACTURE OR CARRY STOCK MOTORS: ALL UNITS ARE PRODUCED TO ORDER IN SPECIAL HIGH-PRECISION INSTRUMENT-TYPE MOTORS AND GENERATORS—EITHER TO THE BASIC SPECIFICATIONS OF THEIR OVER 250 DIFFERENT TYPES, OR TO WHATEVER PHYSICAL OR ELECTRICAL VARIATIONS ARE REQUIRED TO MEET SPECIAL USES.

ELECTRIC INDICATOR CO.
PARKER AVE. STAMFORD, CONN.

ALL UNDER ONE ROOF!

JFD PRECISION MADE

We manufacture over 4000 different items for the Radio service trade and are prepared to ship

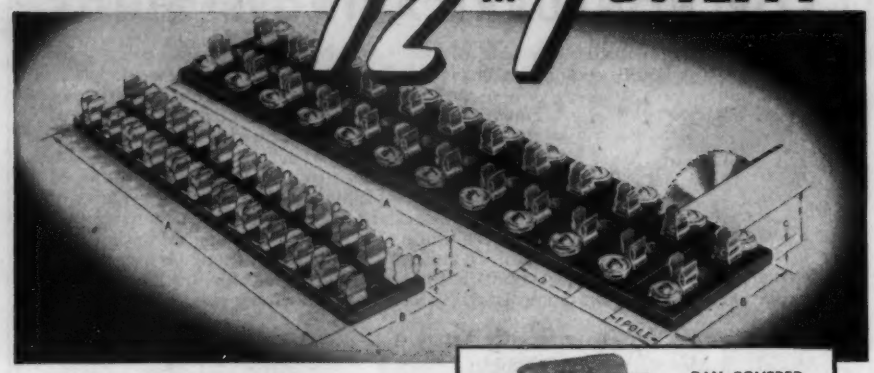


JFD "SOCKETTE" RADIO TUBE ADAPTERS
JFD EXACT DUPLICATE BALLAST TUBES
JFD SERVICEMEN'S ALIGNMENT TOOLS
JFD PHOSPHOR BRONZE DIAL CABLE
JFD FLEXIBLE SHAFTING & CASING
JFD BATTERY ADAPTER HARNESSES
JFD STEPDOWN RESISTANCE CORDS
JFD ADJUSTABLE BALLAST TUBES
JFD RESISTANCE CORD ADAPTERS
JFD SPEAKER CEMENT & SOLVENT
JFD FM & TELEVISION ANTENNAS
JFD REMOTE-O-CABLE REPLACER
JFD PHONO ADAPTER SWITCHES
JFD TRIK HOLD SCREWDRIVERS
JFD MICROPHONE CONNECTORS
JFD FM & TELEVISION LEADS
JFD STEPDOWN TRANSFORMERS
JFD WIRE WINDING MACHINES
JFD MIDGET JACKS & PLUGS
JFD HOME RECORDING DISCS
JFD AUTO RADIO FITTINGS
JFD BAKELITE DIAL KNOBS
JFD PHONOGRAPH NEEDLES
JFD RADIO TUBE SOCKETS
JFD NEUTRALIZING TOOLS
JFD STEPDOWN BALLASTS
JFD RESISTANCE CORDS
JFD AC SERVICE CORDS

JFD TOGGLE SWITCHES
JFD AUTO CONDENSERS
JFD ANTENNA LOOPS
JFD AUTO ANTENNAS
JFD BATTERY PLUGS
JFD NOISE FILTERS
JFD DIAL POINTERS
JFD TUBE SHIELDS
JFD SUPPRESSORS
JFD RADIO WIRE
JFD DIAL BELTS
JFD TEST LEADS
JFD DIAL CORD

J.F.D. MANUFACTURING CO. 4111 FT. HAMILTON PARKWAY BROOKLYN 19, N. Y.

12-IN-1 UTILITY



New 12-pole Littelfuse unit saws into single or multiple mountings.

Be instantly ready to supply any length open fuse-mounting panel; also simplify your parts inventory. Stock all five available styles of this new Littelfuse 12-pole unit. Saw them to 1, 2, 3, 4 or more pole-lengths in your own plant as needed; or order them cut to your specifications.

Solder-terminal types are available in 3 and 8 AG capacities. Terminals are integral parts of clips.

Screw-terminal types available in 3, 4 and 5 AG capacities. Send for Littelfuse Catalog No. 9 today.

CAN COVERED MOUNTINGS FOR 3 AG FUSES

Fatigue-resistant nickel plated phosphor bronze clips—mounted on black Bakelite base. Solder terminals extend through base. Available in single and double pole and combination types.

3 AG "SLO-BLO" LITTELFUSES

High time lag withstands heavy surges—yet is quick on shorts. Designed for use with magnets, solenoids, etc.; and for intermittent duty circuits with heavy starting currents. Anti-fatigue construction.

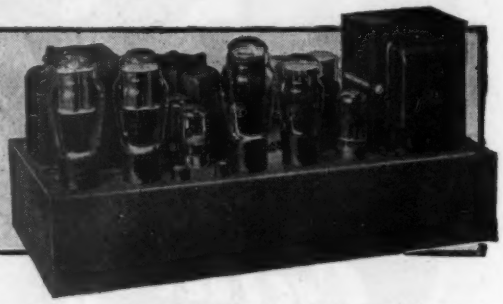
LITTELFUSE FIRST IN CIRCUIT PROTECTION *Incorporated*

4789 N. RAVENSWOOD AVE. CHICAGO 40, U. S. A.

MITE-T-LITE · SWITCH-LITE · IGNITION-FRITZ · NEON INDICATORS · SWITCHES · CIRCUIT BREAKERS · FUSES, MOUNTINGS AND ACCESSORIES

TRIODES.....

The superiority of a triode amplifier is most apparent in the final test...listening



★TRIODE TUBES used throughout. 2—6B4G, 4—7A4, 2—7N7, 1—5U4G, 1—5Y3G.

★Three push-pull stages preceded by an inverter stage.

★Interstage transformer insures good push-pull balance.

★Flat within 1 db to 25 cycles at full power and to 4 cycles at reduced power.

★Flat within 0.2 db to 30,000 cycles.

★0.6% harmonic distortion and 0.2% intermodulation distortion at 5 watts.

★Rated power—30 watts at 2½% total distortion.

★AUTOMATIC BIAS CONTROL greatly increases undistorted power at moderate cost.

★Gain—55 to 120 db in various models.

★Bass and treble compensation—Two-stage tapped condenser-resistor networks.

★Input—38, 150, 500/600 and 500,000 ohms. Output—1.5 to 30 ohms and 500 ohms.

★Power available for other units—250 volts, 0.090 A DC; 6.6 volts, 5 A.

★Fuses—Main power and 6B4G plate line.

★Attractively finished chassis. High quality components. Finest workmanship.

The BROOK HIGH QUALITY AUDIO AMPLIFIER

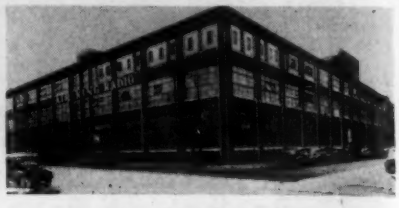
BROOK

Designed by LINCOLN WALSH

BROOK ELECTRONICS, Inc., 34 DeHart Place, Elizabeth 2, N. J.

Naval Air Transport Service. The new APS-42 radar, weighing approximately 150 lb, uses a 5-inch ppi screen to give the pilot a radar map of 220 degrees of terrain ahead of the plane if the radar scanner is mounted in the nose, or a complete 360-degree map if the scanner is placed under the fuselage.

AIR KING PRODUCTS CO., INC., Brooklyn, N. Y., has acquired an additional plant, occupying an entire city block at 170 53rd St., for straight-line manufacture of radio sets.



New Air King plant

RAY-O-VAC CO., Madison, Wisconsin, announces formation of a subsidiary, the Specialty Battery Co., which will manufacture dry batteries for highly specialized purposes in the electronics field.

CASE SCHOOL OF APPLIED SCIENCE, Cleveland, Ohio, recently changed its name to Case Institute of Technology.

NORTH AMERICAN PHILIPS CO., INC., New York City, has acquired a new 2½-story concrete factory building in Mount Vernon, N. Y. for expansion of x-ray production.

SOLAR MFG. CORP. has moved its general offices from New York City to its main Eastern plant in North Bergen, N. J.

MAGNAVOX CO., Ft. Wayne, Indiana, has begun construction of a $1,000,000 factory at Paducah, Kentucky for manufacture of loudspeakers, transformers, and household appliances.

New Magnavox plant now being built

TECHNICRAFT LABORATORIES, INC., Waterbury, Conn., has purchased the microwave and electronic laboratory facilities of the American

★STAR
VITAL AIDS
FOR THE
Manufacturers of 7-pin and 9-pin miniature tube—radios and equipment

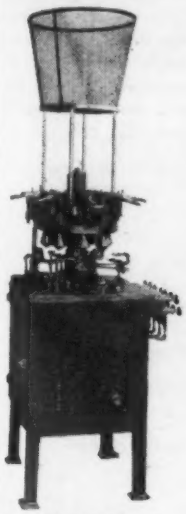
#JE-9 (9-pin)—#JE-10 (7pin)—Star Miniature socket wiring plugs for accurate alignment of miniature socket contacts during wiring. Precision cast of zinc base alloy—pins of stainless steel.

#JE-15 (9-pin)—#JE-13 (7-pin)—Star Miniature tube pin straighteners (with stainless steel insert) to obtain a perfect fit when the tube is placed in the equipment.

Scientifically designed—Precision made
READY FOR IMMEDIATE DELIVERY IN ANY QUANTITIES

STAR EXPANSION PRODUCTS CO. INC.
147 Cedar St. New York 6, N. N. Y.

BAACH-INTERNATIONAL



EIGHT HEAD HOT CUT FLARE MACHINE

- Automatic throughout.
- Can be synchronized with automatic Stem machine.
- Cuts off and flares in one operation.
- Production 1250 flares per hour. For miniature flares, fluorescent starters, standard size lamps, fluorescent and radio tubes.

RANGE OF MACHINE
Glass tubing 27 to 45 gauge
Length of flares 5 mm. to 80 mm.
Forms flares up to 47 mm. diam.
Net weight, 960 lbs. Gross weight 1450 lbs.

Dimensions 24"x24"x72" high

INTERNATIONAL MACHINE WORKS
Manufacturers of High Vacuum Pumps, Automatic Machinery for Incandescent Lamps, Electronic Tubes since 1916.
2027 - 46TH STREET
NORTH BERGEN, N. J., U. S. A.
Tel. UNion 3-7412
Cable Address "Intermach" North Bergen, N. J.

Use LINDE Synthetic Sapphire At Points of Wear

LINDE synthetic sapphire offers definite advantages for small parts at points of wear. Sapphire is hard—takes and retains a high polish. These properties show why.

1. Hardness (Knoop) 1,525 to 2,000
2. Melting Point . . 2,030 deg. C.
3. Unicrystalline Structure
4. Chemical Resistance . All acids
5. Coefficient of Friction . . 0.140 (ring bearing against high-carbon steel pivot)

Write for the LINDE Synthetic Sapphire Technical Data Sheet No. 3. It may suggest further uses where you have problems of small parts wear.

THE LINDE AIR PRODUCTS COMPANY
Unit of Union Carbide and Carbon Corporation
30 East 42nd Street UCC New York 17, N. Y.
In Canada:
Dominion Oxygen Company, Limited, Toronto

The word "Linde" is a trade-mark of The Linde Air Products Company

AMPERITE

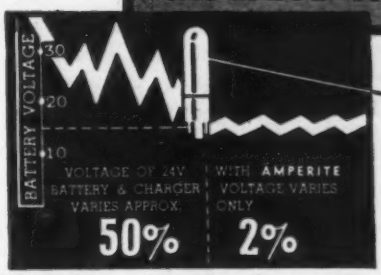
THERMOSTATIC METAL TYPE DELAY RELAYS

PROVIDE DELAYS RANGING FROM 1 TO 120 SECONDS

FEATURES:—Compensated for ambient temperature changes from —40° to 110° F . . . Hermetically sealed; not affected by altitude, moisture or other climate changes . . . Explosion-proof . . . Octal radio base . . . Compact, light, rugged, inexpensive . . . Circuits available: SPST Normally Open; SPST Normally Closed.

PROBLEM? *Send for "Special Problem Sheet" and Bulletin.*

AMPERITE REGULATORS

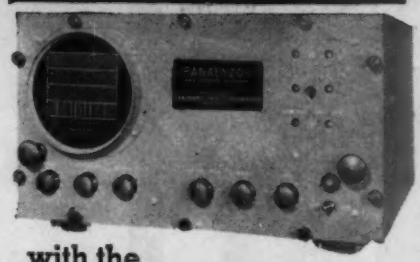


Amperite REGULATORS are the simplest, lightest, cheapest, and most compact method of obtaining current or voltage regulation . . . For currents of .060 to 8.0 Amps . . . Hermetically sealed; not affected by altitude, ambient temperature, humidity.

Write for 4-page Illustrated Bulletin.

AMPERITE CO., 561 Broadway, New York 12, N. Y.
In Canada: Atlas Radio Corp. Ltd. 560 King St. W. Toronto

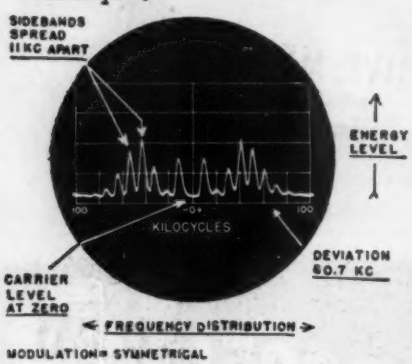
Simpler FM Analysis



with the

PANALYZOR

Eliminates tedious, time consuming point by point frequency checks. It shows simultaneously, in one complete picture, an FM'd carrier and resultant sidebands . . . in terms of relative frequency, amplitude and stability.



A single observation enables determination of such performance details as frequency deviation, energy distribution, sideband content, carrier shift and modulation symmetry . . . Operating procedures are simple . . . interpretations clear cut.

Actually, the PANALYZOR is a panoramic spectrum analyzor which shows, distributed in frequency, discrete quantities of r-f energy as vertical deflections on a cathode-ray tube.

Standard models now available with maximum scanning widths of 50 KC to 20 MC and corresponding resolutions of 2.5 KC to 100 KC.

Write, wire or phone now for recommendations, specifications, prices and delivery time.

PANORAMIC RADIO CORPORATION
242 250 WEST 55TH ST
Exclusive Canadian Representative: Canadian Marconi, Ltd.

Brass Company and the production equipment for the manufacture of flexible waveguides.

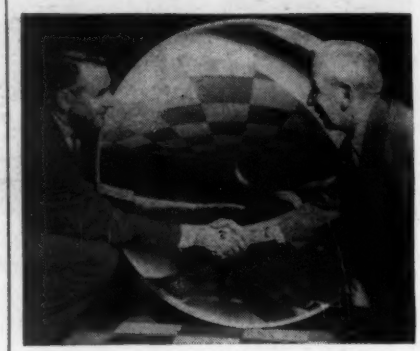
INSULINE CORP. OF AMERICA recently added another 25,000 sq ft to its modern plant facilities in Long Island City, N. Y.

ARLINGTON ELECTRICAL PRODUCTS, INC., 18 W. 25th St., New York 10, N. Y., was recently formed for the manufacture of audio equipment, motion picture items, and allied products.

ARCTURUS RADIO & TELEVISION CORP., Newark, N. J., newly-formed associate company of Standard Arcturus Corp., is now producing television receivers.

KENT CLIFF LABORATORIES, Peekskill, N. Y., has been organized by David F. Sklar for consultation, development, and manufacture of precision electronic and mechanical products.

RCA VICTOR DIVISION announces signing of a contract with Warner Bros. Pictures, Inc. for a joint program of research on large-screen television for theaters. In connection with the announcement, RCA executive vice-president Frank M. Folsom declared, "I am confident that in 1967 this company will be observing the 20th anniversary of large-screen television in the motion picture industry."



RCA vice-president W. W. Watts (left) and Col. N. Levinson of Warner Bros. demonstrate magnifying power of 42-inch spherical mirror used in RCA's large-screen television projection system

PERSONNEL

KENNETH A. SMITH, former RCA field engineer, has joined Primary Processes Ltd. of Los Angeles, Cal., as consultant and development director. He will specialize in re-

FOR PERSONALIZED ATTENTION TO YOUR RESISTANCE WIRE PROBLEMS

CONSULT JELLIFF

When confronted with any resistance problem, take advantage of the diversified experiences of Jelliff in selecting the proper alloys for your specific applications.

For recommendations, literature, prices and delivery of Jelliff Quality Alloys get in touch with our nearest sales representative or communicate direct with Southport, Connecticut. Write or phone for Prompt Action.

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William Maxwell Co., 107 N. Wacker Drive

CLEVELAND, OHIO Phone: MAIN 8585
A. J. Loeb Sales Co., 1836 Euclid Ave. So.

LOS ANGELES, CALIF. Phone: TRINITY 7353
Perlmuth-Colman Associates, 942 Maple Ave.

MINNEAPOLIS, MINN. Phone: GENEVA 3373
Volco Company, 622 McKnight Building

NEW YORK, N. Y. Phone: CALEDONIA 5-1776
R. B. Dana Company, 101 Park Ave.

PHILA., PA. Phone: KINGSLEY 5-1205
S. K. MacDonald, 1531 Spruce St.

PITTSBURGH, PA. Phone: CEDAR 3000
Wm. M. Orr Co., 1228 Brighton Rd.

ROCHESTER, N. Y. Phone: MONROE 5392
J. R. Hanna, P. O. Box 93, Brighton Station

SEATTLE, WASH. Phone: SE-0193
Perlmuth-Colman Associates, 704 Third Avenue

HULL, QUE., CANADA
Mica Co. of Canada, Ltd., P. O. Box 189

LEKTROMESH • RESISTANCE WIRE • FILTERS • STRAINERS • WIRE CLOTH • PARTS • WIRE MESH • DIPPING BASKETS

The C. O. JELLIFF MANUFACTURING CORPORATION
SOUTHPORT, CONN.
WIRE PRODUCTS SINCE 1880

POLARAD
TELEVISION *Equipment*
for studio • laboratory • manufacturer

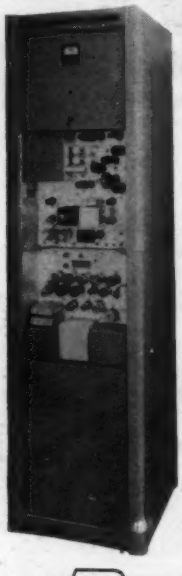
SYNCHRONIZING GENERATOR
Model PT 101—Television

FEATURES

- Built-in 3" oscilloscope with synchronized sweeps for viewing Timing and Video Output pulse wave forms.
- Synchronized marker system for checking pulse width and rise time.
- Extreme stability, insured by deriving all pulses from leading edge of master oscillator pulse.
- Means for checking synchronizing pulses in odd and even fields.

SPECIFICATIONS

525 line, interlaced, 60 fields, 30 frames, RMA Synchronizing pulses held to tolerance specified in the NRTPB report of 1945. Output Pulses: Synchronizing, Video Blanking, Camera Blanking, Horizontal Driving, Vertical Driving Pulses. 5 volts across 100 ohm termination. Dual output jacks. 115 volts 50/60 cps. Complete with tubes.



TELEVISION MONOSCOPE SIGNAL SOURCE
Model PT 102

- Composite Video Signal
- Wide Band Video Amplifier, 6 DB down at 10MC
- Dual outputs for feeding two 75 or 100 lines
- Black positive or Black negative output
- Resolution greater than 600 lines

INPUT: Vertical and Horizontal Driving pulses, Camera and Kinescope Blanking Pulses.

OUTPUT: Composite Video Signal, 3 volts, 100 ohm line 115 volts 50/60 cps. Complete with tubes and including high and low voltage power units.

•

9 FERRY STREET
NEW YORK 7, N. Y.

Polarad *Electronics Company*

Television engineers and consultants to the nation's great television stations.

search and installation of new applications of electronics and ultrasonics in food packaging, processing, and sterilization.

ROBERT F. JONES, JR., Lima, Ohio, received unanimous Senate confirmation as new member of the FCC. He will resign as a congressman to fill the post left by former Commissioner Ray C. Wakefield.

R. F. Jones, Jr.

M. V. Kiebert, Jr.

MARTIN V. KIEBERT, JR., has joined the firm of Sherman Fairchild and Associates, New York City, and will serve as consultant on aircraft telemetering, radar, servo mechanisms, remote control, and instrumentation.

NICHOLAS J. ROSE, now with National Radio Research Laboratories in New York City, will specialize in mechanical applications of electronic controls on a consultant basis.

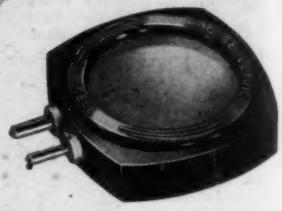
P. C. SANDRETTO has been named director of aviation for the International Telecommunication Laboratories. He joined IT&T late last year upon leaving the Army as a full colonel and receiving a citation and a Bronze Star for his electronics work and its contribution to the efficiency of the B-29 program in the Central Pacific.

CHARLES J. PANNILL, with RCA since 1928, recently retired as president and director of Radiomarine Corp. of America.

HARRY C. INGLES has been elected a director of Radio Corporation of America, succeeding Edward W. Harden who retires after serving on the board since 1919.

L. M. TEMPLE has been appointed chief engineer of the battery division of the Winchester Repeating Arms Company division of Olin Industries, Inc. He was formerly chief engineer of the Eveready divi-

BRADLEY
PHOTO ELECTRIC CELLS



SIMPLIFY PHOTO-ELECTRIC APPARATUS

Luxtron* photocells convert light directly into electrical energy. No external source of voltage is required. Meters and relays can be operated directly from Bradley Luxtron photocells, improving control over manufacturing operations, reducing your costs. They meet the most exacting requirements. Advanced manufacturing techniques make light-actuated Bradley cells the choice all over the world.

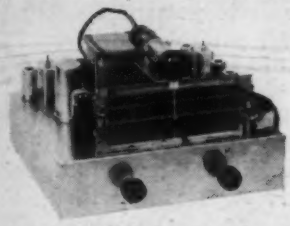
Besides the housed model shown with its plug-in contacts, Bradley also offers tube socket, nut-and-bolt types and pigtail contact mountings. In addition, Luxtron unmounted cells are available in many different sizes and shapes.

*T. M. REG. U. S. PAT. OFF.

Illustrated literature, available on request, shows more models of Bradley photocells, plus a line of copper oxide and selenium rectifiers. Write for "The Bradley Line."

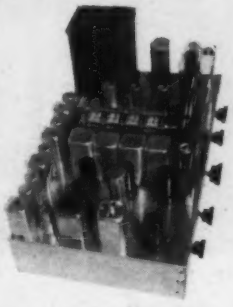
BRADLEY
LABORATORIES, INC.
82 Meadow St. New Haven 10, Conn.

COLLINS
FM and FM/AM TUNERS



11 tube FM chassis unit

17 TUBE FM/AM TUNER. CUSTOM BUILT TO YOUR ORDER



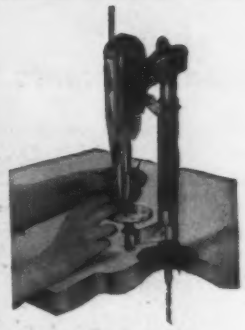
Write for literature on our FM components and matched amplifiers

Collins Audio Products Co., Inc.
126 Park Street Westfield, N. J.

speed production — cut costs with
EJECT-O-MATIC
AUTOMATIC-FEED SOLDERING IRONS AND ACCESSORIES

HEAVY DUTY EJECT-O-MATIC
High-heat 100 and 150-watt models for general heavy electrical, electronic and repair work. Also available in long-nosed "instrument" model. Weight only 22 oz.

STANDARD EJECT-O-MATIC
Popular 50 and 75-watt models for general radio and electrical work. Special long-nosed models available for soldering inside deep receptacles, hard-to-get relays and assemblies. Weight only 18 oz.



VERTI-MOUNT
Treadle operated — leaves both hands free to hold work. Pre-heats and solders work with one easy foot movement. Takes all Eject-O-Matic irons.

complete line now available

TIPS— Multi-clad tips available in eight different sizes and shapes.

SPECIAL TIPS DESIGNED— MINIMUM 10 TO ORDER.

Send for new catalog and prices.

AUTOMATIC FEED SOLDERING IRON
EJECT-O-MATIC
HOUSES · MEASURES · EJECTS THE SOLDER

MULTI-PRODUCTS TOOL CO. 123 SUSSEX AVENUE, NEWARK, N. J.

Linde

ARGON
HELIUM
KRYPTON
NEON
XENON
and Standard or Special
MIXTURES

LINDE rare gases are spectroscopically pure—argon, helium, neon, and standard mixtures are available in one- and two-liter glass bulbs and in cylinders; xenon and krypton are available in liter and fractional-liter bulbs.

The word "Linde" is a trade-mark of

THE LINDE AIR PRODUCTS COMPANY
A Unit of Union Carbide and Carbon Corporation
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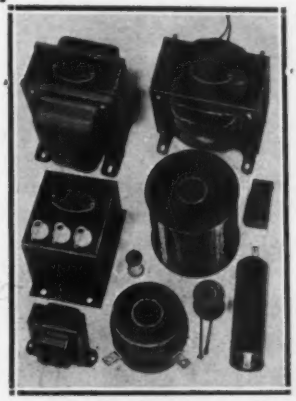

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sion, National Carbon Co., and there developed an economizer circuit for portable radios.

PHILIP I. MERRYMAN, formerly manager of the planning and development division of NBC, has formed a new firm of radio consultants in partnership with H. V. Anderson of New Orleans.

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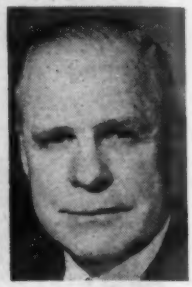

S. Bracken

STANLEY BRACKEN, executive vice-president of Western Electric Co., was elected president to succeed retiring president Clarence G. Stoll.

CLARK C. RODIMAN, managing editor of QST for over 10 years before the war, is now with National Co., Inc., Malden, Mass.

C. C. Rodiman

R. K. Dixon

ROBERT K. DIXON, previously engaged in radar beacon development at Submarine Signal Co., is the newly appointed product manager of broadcast equipment in Raytheon's Commercial Products Division.

GEORGE J. MAKI, previously associated with Collins Radio Co., has established his own office as radiotelegraph consultant at Moraga, Cal.

ARTHUR WINTERS has established Winters Radio Laboratory at 11 Warren St., New York City for specialized servicing of communication equipment and television receivers. His firm is now the authorized Hallicrafters service center for the northeastern area.

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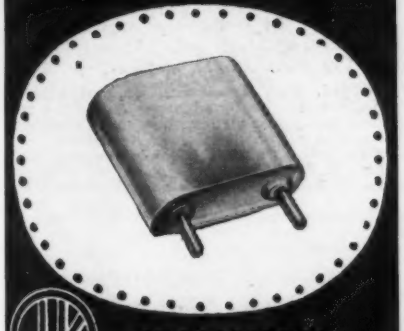

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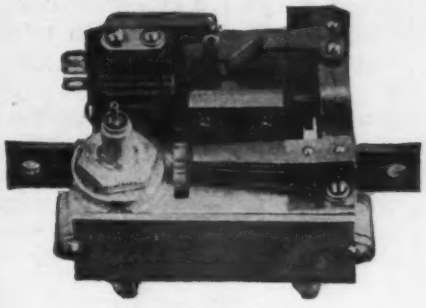

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NEW BOOKS

Klystron Tubes

BY A. E. HARRISON, *Assistant Professor of Electrical Engineering, Princeton University; formerly klystron applications engineer, Sperry Gyroscope Co. McGraw-Hill Book Co., New York, 1947, 271 pages, $3.50.*

BEHAVIOR of various types of klystrons is explained as a detailed introduction to them for both tube designers and electronic application engineers. The book coordinates functions of cavities and electron beams as used in amplifier, multiplier, oscillator, and detector klystrons. Methods of modulation and measurement are also described.—F.R.

Theory and Application of Radio-Frequency Heating

BY GEORGE H. BROWN AND CYRIL N. HOYLER, *research engineers at RCA Laboratories, and* RUDOLPH A. BIERWIRTH, *chief engineer of Sound Inc., Chicago. D. Van Nostrand Co., Inc., New York, N. Y., 1947, 370 pages, $6.50.*

FOR INDUSTRIAL engineers considering the use of r-f generators for solving heating problems, the authors have combined in this one distinguished volume the pertinent basic principles, design equations, and practical application data for both induction and dielectric forms of electronic heating.

Maximum emphasis is placed on the nature of current flow in various metal shapes and on the design of work coils for surface or all-through heating of good conductors, and rightly so because this form of electronic heating predominates in manufacturing plants. The first ten chapters thus lay the groundwork for chapter 11 on annealing brass and bronze, chapter 12 on typical induction heating problems such as paint baking and multiple soldering, three following chapters on heat flow equations for metals, and chapter 16 on case hardening of steel. Included among the first ten chapters are details of such specific applications as heat treatment of knitting needles and surgical needles, heating of thin metal strip, and heating of laminar sheets such as copper-clad steel.

Of the eight chapters devoted to heating of poor conductors, chapter

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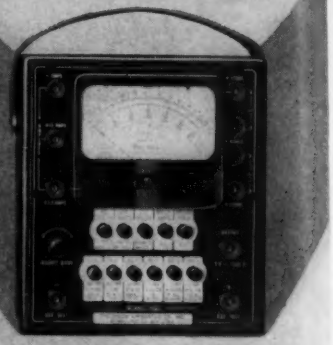

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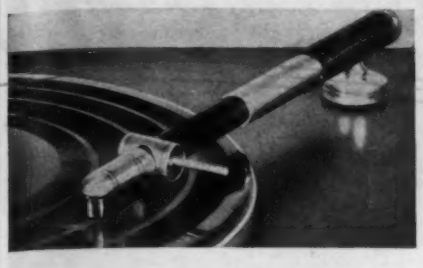

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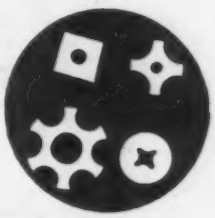

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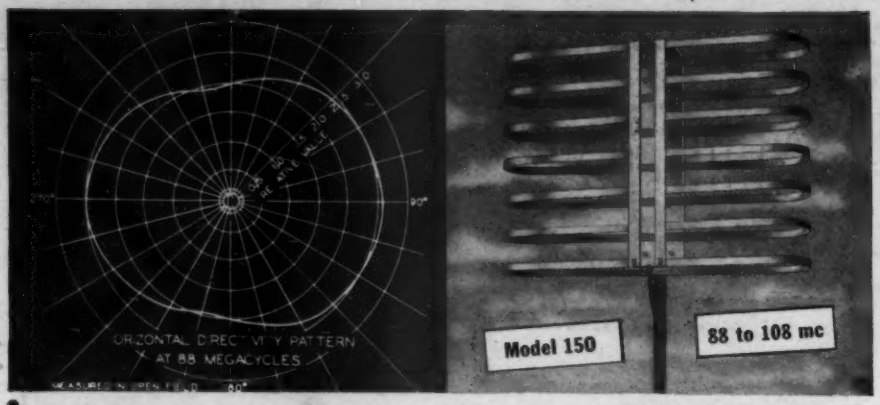

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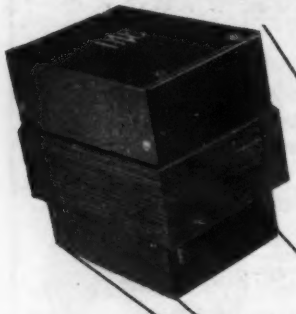

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17 deals solely with theory and the remainder cover specific applications. These include wood-gluing presses, r-f sewing machines for sheet plastics, r-f dehydration, pasteurization, sterilization, and cooking of food, and heating of drugs and other materials.

The mathematical derivations throughout the book, supplemented by a three-part mathematical appendix, make this excellent book equally as valuable to the design engineer as to the industrial engineer concerned primarily with the engineering of economic electronic heating applications in plants.—J.M.

The Physical Principles of Wave Guide Transmission And Antenna Systems

BY W. H. WATSON, *Professor of Mathematics, University of Saskatchewan. Oxford University Press, London, 1947, 207 pages, $7.00.*

PROPERTIES of slot couplings and antenna arrays fed from waveguides were extensively studied during the war. In this presentation of the studies, a preliminary chapter reviews the impedance concept used in exploring guided wave phenomena. Propagation in waveguides and auxiliary components is described as background for the development of slot couplings and radiators, the latter constituting the major portion of this book for design engineers and physicists. —F.R.

German-English Dictionary for Electronics Engineers and Physicists

BY BERNARD R. REGEN AND RICHARD R. REGEN. *Published by J. W. Edwards, Ann Arbor, Mich., 1946, 358 pages. $6.00.*

NEARLY 21,000 German technical terms in the fields of electronics and physics are listed alphabetically with English equivalents, providing an invaluable key to the great mass of German-language literature collected in Germany since the war and available from the Superintendent of Documents, Washington, D. C. In many instances brief explanatory definitions have been added to clarify the

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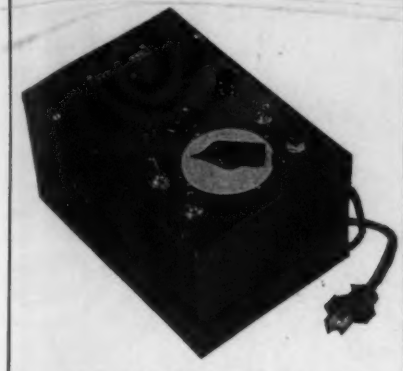

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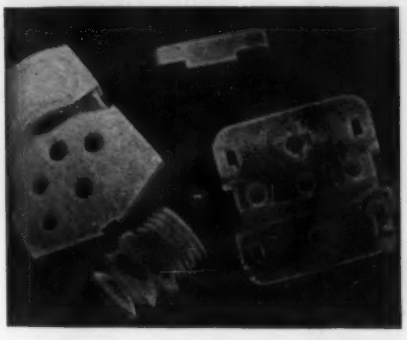

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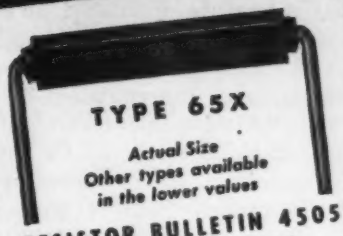

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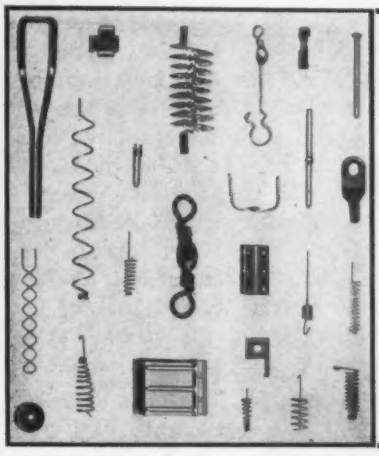

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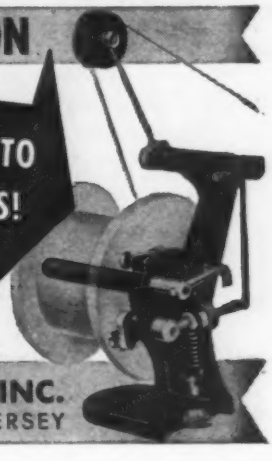

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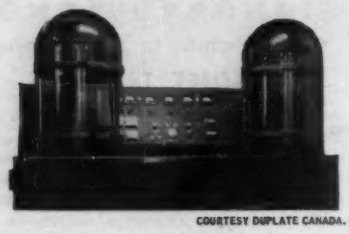
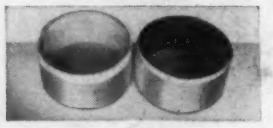


COURTESY DUPLATE CANADA, LTD.

Use of DC 703 in production units for the thermal evaporation of metallic and non-metallic films indicates that the exceptional heat stability of this DC Silicone Diffusion Pump Fluid results in better adhesion of magnesium fluoride films to glass.

After using DC 703 for the past year in two production units for evaporating magnesium fluoride and rhodium, Dr. Bateson of Duplate Canada Ltd., Oshawa, Ontario, reports that he has had no difficulty in producing equal or higher vacuum with pumping speeds equal to those of a high grade organic pumping fluid.

Even more significant are Dr. Bateson's observations that backstreaming seems to be reduced by use of DC 703; that the Silicone-vapor which does migrate into the system is more heat-stable; and that primary adhesion of magnesium fluoride films to glass is therefore improved. Preliminary observations also indicate that use of DC 703 results in brighter first surface mirrors.



That DC 703 causes less contamination of the system is further indicated by these aluminum shields. Both were normally exposed to the heaters in the bell jar for 9 months. Shield at left, exposed to DC 703, is quite clean. The other shield, exposed to a high grade organic pumping fluid, is covered with a dark brown deposit.

May we also remind you of DC 702. Both DC 702 and DC 703 are useful in electron microscopes, high vacuum dehydration and in evacuating electronic tubes. Write for pamphlet No. 9-4.

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FIRST IN SILICONES

distinction between similar terms in both languages.

Terminology current in 1939 is covered, along with all other terms falling within the scope of the dictionary that could be found in books, scientific journals, and patent literature published in Germany during the war and available to the authors up to the time of publication. Thoroughness of coverage of electronic terms is evidenced by inclusion of 75 entries for antennas, 65 for recording, 80 for electrons, 100 for frequency, 85 for grids, grates, and lattices, 40 for cathodes, 40 for capacitors, 45 for loudspeakers, 70 for direction finding, 150 for tubes, 120 for waves, etc. A patent-practice vocabulary is included, adding greatly to the value of this book in an engineer's library.—J.M.

Television Primer of Production and Direction

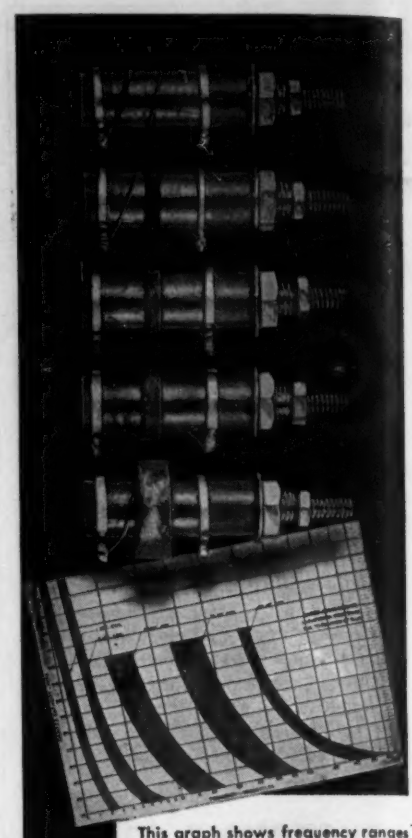
BY LOUIS A. SPOSA. *McGraw-Hill Book Co., Inc., New York, N. Y., 1947, 237 pages, $3.50.*

THOUGH dealing primarily with techniques of producing television programs, this little book contains much information of value to engineers and their assistants at television stations, as well as those concerned with the design and manufacture of cameras and other equipment used in television studios. The greater the familiarity of an engineer with the end uses of his product, the more useful and versatile will his product be.—J.M.

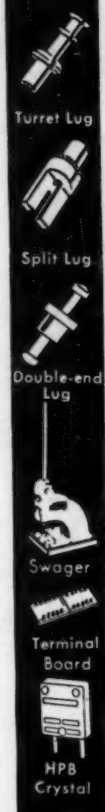
Writing the Technical Report

BY J. RALEIGH NELSON, *Professor Emeritus, University of Michigan, McGraw-Hill Book Co., New York, 1947, 388 pages, $3.00.*

ENGINEERS are judged not on what they know, but on the extent to which they clearly present their knowledge to company executives, clients, and other engineers. Frequently report writing is neglected because of pressure to learn additional technical material. The result, as editorialized recently in the British *Electronic Engineering*, is that excessive verbiage, ambig-



This graph shows frequency ranges covered by each unit. Write us for your full-size copy.



Five Standard Slug-Tuned LS3 Coils Cover ½ to 184 mc

For strip amplifier work, the compact (1⅛" high when mounted) LS3 Coil is ideal. Also for Filters, Oscillators, Wave-Traps or any purpose where an adjustable inductance is desired.

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CTC LS3 Coils are easy to assemble, one ¼" hole is all you need. Each unit is durably varnished and supplied with required mounting hardware.

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uity, and vague expressions in reports are taken by readers to indicate lack of thought. The present book examines report organization and preparation, and describes good practices of style and format. An especially valuable feature is the criticism of typical reports. Suggestions for class use are included at the end, should the book be used in colleges.—F.R.

The Future of Television

BY ORRIN E. DUNLAP, JR. *Harper & Brothers, New York, N. Y., revised edition, 1947, 194 pages, $3.00.*

REVISION of the 1942 edition brings it up to date and looks still farther ahead. The appendix now lists 150 historic steps in television from 1867 (Maxwell's wave theory) to 1947. Though dealing chiefly with economic and programming aspects, engineers will find here many answers to the kinds of television questions posed by nontechnical friends.—J.M.

Electronic Engineering Principles

BY JOHN D. RYDER, *Professor of Electrical Engineering, Iowa State College, Prentice-Hall, New York, 1947, 397 pages, $6.65.*

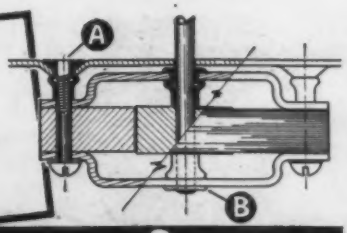
WRITTEN to present electronics to all engineering students, this book embraces electronics, tubes and tube characteristics, and the most common amplifier, rectifier, and control circuits. Although intended for class-room use, the book is sufficiently clear and complete in itself to be read by those studying the subject alone.—F.R.

Vector and Tensor Analysis

BY LOUIS BRAND, *Professor of Mathematics, University of Cincinnati, John Wiley and Sons, New York, 1947, 439 pages, $5.50.*

THE SUBJECT is presented in the form of assertions about the mathematical properties of vectors and tensors, thus giving a grounding in their characteristics and manipulation. However, the mathematical rigor is at times incomplete, and the application of the techniques to engineering is only sketchy, so that the book is more suited to college courses than personal study.—F.R.

HOW RIVNUTS SOLVED A FASTENING PROBLEM



. . . and cut costs 2 ways!

A MANUFACTURER wanted to make a new motor that would give good service—yet he wanted to produce it at minimum cost. B. F. Goodrich Rivnuts helped him solve the problem.

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Rivnuts thus cut costs of the motor two ways: 1) lower material costs, and 2) reduced production time. If you have a fastening problem, why not put it up to Rivnut engineers? Write The B. F. Goodrich Company, Department E-97, Akron, Ohio.

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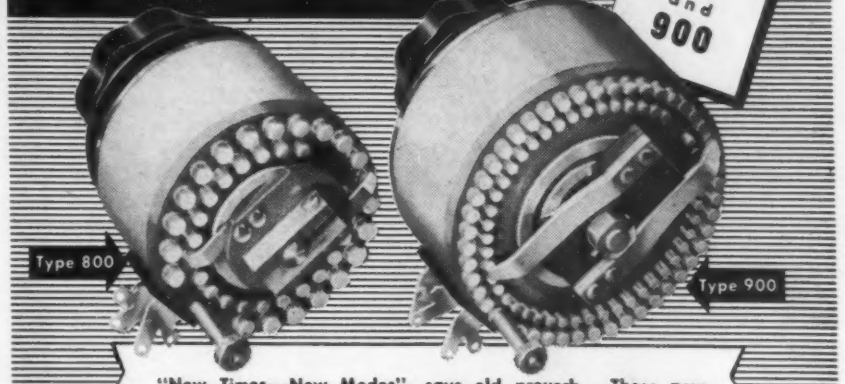
TYPE 800 and 900



"New Times—New Modes", says old proverb. These new attenuators were born to meet new war-created demands. They represent a new medium frame size: Type 800 (2¼" dia.) and a larger size: Type 900 (3" dia.). The Type 800 is supplied as potentiometer, rheostat, ladder and T-pad up to 20 steps. The larger size Type 900 is similarly furnished with up to 45 steps. Write for new bulletin.

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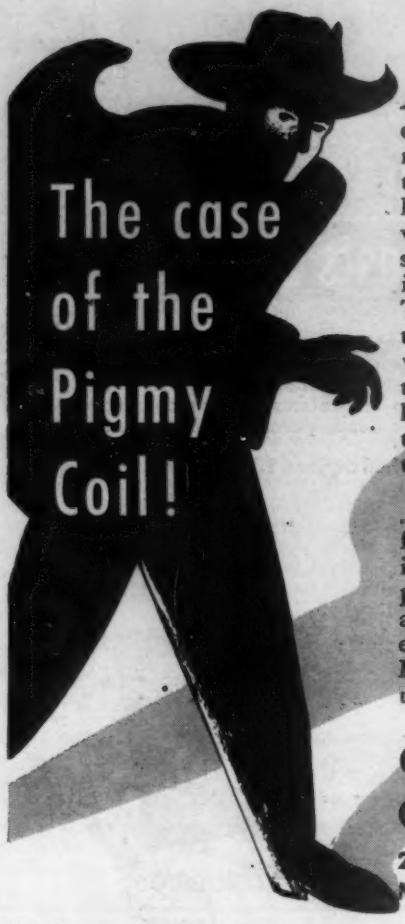

A PROMINENT manufacturer redesigned his product for lower costs and better performance. But the new design called for a coil smaller than he had ever used before. Central Paper Company research engineers were consulted on the problem. They suggested reducing the bulk of the insulating paper used in winding coil. They developed an entirely new paper, thinner than any ever used for coil winding before, yet, because of its controlled and greatly reduced porosity, higher dielectric strength and antitarnish properties, the ideal paper for this particular application.

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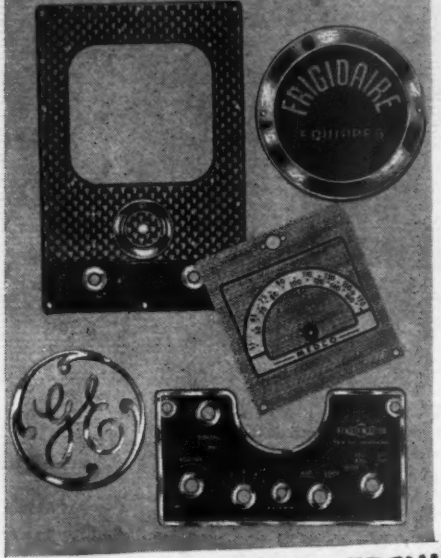

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Backtalk

This department is operated as an open forum where our readers may discuss problems of the electronics industry or comment upon articles which ELECTRONICS has published.

Electrometer Addendum

Dear Sir:

THE TIME necessary for an electrometer tube to reach equilibrium can be substantially reduced if the supply voltages are applied only after the cathode has reached operating temperature. This could easily be accomplished in the amplifier described in my paper, published in the June 1947 issue of ELECTRONICS (p 138), if a double-pole time delay relay of 1-2 minutes delay were inserted into the screen and plate connections of the 12BE6 tube (Fig. 1).

H. S. ANKER
Department of Biochemistry
The University of Chicago
Chicago, Illinois

Reverberation Time Nomographs

Dear Sir:

IT IS pleasing to note Mr. R. C. Coile's attempt to reduce the computations involved in the acoustic treatment of rooms to a simple set of nomographs in his paper on Reverberation Time Nomographs in the April issue of ELECTRONICS.

The alert acoustical engineer will recognize immediately the limitations of the nomograph in Fig. 1 of Mr. Coile's paper. However, judging by the inquiries on the use of acoustic materials received at this Bureau, a novice might be led far astray by using the incomplete information in the nomograph.

First, in the absence of notes or textual reference to the contrary, it might appear offhand that any acoustic material has a unique coefficient of sound absorption at a given frequency. Actually, the coefficient depends very much on how the material is mounted. Published lists[1] show that the variation in sound absorbing power of the same material can vary by as much as

100 percent for different methods of mounting. Consequently, the mounting information should be included in the nomograph along with the material. Otherwise, separate nomographs would have to be constructed for each type of mounting or for each material. Also, the inclusion of a numerical scale on the absorption coefficients scale might increase the usefulness of the nomograph.

Second, it might be inferred from the nomograph that the sound absorption coefficient is directly proportional to the area of an acoustic material. For larger areas, say greater than 60 or 70 square feet, this is approximately true. Occasionally, however, acoustic materials are applied in smaller patches. Experiments[2] show that the coefficient determined for small patches are appreciably greater than for larger areas.

PETER CHRZANOWSKI,
Physicist, Sound Section
National Bureau of Standards
Washington, D. C.

(1) "Sound Absorption Coefficients of the More Common Acoustic Material", Letter Circular published periodically by National Bureau of Standards, Washington, D. C.
"Sound Absorption Coefficients of Architectural Acoustical Materials", published periodically by Acoustical Materials Association, 205 West Monroe Street, Chicago, Ill.

(2) V. L. Chrisler, Dependence of Sound Absorption upon the Area and Distribution of the Absorbent Material, *Jour. Res. N. B. S.*, 13, p 169, 1934.
J. S. Parkinson, Area and Pattern Effects in the Measurement of Sound Absorption, *Jour. Acous. Soc. Am.*, 2, p 112, 1930.

Dear Sir:

DR. CHRZANOWSKI'S questions as to limitations in the use of the nomographs for Reverberation Time given in April ELECTRONICS are well taken. A precise calculation of reverberation time should include the method of mounting the acoustic material, the consideration of the size of the patch as well as its relation to the geometry of the room, the use of the proper acoustical coefficients for a particular frequency, the effect of humidity, and should more properly not use the approximate formula which the nomograph represents.

Surprisingly enough, however, the approximate values that the nomographs give are very close to practical measured values for average rooms such as are used in broadcast stations, offices, and homes.

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Colton & Foss, Inc.
Washington, D. C.

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ADDITIONS and CORRECTIONS

to the

JUNE 1947–1948 ELECTRONICS DIRECTORY ISSUE

CONSULTING ENGINEERS

The following names of consultants were received after publication of a more extensive list in ELECTRONICS BUYERS' GUIDE for June 1947, Mid-Month issue, starting on page 132. Refer to the number key, printed on that page, for a breakdown of the specific Communications and Industrial services available. An asterisk following the name indicates that patents are handled as well.

ADDITIONS:—

Name	Communications	Industrial	Design	Models	Area
BARONE CO. S. A. 143-5 West 22nd St., New York 11, N. Y. Arthur J. Odgers Leslie Norde	1-2-4-10-12-17-20		•	•	international
BUILDER, G., ESQ.* 7 Appian Way, Burnwood, N. S. W., Australia		1	•	•	national
EDGERTON, GERMESHAUSEN & GRIER 77 Massachusetts Ave., Cambridge, Mass.		1-17-22	•	•	national
ELLSWORTH, A. R. 8469 Hollywood Blvd., Hollywood, Calif.	15	15			
FERRAMENTA, M. Rua Joaquim Tavora 300, Santos Est De Sao Paulo, Brazil	1-2	1-2	•		statewide
HOLLEY, ARMANDO* Calle Constitucion 097, Santiago, Chile, S.A.	15	15	•		national
IVINS, H. Y. Box 82, Plymouth, Ind.	1		•		150 miles
LEATHAM, W. G. P. O. Box 11, Lower Hutt, N. Z.	1	1	•		national
LUNDY, CLARENCE 428 Boulevard, Bayonne, N. J.		5	•		
NIAGARA ELECTRON LABORATORIES Box 141, Andover, N. Y. P. M. Hackett		1-5-6 8-17	•	•	
PACENT ENGINEERING CORP.* 79 Madison Ave., New York 16, N. Y. L. G. Pacent Edward W. Noll	1-4-10-12	12-13-20-25-30	•		national
PARDEE, H.* 549 W. Washington Blvd., Chicago, Ill.		20	•		national
PATRICK RESEARCH & DEVELOPMENT LABS.* 5208 N. Keystone Ave., Indianapolis, Ind. E. Patrick		1-2		•	national
POLLACK, DALE 352 Pequot Ave., New London, Conn.	1-2-5-10 11-12-28		•	•	national
RADIO INVENTIONS, INC.* 155 Perry St., New York, N Y.		5	•	•	national
RUBIN, H. B.* 1 Flint St., Poquonock Bridge, Conn.	20		•	•	1500 miles
SACAZES, J. M.* 35 Rue de la Dalbade, Toulouse, Haute Garonne, France	17			•	France
STEPHENS, HERBERT V. 83 Edsen St., Buffalo 10, N. Y.					
WHEELER LABORATORIES, INC. 259-09 Northern Blvd., Great Neck, N. Y. Harold A. Wheeler	5-10-11 12-19-25 26	5-8			national
WISE, ROGER 7 Gracie Square, New York, N. Y.					
ZELINGER, GEZA P. O. Box 1239, Haifa, Palestine	10		•	•	statewide

CORRECTIONS:—

GALANTY, HAROLD
287 Conklin Ave., Hillside, N. J. (not N. Y.)

HAYNES, A.
2227 S. 10th Ave., Broadview, Maywood, Ill. (not Mayview)

PRODUCT LISTINGS

The following listings are to be used in conjunction with the June 1947–1948 ELECTRONICS BUYERS' GUIDE for information on manufacturers' names, addresses, and products omitted or incorrectly listed in that issue. Product classification names are the same as used in the June Guide.

ADAPTERS—*Battery*

Eby, Inc., Hugh H., 18 W. Chelton Ave., Philadelphia 13, Pa.

ADHESIVES

Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.

ADHESIVES—*Waterproof & Corrosion Proof Compounds*

Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.

AMPLIFIERS—*Audio Frequency*

Kinsey Radio Engineering Co., Inc., 1208 Farnam St., Omaha 2, Nebraska

Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.

ATTENUATORS

Daven Co., The, 191 Central Ave., Newark 4, N. J.

AMPLIFIERS—*Audio Input Systems*

Northern Radio Co., 143-145 W. 22nd St., New York 11, N. Y.

BATTERIES—*Dry*

Bowers Battery & Spark Plug Co., Reading, Pa.

Burgess Battery Co., Battery Div., Freeport, Ill.

BATTERIES—*Standard Cell*

Bowers Battery & Spark Plug Co., Reading, Pa.

BATTERY CHARGERS

Bowers Battery & Spark Plug Co., Reading, Pa.

BOXES—*Metal*

Paul & Beekman Div., Portable Products Corp., 1801 Cortland St., Philadelphia 40, Pa.

BUSHINGS—*Nonmetallic*

Franklin Fibre-Lamitex Corp., 12th & French Sts., Wilmington, Del.

CABINETS—*Plastic*

Norton Labs., Inc., Lockport, N. Y.

CAPACITORS—*Variable Receiver Tuning*

Airadio Inc., Melrose Ave. & Battery Pl., Stamford, Conn.

CEMENTS—*Radio*

Moyen, C. P., 908 Chicago Ave., Chicago, Ill.

COATINGS—*Protective*

Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.
Sun Oil Co., 1608 Walnut St., Philadelphia 3, Pa.

CODE & TAB MARKERS

Smith Laboratories, Melvin L., 16 Field St., Kane, Pa.

COIL WINDING MACHINES

Seifert, Inc., E. R., 202 South Beach St., Syracuse, N. Y.

COMMUNICATION SYSTEMS—*Facsimile*

Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.

COMMUNICATION SYSTEMS — *Radio-teletype*

Barker & Williamson, 235 Fairfield Ave., Upper Darby, Pa.

CONNECTORS—*Cable Connectors & Couplings*

Scintilla Magneto Div., Bendix Aviation Corp., Sidney, N. Y.

CONNECTORS—*Coaxial Cable*

Scintilla Magneto Div., Bendix Aviation Corp., Sidney, N. Y.

CONNECTORS—*Receptacle*

Scintilla Magneto Div., Bendix Aviation Corp., Sidney, N. Y.

CONTROLS—*Temperature*

Philadelphia Thermometer Co., 4401 N. Sixth St., Philadelphia, Pa.

CONVERTERS—*Frequency Shift*

Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.

CRYSTAL HOLDERS

Smith Laboratories, Melvin L., 16 Field St., Kane, Pa.

CRYSTALS—*i-f Filter*

Bliley Electric Co., Union Station Bldg., Erie, Pa.

CRYSTALS—*Quartz*

Smith Laboratories, Melvin L., 16 Field St., Kane, Pa.

CRYSTALS—*Receiver Control*

Bliley Electric Co., Union Station Bldg., Erie, Pa.
Smith Laboratories, Melvin L., 16 Field St., Kane, Pa.

DIELECTRICS—*Liquids & Solids*

Sun Oil Co., 1608 Walnut St., Philadelphia 3, Pa.

DIES

Atlas Tool and Design Co., Castor & Kensington Avenues, Philadelphia 24, Pa.

DRAFTING EQUIPMENT

Ozalid Products Div., General Aniline & Film Corp., 770 Ansco Rd., Johnson City, N. Y.

ESCUTCHEONS

Wilson Plastics Div., Wilson Magazine Camera Co., 6022 Media St., Philadelphia 31, Pa.

FILTERS—*Electric Wave Section*

Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.

GENERATORS—*Gas-Engine Driven*

Burke Electric Co., 12th & Cranberry Sts., Erie, Pa.

GENERATORS—*Hand Driven*

Burke Electric Co., 12th & Cranberry Sts., Erie, Pa.

GRILLES—*Molded Plastic*

American Insulator Corp., New Freedom, Pa.
Norton Laboratories, Inc., Lockport, N. Y.

HEARING AIDS

Schulmerich Electronics, Inc., Sellersville, Pa.

HEATING, ELECTRONIC — *Accessories and Associated Equipment*

Sherman Industrial Electronics Co., Inc., 503 Washington Ave., Belleville 9, N. J.

HEATING, ELECTRONIC—*Dielectric*

Sherman Industrial Electronics Co., Inc., 503 Washington Ave., Belleville 9, N. J.

HEATING, ELECTRONIC—*Induction*

Sherman Industrial Electronics Co., Inc., 503 Washington Ave., Belleville 9, N. J.

INDICATORS—*Temperature*

Philadelphia Thermometer Co., 4401 N. Sixth St., Philadelphia, Pa.

INSTRUMENTS—*Electronic Musical*

Wurlitzer Mfg. Co., North Tonawanda, N. Y.

INSULATING COMPOUNDS

Impregnating Compounds

Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.
Sun Oil Co., 1608 Walnut St., Philadelphia 3, Pa.

Insulating Varnish

Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.

Phenolic Molding Compounds

Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.

Protective Sealers

Sun Oil Co., 1608 Walnut St., Philadelphia 3, Pa.

Resins

Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.

Varnish Cambric Coatings

National Varnished Products Corp., 211 Randolph Ave., Woodbridge, N. J.

Waxes

Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.
Sun Oil Co., 1608 Walnut St., Philadelphia 3, Pa.

INSULATION MATERIALS

Ceramic

Scintilla Magneto Div., Bendix Aviation Corp., Sidney, N. Y.

Fabric

National Varnished Products Corp., 211 Randolph Ave., Woodbridge, N. J.

Paper

National Varnished Products Corp., 211 Randolph Ave., Woodbridge, N. J.

INSULATION PARTS

Ceramic Insulation Parts

Scintilla Magneto Div., Bendix Aviation Corp., Sidney, N. Y.

Extruded Insulation Parts

National Varnished Products Corp., 211 Randolph Ave., Woodbridge, N. J.

Fabricated Insulation Parts

Accurate Mfg. Co., Inc., Philadelphia 32, Pa.
National Varnished Products Corp., 211 Randolph Ave., Woodbridge, N. J.
Norton Laboratories, Inc., Lockport, N. Y.

Molded Insulation Parts

American Insulator Corp., New Freedom, Pa.
Norton Laboratories, Inc., Lockport, N. Y.
Scintilla Magneto Div., Bendix Aviation Corp., Sidney, N. Y.

Plastic Insulation Parts

Accurate Mfg. Co., Inc., Philadelphia 32, Pa.
American Insulator Corp., New Freedom, Pa.
Norton Laboratories, Inc., Lockport, N. Y.

Rods

National Varnished Products Corp., 211 Randolph Ave., Woodbridge, N. J.

Stampings & Punchings—Paper

National Varnished Products Corp., 211 Randolph Ave., Woodbridge, N. J.

INSULATORS—*Ceramic*

Scintilla Magneto Div., Bendix Aviation Corp., Sidney, N. Y.

KEYERS & COUPLERS

Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.

KNOBS—*Plastic Molded*

American Insulator Corp., New Freedom, Pa.
Norton Laboratories, Inc., Lockport, N. Y.

LIGHTS—*Photoelectric Source*

Eby, Inc., Hugh H., 18 W. Chelton Ave., Philadelphia 13, Pa.

LUGS AND TERMINALS

Lambert Mfg. Co., 639 West 12th St., Erie, Pa.

METAL PARTS—*Powdered*

Stackpole Carbon Co., St. Marys, Pa.

METAL PARTS—*Small Metal Stampings*

Accurate Mfg. Co., Inc., Philadelphia 32, Pa.
Atlas Tool and Design Co., Caster & Kensington Avenues, Philadelphia 24, Pa.
Lambert Mfg. Co., 639 West 12th St., Erie, Pa.

METALS

Brass and Bronze

Phosphor Bronze Smelting Co., 2200 Washington Ave., Philadelphia 46, Pa.

Phosphor Bronze

Phosphor Bronze Smelting Co., 2200 Washington Ave., Philadelphia 46, Pa.

MONITORS—*Frequency*

Northern Radio Co., Inc., 143-145 West 22nd St., New York 11, N. Y.

(continued on next page)

(continued from preceding page)

MONITORS—(FS) Transmitter
Northern Radio Co., Inc., 143-145 West 22nd St., New York 11, N. Y.

MOTORS—Fractional hp
Marco Industries, Depew, N. Y.
Smith Laboratories, Melvin L., 16 Field St., Kane, Pa.

MOTORS—Phonograph
Marco Industries, Depew, N. Y.

MOTORS—Synchronous
Marco Industries, Depew, N. Y.

MOTORS—Timing
Marco Industries, Depew, N. Y.

MUSIC SYSTEMS—Industrial and Commercial
Schulmerich Electronics, Inc., Sellersville, Pa.
York Electric & Machine Co., Carillotone Div., 30-34 N. Penn St., York, Pa.

OSCILLATORS—Crystal
Smith Laboratories, Melvin L., 16 Field St., Kane, Pa.

OSCILLATORS—Radio Frequency
Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.

OSCILLATORS—Ultrasonic
Barker & Williamson, 235 Fairfield Ave., Upper Darby, Pa.
Brush Development Co., 3405 Perkins Ave., Cleveland 14, Ohio
Fisher Scientific Co., 717 Forbes St., Pittsburgh 19, Pa.
General Electric Co., Schenectady 5, N. Y.
General Motors Co., Research Labs. Div., 485 W. Milwaukee Ave., Detroit, Mich.

PLASTIC MATERIALS

Phenol Formaldehyde—(Phenolics)
Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.

Resins
Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.

PLATES—Name
Wilson Plastics Div., Wilson Magazine Camera Co., 6022 Media St., Philadelphia 31, Pa.

POTENTIOMETERS
I-T-E Circuit Breaker Co., 19th & Hamilton Sts., Philadelphia 30, Pa.

POTS—Soldering
Trent Co., Harold E., Liverington Ave. at Wilde St., Philadelphia 27, Pa.

POWER SUPPLIES
National Radio Service Co., Reisterstown Rd. at Cold Spring Lane, Baltimore 15, Md.

POWER SUPPLIES—Electronically Regulated
National Radio Service Co., Reisterstown Rd. at Cold Spring Lane, Baltimore, 15, Md.

POWER SUPPLIES—Voltage Regulated
National Radio Service Co., Reisterstown Rd. at Cold Spring Lane, Baltimore, 15, Md.

RECEIVERS—a-m Fixed Frequency
Kinsey Radio Engineering Co., Inc., 1208 Farnam St., Omaha 27, Nebraska

RECEIVERS—Automobile
Kinsey Radio Engineering Co., Inc., 1208 Farnam St., Omaha 27, Nebraska

RECEIVERS—Diversity Reception
Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.

RECEIVERS—Fixed Frequency
Kinsey Radio Engineering Co., Inc., 1208 Farnam St., Omaha 2, Nebraska

RECEIVERS—f-m
Kinsey Radio Engineering Co., Inc., 1208 Farnam St., Omaha 2, Nebraska

RECEIVERS, HOME—a-m Portables
Bowers Battery & Spark Plug Co., Reading, Pa.

RECEIVERS, HOME—a-m Table Models
Bowers Battery & Spark Plug Co., Reading, Pa.
Kinsey Radio Engineering Co., Inc., 1208 Farnam St., Omaha 2, Nebraska

RECEIVERS, HOME—f-m Table Models
Kinsey Radio Engineering Co., Inc., 1208 Farnam St., Omaha 2, Nebraska

RECORDERS—Sound and Phonograph
Presto Recording Corp., 242 West 55th St., New York, N. Y.

RELAYS—Mercury
Mack Electrical Devices Co., Elkins Park, Pa.

RESISTANCE SPECIALTIES
Shallcross Mfg. Co., 10 Jackson Ave., Collingdale, Pa.

RESISTORS—Fixed
I-T-E Circuit Breaker Co., 19th & Hamilton Sts., Philadelphia 30, Pa.

RESISTORS—Wire Wound
Stackpole Carbon Co., St. Marys, Pa.

RHEOSTATS
I-T-E Circuit Breaker Co., 19th & Hamilton Sts., Philadelphia 30, Pa.

RINGS—Oil Sealing
Stackpole Carbon Co., St. Marys, Pa.

SEALS—Carbon
Stackpole Carbon Co., St. Marys, Pa.

SOCKETS—Miniature Sockets & Connectors
Eby, Inc., Hugh H., 18 W. Chelton Ave., Philadelphia 13, Pa.

SOUND SYSTEMS—Complete
Schulmerich Electronics, Inc., Sellersville, Pa.
York Electric and Machine Co., Carillotone Div., 30-34 N. Penn St., York, Pa.

SWITCH PARTS

Contacts
Keystone Carbon Co., St. Marys, Pa.
Stackpole Carbon Co., St. Marys, Pa.

TOOLS—Screwdrivers and Small Insulated
Utica Drop Forge & Tool Corp., 2415 Whitesboro St., Utica 4, N. Y.

TRANSMITTERS—Aircraft
Supreme Transmitter Corp., 280 Ninth Ave., New York 1, N. Y.

TRANSMITTERS—Amateur
Supreme Transmitter Corp., 280 Ninth Ave., New York 1, N. Y.

TRANSMITTERS—f-m
Radio Engineering Labs. Inc., 35-54 36th St., Long Island City, N. Y.

TRANSMITTERS — Fixed Station Communication
Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.

TRANSMITTERS—Marine
Supreme Transmitter Corp., 280 Ninth Ave., New York 1, N. Y.

TRANSMITTERS—Television
Du Mont Labs. Inc., Allen B., 2 Main Ave., Passaic, N. J.

TUBE PARTS

Anodes, Metallized Graphite
Stackpole Carbon Co., St. Marys, Pa.

TURNTABLES—Recording and Transcription
Gray Research & Development Co., 50 N. Central Ave., Elmsford, N. Y.

WASHERS—Insulating
Franklin Fibre-Lamitex Corp., 12th & French Sts., Wilmington, Del.

WIRE—Hookup
Precision Tube Co., 3828 Terrace St., Philadelphia 28, Pa.

NAME and ADDRESS ADDITIONS and CORRECTIONS

Accurate Mfg. Co., Inc., Philadelphia 32, Pa.
American Insulator Corp., New Freedom, Pa.
Atlas Tool and Design Co., Caster & Kensington Avenues, Philadelphia 24, Pa.
Bowers Battery & Spark Plug Co., Reading, Pa.
Burgess Battery Co., Battery Div., Freeport, Ill.
Durez Plastics & Chemicals, Inc., 1921 Walck Rd., North Tonawanda, N. Y.
Electro-Physical Laboratories, Inc., 290-8 Dyckman St., New York 34, N. Y.
General Motors Co., Research Labs. Div., 485 W. Milwaukee Ave., Detroit, Mich.
Guardian Electric Mfg. Co., 1621 West Walnut St., Chicago 12, Ill.
Instrument Electronics, 42-17A Douglaston Parkway, Douglaston, L. I., N. Y.
Kinsey Radio Engineering Co., Inc., 1208 Farnam St., Omaha 27, Nebraska
Lambert Mfg. Co., 639 West 12th St., Erie, Pa.
Mack Electrical Devices Co., Elkins Park, Pa.
Marco Industries, Depew, N. Y.
Moyen, C. P., 908 Chicago Ave., Chicago, Ill.
National Radio Service Co., Reisterstown Rd. at Cold Spring Lane, Baltimore 15, Md.
Northern Radio Co., Inc., 143-145 W. 22nd St., New York 11, N. Y.
Norton Laboratories, Inc., Lockport, N. Y.
Ozalid Products Div., General Aniline & Film Corp., 770 Ansco Rd., Johnson City, N. Y.
Paragon Electric Co., Two Rivers, Wisconsin
Philadelphia Thermometer Co., 4401 N. Sixth St., Philadelphia, Pa.
Phosphor Bronze Smelting Co., 2200 Washington Ave., Philadelphia 46, Pa.
Ripley Co., Inc., P. O. Box 31, Middletown, Conn.
Schulmerich Electronics, Inc., Sellerville, Pa.
Scintilla Magneto Div., Bendix Aviation Corp., Sidney, N. Y.
Seifert, Inc., E. R., 202 South Beech St., Syracuse, N. Y.
Sherman Industrial Electronics Co., Inc., 503 Washington Ave., Belleville 9, N. J.
Simpson Mfg. Co., Inc., Mark, 32-28 49th St., Long Island City 3, N. Y.
Smith Laboratories, Melvin L., 16 Field St., Kane, Pa.
Sun Oil Co., 1608 Walnut St., Philadelphia 3, Pa.
Supreme Transmitter Corp., 280 Ninth Ave., New York 1, N. Y.
Trent Co., Harold E., Liverington Ave. at Wilde St., Philadelphia 27, Pa.
Utica Drop Forge & Tool Corp., 2415 Whitesboro St., Utica 4, N. Y.
Wurlitzer Mfg. Co., North Tonawanda, N. Y.
York Electric & Machine Co., Carillotone Div., 30-34 N. Penn St., York, Pa.

TRADE NAMES ADDITIONS and CORRECTIONS

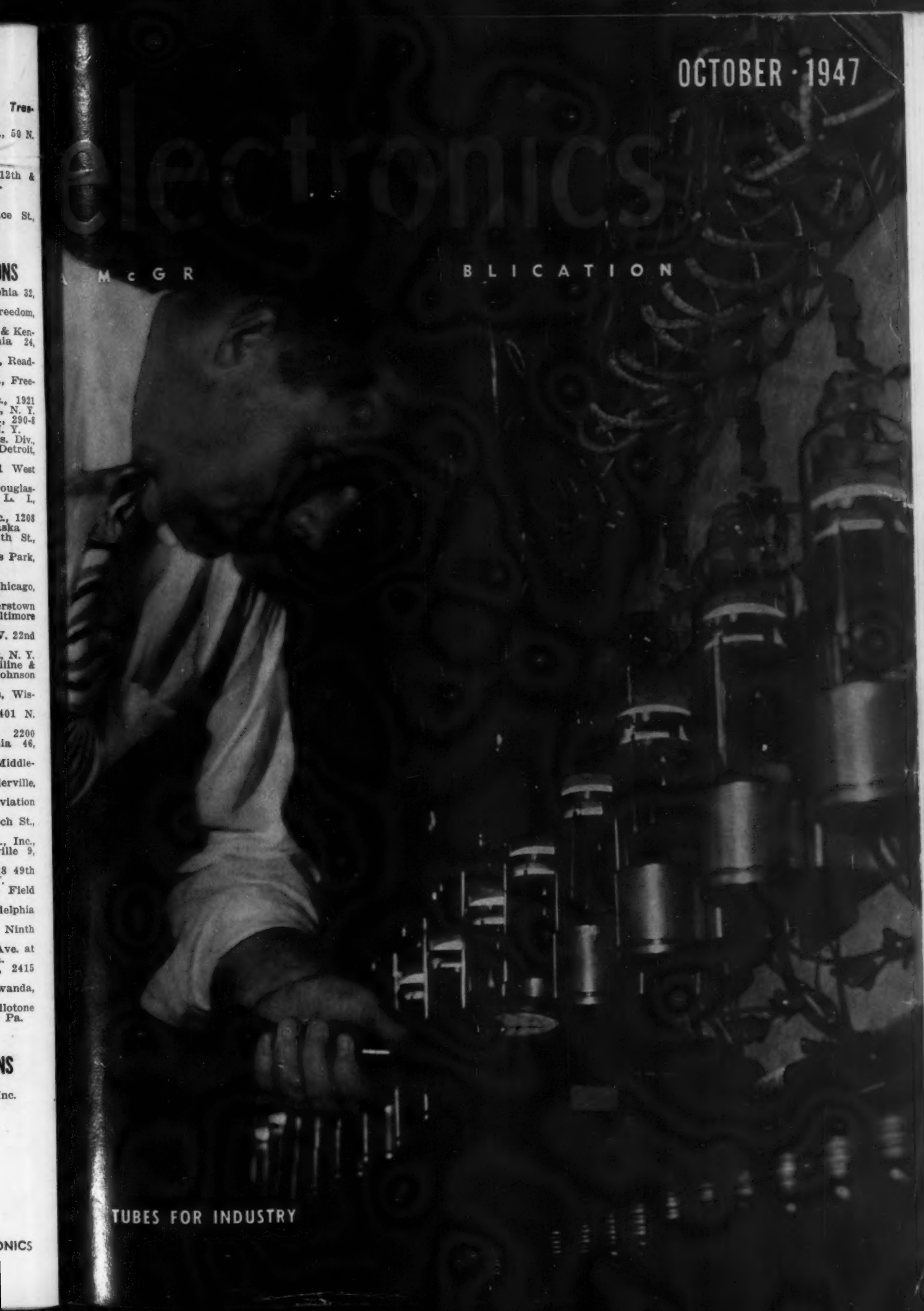
KREC
Kinsey Radio Engineering Co., Inc.

LITTLE-SIX
Burgess Battery Co.
Battery Div.

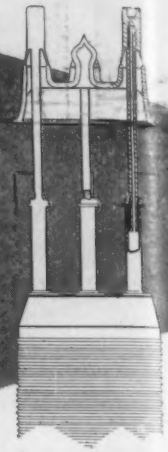
SELSYN
General Electric Co.
Schenectady 5, N. Y.

STROWGER
Automatic Electric Co.

TWIN-SIX
Burgess Battery Co.
Battery Div.

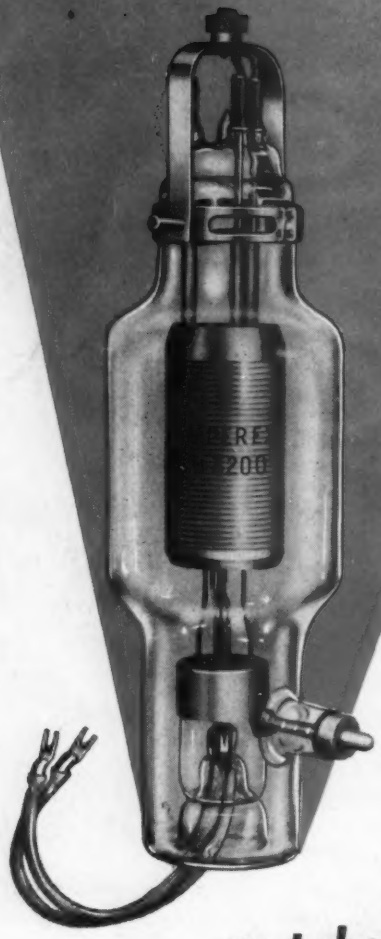

electronics
OCTOBER · 1947
McGR
BLICATION
TUBES FOR INDUSTRY

the LITTLE differences make the BIG differences



Would you say there's a little or a big difference between a winner and a loser? And could a loser be a winner? No, we're not being philosophic . . . we're just thinking about the Amperex ZB3200.

This tube is a winner but that's because it's also a good loser. In tubes, one of the problems is to lose heat efficiently but in such a way that we prevent overheating of the glass-to-metal seals. So even though the plate terminals are that special Amperex copper one-piece construction, feather-edge seals to glass, we wanted to keep the heat away from them. To do this we chose tubular supports; molybdenum next to the anode to withstand the high temperature but with a thin wall to reduce thermal conduction and large outer surface to increase radiation; then a smaller tube of nickel to connect to the one-piece copper terminal, nickel because it is also a good loser in the heat conduction race but has the required strength and rigidity. So we start out with two losers and wind up with a winner.



Now this isn't exactly a "little" difference, but we do have hundreds of little and medium-size differences in the design and construction of the many transmitting, rectifying and special purpose tubes that comprise the extensive Amperex line.

re-tube with Amperex

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OCTOBER • 1947

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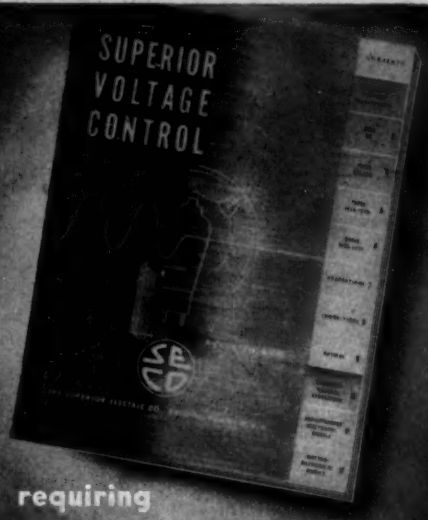
New Bulletin 547

Features VOLTAGE CONTROL by SUPERIOR ELECTRIC

All the information required for selection of the correct voltage control apparatus is included—

FOR INSTANCE: If you have an application requiring variable voltage to a load of about 1 KVA turn to page 4.



POWERSTAT TYPES 116 - 216

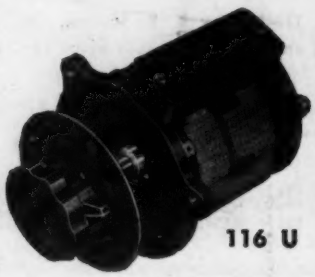
POWERSTATS types 116 and 216 are alike in appearance and physical size . . type 116 operates on 115 volts to deliver 0 - 135 volts, 7.5 amperes — type 216 operates on 230 volts to deliver 0 - 270 volts, 3.0 amperes. Incorporated into each type is protective screening, output receptacle, "on-off" switch, input cord-plug, and fuse protection.

When back-of-panel mounted as part of other equipment, these POWERSTATS are furnished uncased as shown in the illustration of type 116U. Ganged assemblies are supplied for open-delta or wye connected three phase duty. All single and multiple types 116 and 216 can be adapted to motor-operation. Type MT116 is a typical motor-driven unit.

116-2D

116

116 U



MT 116

To get your copy of Bulletin 547 . . . and for more information on POWERSTATS types 116 and 216, write

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